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Adaptive Fuzzy Tracking Control of Uncertain Nonlinear Multi-Agent Systems with Unknown Control Directions and a Dead-Zone Fault

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Abstract: In this paper, a class of uncertain nonlinear multi-agent systems with unknown control directions and a dead-zone fault is addressed, where unknown control gains exist in each subsystem. In terms of the approximation characteristic of a fuzzy logic system, it is used to approximate uncertain nonlinear dynamics, and then the relevant adaptive control laws are designed. Considering the presence of unknown control directions and a dead-zone fault, the Nussbaum gain function technique is introduced to design the intermediate control law and the adaptive fuzzy control law. A theoretical analysis shows that the tracking control problem of the given multi-agent systems can be effectively solved through the application of the proposed adaptive fuzzy control law and the tracking errors can converge to a small neighborhood of zero through an adjustment of the relevant parameters. Finally, the effectiveness of the theoretical analysis results is verified by two simulation cases.

Keywords: uncertain nonlinear multi-agent systems; unknown control direction; dead-zone fault; fuzzy logic system

MSC: 93D21; 93D50

Citation: Deng, X.; Zhang, X. Adaptive Fuzzy Tracking Control of Uncertain Nonlinear Multi-Agent Systems with Unknown Control Directions and a Dead-Zone Fault. *Mathematics* **2022**, *10*, 2655. <https://doi.org/10.3390/math10152655>

Academic Editors: Jan Awrejcewicz, José A. Tenreiro Machado, José M. Vega, Hari Mohan Srivastava, Ying-Cheng Lai, Hamed Farokhi and Roman Starosta

Received: 21 June 2022

Accepted: 26 July 2022

Published: 28 July 2022

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1. Introduction

In recent years, the control problems of multi-agent systems have been extensively studied in complex systems such as multi-UAV systems [1], multi-sensor network systems [2], and microgrid systems [3]. In order to solve the control problems of multi-agent systems, many control strategies have been proposed and put into practice, achieving good control. For example, the iterative learning control law was proposed in [4,5], where the tracking problem of nonlinear uncertain multi-agent systems was solved. In [6], the authors designed a finite time adaptive neural network controller using the command filter control technology, where the guaranteed cost control of nonstrict-feedback uncertain multi-agent systems with input nonlinearity was realized. Moreover, adaptive consensus control laws, which were used to study the tracking control of heterogeneous nonlinear multi-agent systems, were presented in [7,8]. In addition, the pulse consensus control law [9] and the event-triggered consensus control law [10,11] were successfully applied for the control of multi-agent systems.

However, it should be noted that in some actual systems, the sign of control gains is sometimes not predicted in advance [12–14], which leads to the problem of an unknown

control direction. This makes it impossible to directly apply some existing achievements, such as those mentioned in [15–17], and many scholars have carried out research on this topic. In [18], to solve the control problem of an uncertain system with unknown control direction and unknown input power, an adaptive parameterized controller was designed to ensure that the signals of the closed-loop system were globally bounded. In [19], an adaptive robust tracking control law was developed for a class of uncertain nonlinear systems with unknown control directions, and the specified tracking control was achieved. Furthermore, on the basis of unknown control directions, the results of [13,20–22] further considered the existence of non-ideal conditions, such as time delay, input nonlinearity, output saturation, and actuator failure, achieving the desired control level by applying the proposed control laws. Accordingly, the control problems of multi-agent systems with unknown control directions have also been studied to a certain extent. In [23], for the nonlinear multi-agent systems with unknown non-uniform control directions, the authors proposed the distributed control law to ensure that the consensus tracking problem was solved. Based on the hybrid Nussbaum control method, the consensus tracking control of high-energy nonlinear multi-agent systems with unknown control directions was studied in [24]. Additionally, for the consensus control problem of strict-feedback nonlinear leaderless multi-agent systems with unknown control directions, a decentralized inversion adaptive control law was presented in [25], where the local error surface of multi-agent systems remained bounded and converged to zero. Moreover, compared with [25], the distributed adaptive control law designed in [26] combined the conversion mechanism and did not consider the Nussbaum gain control technology. From the above description, although many achievements have been made in the research on the problem of unknown control directions, there remains little discussion on the existence of actuator fault with unknown control directions.

As an important part of multi-agent systems, when the actuator breaks down it inevitably causes difficulties in the control of the system and even leads to the failure of the system. The distributed consensus control law and the adaptive cooperative control law were proposed in [27–29], where the consensus tracking control problems of nonlinear multi-agent systems with dead-zone inputs were solved. In [30], a distributed adaptive control strategy, which combined the backstepping control technology and the Nussbaum gain function technology, was design to achieve the progressive tracking control of nonlinear multi-agent systems with backlash such as hysteresis faults. In addition, the control problems of multi-agent systems with input saturation, input hysteresis, input quantization, and time-varying faults have also been studied in great depth [31–34]. However, as a kind of input nonlinearity, the occurrence of an actuator dead-zone fault can easily lead to a decline in system control performance and even to the instability of the closed-loop system. Furthermore, the existence of uncertainty causes difficulties in the control of the system. Some control strategies, such as fuzzy active disturbance rejection control [35], indirect adaptive iterative learning control [36], and adaptive sliding mode control [37], have been proposed and applied by researchers. For multi-agent systems, it is also important to consider the existence of uncertainty. Therefore, it is practical to study an uncertain multi-agent system with unknown control directions and a dead-zone fault.

Motivated by the above-mentioned discussions, a class of uncertain multi-agent systems with unknown control directions and an actuator dead-zone fault is considered in this paper. To solve the tracking control problem of the multi-agent systems, a fuzzy logic system and Nussbaum gain function technology are simultaneously considered, and then the adaptive control law, the intermediate control law, and the adaptive fuzzy control law are designed. To this end, the main contributions of this paper can be summarized as follows:

- (i) The control problem of uncertain multi-agent systems with unknown control directions and an actuator dead-zone fault is studied, where unknown control gains

exist in each subsystem of the multi-agent systems. Compared with [6,8,9,23], the system model considered in this paper is more general.

- (ii) Considering the approximation characteristics of the fuzzy logic system, the unknown nonlinear dynamics in the analysis process are approximated, and the adaptive control laws are designed. Compared with [25,26], the analysis process is effectively simplified.
- (iii) The Nussbaum gain function technology is used in the design of the intermediate control law and of the adaptive fuzzy control law to solve the desired tracking control problem. Compared with [10,24,26,27], the control law designed in this paper can meet the control requirements when the control directions are unknown and coexists with the actuator dead-zone fault.
- (iv) Based on the designed Lyapunov function, the effectiveness of the proposed control law is proven. The simulation results show that the tracking errors can finally converge to a small neighborhood of zero following adjustments to the relevant parameters.

The rest of this paper unfolds as follows: Section 2 introduces the multi-agent systems model, and some preliminaries are given in this section. The main results are provided in Section 3, which mainly involves the design of the adaptive fuzzy control law and the discussion of stability. In Section 4, the simulation analysis is described to illustrate the effectiveness of theoretical results, and the conclusions are briefly drawn in Section 5.

Notations: $|A|$ stands for the absolute value of constant A . $X_{i,\min}$ and $X_{i,\max}$ are the minimum and maximum values of variable X_i , respectively. $diag\{x_1, \dots, x_n\}$ denotes a diagonal matrix with diagonal elements x_1, \dots, x_n . $(\cdot)^T$ represents the transposition operation. \hat{W} is the estimate of W , and $\tilde{W} = W - \hat{W}$ stands for the estimation error. $\lambda_{\min}(\cdot)$ and $\lambda_{\max}(\cdot)$ represent the smallest and largest eigenvalues of matrix (\cdot) .

2. Problem Formulation and Preliminaries

In this section, the problem formulation is provided, and some preliminaries, including graph theory, the fuzzy logic system, and some lemmas, are provided for a subsequent analysis.

2.1. Problem Formulation

Consider a class of uncertain nonlinear multi-agent systems with unknown control directions and a dead-zone fault, which is composed of one leader agent and n follower agents. The dynamics of follower agent i is described as

$$\begin{cases} \dot{x}_i(t) = a_{i1}v_i(t) + f_i(x_i(t), v_i(t)) + \Delta_{i1}(t) \\ \dot{v}_i(t) = a_{i2}u_i^F(t) + g_i(x_i(t), v_i(t)) + \Delta_{i2}(t) \end{cases} \quad (1)$$

where $i = 1, \dots, n$, $x_i(t)$ and $v_i(t)$ represent the position vector and velocity vector; $g_i(x_i(t), v_i(t))$ and $f_i(x_i(t), v_i(t))$ are unknown smooth nonlinear functions, and for convenience, the functions $g_i(x_i(t), v_i(t))$ and $f_i(x_i(t), v_i(t))$ are denoted by g_i and f_i , respectively; a_{i1} and a_{i2} represent non-zero unknown constants; $\Delta_{i1}(t)$ and $\Delta_{i2}(t)$ are uncertain dynamics; and $u_i^F(t)$ represents the system input and is supposedly affected by the dead-zone fault. According to [29], the model of a dead-zone fault is:

$$u_i^F(t) = \begin{cases} \omega(u_i(t) - b_{ir}), & u_i(t) \geq b_{ir} \\ 0, & -b_{il} < u_i(t) < b_{ir} \\ \omega(u_i(t) + b_{il}), & u_i(t) \leq -b_{il} \end{cases} \quad (2)$$

where $\omega > 0$ is an unknown bounded constant that represents the slope of the dead-zone; $b_{il} > 0$ and $b_{ir} > 0$ represent the left and right breakpoints of the dead-zone, respectively. By applying the mean value theorem, (2) can be rewritten as:

$$u_i^F(t) = \omega u_i(t) + \phi_i(t) \tag{3}$$

where $\phi_i(t)$ is a bounded function and satisfies $|\phi_i(t)| \leq \bar{\phi}$, and the expression of $\phi_i(t)$ is shown as:

$$\phi_i(t) = \begin{cases} -\omega b_{ir}, & u_i(t) \geq b_{ir} \\ -\omega u_i(t), & -b_{il} < u_i(t) < b_{ir} \\ \omega b_{il}, & u_i(t) \leq -b_{il} \end{cases} \tag{4}$$

The objective of this paper is to design an adaptive fuzzy control law $u_i(t)$ for the system (1), so that the output of each follower agent can track the trajectory of the leader agent when the control directions are unknown and an actuator dead-zone fault occurs, and the tracking error of each follower agent can converge to a small neighborhood of zero.

Assumption 1. The unknown constants a_{i1} and a_{i2} are bounded; that is, there exist $0 < a_{1,\min} \leq |a_{i1}| \leq a_{1,\max}$ and $0 < a_{2,\min} \leq |a_{i2}| \leq a_{2,\max}$. To not lose generality, we further assume that $a_{1,\min} \leq a_{i1} \leq a_{1,\max}$ and $a_{2,\min} \leq a_{i2} \leq a_{2,\max}$.

Assumption 2. The uncertain dynamics $\Delta_{i1}(t)$ and $\Delta_{i2}(t)$ are bounded and satisfy $|\Delta_{i1}(t)| \leq \Delta_{i1}^*$ and $|\Delta_{i2}(t)| \leq \Delta_{i2}^*$.

2.2. Graph Theory

Let $\mathcal{G} = (\mathcal{V}, \mathcal{E}, \mathcal{A})$ denote a directed graph with n nodes, where $\mathcal{V} = \{v_1, \dots, v_n\}$ is the set of vertices, $\mathcal{E} = \{(i, j), i, j \in \mathcal{V}, \text{ and } i \neq j\}$ is the set of edges, and $\mathcal{A} = [a_{ij}] \in \mathbb{R}^{n \times n}$ is the weighted adjacency matrix of \mathcal{G} . If there is an edge between node i and j , then $a_{ij} = a_{ji} \neq 0$, and otherwise, $a_{ij} = a_{ji} = 0$. The set of neighbors of node i is denoted by $\mathcal{N}_i = \{v_j : (v_i, v_j) \in \mathcal{E}\}$. The Laplacian matrix of \mathcal{G} is denoted by $\mathcal{L} = \mathcal{D} - \mathcal{A}$, where $\mathcal{D} = \text{diag}\{d_1, \dots, d_n\}$ with $d_i = \sum_{j=1}^n a_{ij}$. The graph \mathcal{G} is connected if there is a path between any two vertices.

An extended graph is defined as $\bar{\mathcal{G}} = (\bar{\mathcal{V}}, \bar{\mathcal{E}})$, which is associated with the leader agent and follower agents. Let the leader adjacency matrix be $\mathcal{B} = \text{diag}\{b_1, \dots, b_n\}$, and if the follower agent i obtains the information of leader agent, then $b_i = 1$; otherwise, $b_i = 0$.

Assumption 3 [38]. The directed graph $\bar{\mathcal{G}}$ contains a spanning tree, and the leader node is the root node.

2.3. Fuzzy Logic System

In the subsequent analysis, the fuzzy logic system is considered to approximate unknown uncertain dynamics. The fuzzy rule base is composed of “if-then” rules in the following form:

$$R^l : \text{if } x_1 \text{ is } F_1^l, \dots, \text{ and } x_n \text{ is } F_n^l, \text{ then } h \text{ is } G^l, \quad l = 1, \dots, M,$$

where x_i , $i = 1, \dots, n$, and h are the fuzzy logic system’s input and output, respectively; M is the total number of “if-then” rules; F_1^l, \dots, F_n^l and G^l are fuzzy

sets for linguistic variables. Additionally, by applying singleton fuzzifier, product inference, and a defuzzifier [32,39], the fuzzy logic system can be formulated as:

$$h(\mathbf{x}) = \frac{\sum_{l=1}^M \bar{h}_l \prod_{i=1}^n \mu_{F_i^l}(x_i)}{\sum_{l=1}^M \left[\prod_{i=1}^n \mu_{F_i^l}(x_i) \right]} \tag{5}$$

where $\bar{h}_l = \max_{h \in R} \mu_{G^l}(h)$.

If $\mathbf{W}^T = [\bar{h}_1, \dots, \bar{h}_M] = [w_1, \dots, w_M]$ and $\boldsymbol{\varphi}(\mathbf{x}) = [\varphi_1(\mathbf{x}), \dots, \varphi_M(\mathbf{x})]^T$, then the vector from the fuzzy logic system (5) can be written as:

$$h(\mathbf{x}) = \mathbf{W}^T \boldsymbol{\varphi}(\mathbf{x}) \tag{6}$$

where $\varphi_l(\mathbf{x}) = \left(\prod_{i=1}^n \mu_{F_i^l}(x_i) \right) / \left(\sum_{l=1}^M \left[\prod_{i=1}^n \mu_{F_i^l}(x_i) \right] \right)$, $l = 1, \dots, M$; $\mathbf{x} = [x_1, \dots, x_n]^T$ is the input of the fuzzy logic system.

Lemma 1 [21]. *For any continuous function $f(\mathbf{x})$ defined on a compact set Ω and any given positive constant ε , there exists a fuzzy logic system $f^*(\mathbf{x}) = \mathbf{W}^{*T} \boldsymbol{\varphi}(\mathbf{x})$ in the form of (15) such that*

$$\sup_{\mathbf{x} \in \Omega} |f^*(\mathbf{x}) - f(\mathbf{W}, \boldsymbol{\varphi})| \leq \varepsilon \tag{7}$$

where \mathbf{W}^* is the ideal parameter vector, and ε is the approximation accuracy and can be arbitrarily small.

2.4. Definition and Lemmas

Definition 1 [21]. *The smooth continuous function $N(\kappa)$ is called the Nussbaum gain function if the following properties hold:*

$$\begin{cases} \limsup_{s \rightarrow \infty} \frac{1}{s} \int_0^s N(\kappa) d\kappa = +\infty \\ \liminf_{s \rightarrow \infty} \frac{1}{s} \int_0^s N(\kappa) d\kappa = -\infty \end{cases} \tag{8}$$

Lemma 2 [21]. *Let $V(t)$ and $\kappa(t)$ be smooth functions defined on $[0, t_f)$ with $V(t) \geq 0$, and $N(\kappa)$ be a Nussbaum gain function. If the following inequality holds:*

$$V(t) \leq e^{-k_0 t} \int_0^t e^{k_0 \tau} (G(\bar{\mathbf{x}}_n) N(\kappa) + 1) \dot{\kappa}(\tau) d\tau + c_0, t \in [0, t_f) \tag{9}$$

then $V(t)$, $\kappa(t)$, and $\int_0^t e^{k_0 \tau} (G(\bar{\mathbf{x}}_n) N(\kappa) + 1) \dot{\kappa}(\tau) d\tau$ are bounded by $[0, t_f)$, where $G(\bar{\mathbf{x}}_n)$ satisfies $G_m \leq |G(\bar{\mathbf{x}}_n)| \leq G_M$ with $\bar{\mathbf{x}}_n$ being the system state vector, and G_m, G_M, c_0 , and k_0 are positive constants.

Lemma 3 [15]. *For any $x \in R$ and $y \in R$, the following inequality holds:*

$$xy \leq \frac{\zeta^p}{p} |x|^p + \frac{1}{q\zeta^q} |y|^q \tag{10}$$

where $\zeta > 0$, $p > 1$, $q > 1$, and $(p-1)(q-1) = 1$.

Lemma 4 [5]. For any $b \in \mathbb{R}$ and $\mathcal{G} > 0$, the hyperbolic tangent function satisfies:

$$0 \leq |b| - b \tanh\left(\frac{b}{\mathcal{G}}\right) \leq 0.2785\mathcal{G} \tag{11}$$

3. Adaptive Control Law Design and Stability Analysis

In order to solve the tracking control problem of an uncertain nonlinear multi-agent system (1) with unknown control directions and a dead-zone fault, in this section, the fuzzy logic system and Nussbaum gain function technology are introduced to design the adaptive fuzzy control law, intermediate control law, and adaptive control laws.

3.1. Adaptive Fuzzy Control Law Design

Considering the system (1), the consensus tracking error z_{i1} and velocity tracking error z_{i2} , respectively, are defined by:

$$z_{i1} = \sum_{j \in \mathcal{N}_i} a_{ij}(x_i - x_j) + b_i(x_i - x_d) \tag{12}$$

$$z_{i2} = v_i - \alpha_i \tag{13}$$

where x_d represents the trajectory of the leader agent, and α_i is the intermediate control law to be designed.

Let $\delta_{xi} = x_i - x_d$ stand for the position tracking error of the i th follower agent, then the vector form of (12) is written as:

$$Z_1 = \mathcal{H}\delta_x \tag{14}$$

where $Z_1 = [z_{11}, \dots, z_{n1}]^T$, $\mathcal{H} = \mathcal{L} + \mathcal{B}$, $\delta_x = [\delta_{x1}, \dots, \delta_{xn}]^T = \mathbf{x} - \mathbf{1} \otimes x_d$ with $\mathbf{x} = [x_1, \dots, x_n]^T$ and $\mathbf{1} = [1, \dots, 1]^T$.

Define the candidate Lyapunov function V_1 as:

$$V_1 = \frac{1}{2} Z_1^T \mathcal{H}^{-1} Z_1 \tag{15}$$

According to (12)–(14), the time derivative of V_1 is:

$$\begin{aligned} \dot{V}_1 &= Z_1^T \mathcal{H}^{-1} \mathcal{H} \dot{\delta}_x \\ &= \sum_{i=1}^n [a_{i1} z_{i1} \alpha_i + a_{i1} z_{i1} z_{i2} + z_{i1} (f_i + \Delta_{i1} - \dot{x}_d)] \end{aligned} \tag{16}$$

Considering Lemma 3 and letting $\zeta = 1$, $p = 2$, $q = 2$, $x = a_{i1} z_{i1}$, and $y = z_{i2}$, we have:

$$a_{i1} z_{i1} z_{i2} \leq \frac{(a_{i1} z_{i1})^2}{2} + \frac{z_{i2}^2}{2} \tag{17}$$

Substituting (17) into (16) yields:

$$\dot{V}_1 \leq \sum_{i=1}^n \left[a_{i1} z_{i1} \alpha_i + z_{i1} (F_{i1} + \Delta_{i1} - \dot{x}_d) + \frac{z_{i2}^2}{2} \right] \tag{18}$$

where $F_{i1} = a_{i1}^2 z_{i1} / 2 + f_i$, a fuzzy logic system is introduced to approximate F_{i1} , then we obtain:

$$F_{i1} = (\mathbf{W}_{i1}^*)^T \boldsymbol{\varphi}(\mathbf{X}_{i1}) + \varepsilon_{i1} \tag{19}$$

where $\mathbf{X}_{i1} = [x_i, v_i, x_j, x_d]^T$, $i = 1, \dots, n$, $j \in \mathcal{N}_i$, and ε_{i1} is the approximation error.

Noting Assumption 2 and Lemma 1, then there exists

$$|\varepsilon_{i1} + \Delta_{i1}| \leq \varepsilon_{i1}^* \tag{20}$$

where ε_{i1}^* is an unknown positive constant.

Let

$$(\mathbf{W}_{i1})^T \boldsymbol{\Phi}(\mathbf{X}_{i1}) = (\mathbf{W}_{i1}^*)^T \boldsymbol{\varphi}(\mathbf{X}_{i1}) + \varepsilon_{i1}^* \tanh\left(\frac{\varepsilon_{i1}^* z_{i1}}{\vartheta}\right) - \dot{x}_d \tag{21}$$

where

$$\mathbf{W}_{i1}^T = [\mathbf{W}_{i1}^*, 1] \tag{22}$$

$$\boldsymbol{\Phi}(\mathbf{X}_{i1}) = \left[\boldsymbol{\varphi}(\mathbf{X}_{i1}), \delta_{i1}^* \tanh\left(\frac{\delta_{i1}^* z_{i1}}{\vartheta}\right) - \dot{x}_d \right]^T \tag{23}$$

Substituting (19) and (21) into (18), we have:

$$\begin{aligned} \dot{V}_1 &\leq \sum_{i=1}^n \left[a_{i1} z_{i1} \alpha_i + z_{i1} \left((\mathbf{W}_{i1}^*)^T \boldsymbol{\varphi}(\mathbf{X}_{i1}) + \varepsilon_{i1} + \Delta_{i1} - \dot{x}_d \right) + \frac{z_{i2}^2}{2} \right] \\ &= \sum_{i=1}^n \left[a_{i1} z_{i1} \alpha_i + z_{i1} \left((\mathbf{W}_{i1})^T \boldsymbol{\Phi}(\mathbf{X}_{i1}) + \varepsilon_{i1} + \Delta_{i1} - \varepsilon_{i1}^* \tanh\left(\frac{\varepsilon_{i1}^* z_{i1}}{\vartheta}\right) \right) + \frac{z_{i2}^2}{2} \right] \end{aligned} \tag{24}$$

The intermediate control law α_i and adaptive control law $\dot{\kappa}_{i1}$ are designed as follows:

$$\alpha_i = N(\kappa_{i1}) \left((\hat{\mathbf{W}}_{i1})^T \boldsymbol{\Phi}(\mathbf{X}_{i1}) + \rho_{i1} z_{i1} \right) \tag{25}$$

$$\dot{\kappa}_{i1} = z_{i1} \left((\hat{\mathbf{W}}_{i1})^T \boldsymbol{\Phi}(\mathbf{X}_{i1}) + \rho_{i1} z_{i1} \right) \tag{26}$$

where $\hat{\mathbf{W}}_{i1}$ represents the estimate of \mathbf{W}_{i1} , and ρ_{i1} is the positive constant to be designed.

Considering Lemma 4, one obtains $(\varepsilon_{i1} + \Delta_{i1}) - \varepsilon_{i1}^* \tanh(\varepsilon_{i1}^* z_{i1} / \vartheta) \leq 0.2785\vartheta$; then, substituting (25) and (26) into (24), we have:

$$\begin{aligned} \dot{V}_1 &\leq \sum_{i=1}^n \left[(a_{i1} N(\kappa_{i1}) \dot{\kappa}_{i1} + \dot{\kappa}_{i1}) - \rho_{i1} z_{i1}^2 + z_{i1} (\tilde{\mathbf{W}}_{i1})^T \boldsymbol{\Phi}(\mathbf{X}_{i1}) + z_{i1} (\varepsilon_{i1} + \Delta_{i1}) - z_{i1} \varepsilon_{i1}^* \tanh\left(\frac{\varepsilon_{i1}^* z_{i1}}{\vartheta}\right) + \frac{z_{i2}^2}{2} \right] \\ &\leq \sum_{i=1}^n \left[(a_{i1} N(\kappa_{i1}) \dot{\kappa}_{i1} + \dot{\kappa}_{i1}) - \rho_{i1} z_{i1}^2 + z_{i1} (\tilde{\mathbf{W}}_{i1})^T \boldsymbol{\Phi}(\mathbf{X}_{i1}) + \frac{z_{i2}^2}{2} + 0.2785\vartheta \right] \end{aligned} \tag{27}$$

where $\tilde{\mathbf{W}}_{i1} = \mathbf{W}_{i1} - \hat{\mathbf{W}}_{i1}$.

With reference to the description in [40], the adaptive control law $\dot{\hat{\mathbf{W}}}_{i1}$ can be designed as:

$$\dot{\hat{\mathbf{W}}}_{i1} = \gamma_{i1} \left(\boldsymbol{\Phi}(\mathbf{X}_{i1}) z_{i1} - \eta_{i1} \hat{\mathbf{W}}_{i1} \right) \tag{28}$$

where γ_{i1} and η_{i1} are designed as positive constants.

Furthermore, the candidate Lyapunov function V_2 is defined as:

$$V_2 = \frac{1}{2} \mathbf{Z}_2^T \mathbf{Z}_2 \tag{29}$$

where $Z_2 = [z_{i2}, \dots, z_{n2}]^T$.

With reference to (3) and (13), the time derivative of V_2 is:

$$\begin{aligned} \dot{V}_2 &= \sum_{i=1}^n z_{i2} (\dot{v}_i - \dot{\alpha}_i) \\ &= \sum_{i=1}^n z_{i2} (a_{i2}(\omega u_i(t) + \phi_i(t)) + g_i + \Delta_{i2} - \dot{\alpha}_i) \\ &= \sum_{i=1}^n (\omega a_{i2} z_{i2} u_i(t) + a_{i2} \phi_i(t) z_{i2} + z_{i2} (g_i + \Delta_{i2}) - z_{i2} \dot{\alpha}_i) \end{aligned} \tag{30}$$

Similarly, considering Lemma (3), we have:

$$a_{i2} \phi_i(t) z_{i2} \leq \frac{(a_{i2} z_{i2})^2}{2} + \frac{\phi_i^2(t)}{2} \tag{31}$$

Substituting (31) into (30) obtains:

$$\dot{V}_2 \leq \sum_{i=1}^n \left(\omega a_{i2} z_{i2} u_i(t) + z_{i2} (F_{i2} + \Delta_{i2}) - z_{i2} \dot{\alpha}_i + \frac{\phi_i^2(t)}{2} \right) \tag{32}$$

where $F_{i2} = a_{i2}^2 z_{i2} / 2 + g_i$, and a fuzzy logic system is introduced to approximate F_{i2} ; then, we obtain:

$$F_{i2} = (\mathbf{W}_{i2}^*)^T \boldsymbol{\varphi}(\mathbf{X}_{i2}) + \varepsilon_{i2} \tag{33}$$

where $\mathbf{X}_{i2} = [x_i, v_i, x_j, x_d]^T$, $i = 1, \dots, n$, $j \in \mathcal{N}_i$, and ε_{i2} stands for the approximation error.

Noting Assumption 2 and Lemma 1, we have:

$$|\varepsilon_{i2} + \Delta_{i2}| \leq \varepsilon_{i2}^* \tag{34}$$

where ε_{i2}^* is an unknown positive constant.

Let

$$(\mathbf{W}_{i2})^T \boldsymbol{\Phi}(\mathbf{X}_{i2}) = (\mathbf{W}_{i2}^*)^T \boldsymbol{\varphi}(\mathbf{X}_{i2}) + \varepsilon_{i2}^* \tanh\left(\frac{\varepsilon_{i2}^* z_{i2}}{g}\right) \tag{35}$$

where

$$\mathbf{W}_{i2}^T = [\mathbf{W}_{i2}^*, 1] \tag{36}$$

$$\boldsymbol{\Phi}(\mathbf{X}_{i2}) = \left[\boldsymbol{\varphi}(\mathbf{X}_{i2}), \delta_{i2}^* \tanh\left(\frac{\delta_{i2}^* z_{i2}}{g}\right) \right]^T \tag{37}$$

Furthermore, substituting (33) and (35) into (32) yields:

$$\begin{aligned} \dot{V}_2 &\leq \sum_{i=1}^n \left(\omega a_{i2} z_{i2} u_i(t) + z_{i2} ((\mathbf{W}_{i2}^*)^T \boldsymbol{\varphi}(\mathbf{X}_{i2}) + \varepsilon_{i2} + \Delta_{i2}) - z_{i2} \dot{\alpha}_i + \frac{\phi_i^2(t)}{2} \right) \\ &= \sum_{i=1}^n \left[\omega a_{i2} z_{i2} u_i + z_{i2} \left((\mathbf{W}_{i2})^T \boldsymbol{\Phi}(\mathbf{X}_{i2}) + (\varepsilon_{i2} + \Delta_{i2}) - \varepsilon_{i2}^* \tanh\left(\frac{\varepsilon_{i2}^* z_{i2}}{g}\right) \right) - z_{i2} \dot{\alpha}_i + \frac{\phi_i^2(t)}{2} \right] \end{aligned} \tag{38}$$

The adaptive fuzzy control law $u_i(t)$ and adaptive control law $\hat{\kappa}_{i2}$ are designed as follows:

$$u_i(t) = N(\kappa_{i2}) \left((\hat{\mathbf{W}}_{i2})^T \boldsymbol{\Phi}(\mathbf{X}_{i2}) + \frac{z_{i2}}{2} + \rho_{i2} z_{i2} - \dot{\alpha}_i \right) \tag{39}$$

$$\dot{\kappa}_{i2} = z_{i2} \left((\hat{W}_{i2})^T \Phi(X_{i2}) + \frac{z_{i2}}{2} + \rho_{i2} z_{i2} - \dot{\alpha}_i \right) \tag{40}$$

where \hat{W}_{i2} represents the estimate of W_{i2} , and ρ_{i2} is the positive constant to be designed.

By substituting (39) and (50) into (38), and considering Lemma 4, we have:

$$\dot{V}_2 \leq \sum_{i=1}^n \left(\omega a_{i2} N(\kappa_{i2}) \dot{\kappa}_{i2} + \dot{\kappa}_{i2} - \rho_{i2} z_{i2}^2 + z_{i2} (\tilde{W}_{i2})^T \Phi(X_{i2}) - \frac{z_{i2}^2}{2} + \frac{\phi_i^2(t)}{2} + 0.27859 \right) \tag{41}$$

where $\tilde{W}_{i2} = W_{i2} - \hat{W}_{i2}$.

Similarly, we design the adaptive control law $\dot{\hat{W}}_{i2}$ as:

$$\dot{\hat{W}}_{i2} = \gamma_{i2} \left(\Phi(X_{i2}) z_{i2} - \eta_{i2} \hat{W}_{i2} \right) \tag{42}$$

where γ_{i2} and η_{i2} are the designed positive constants.

3.2. Stability Analysis

In order to verify the validity of the proposed adaptive fuzzy control law, the following theorem is provided.

Theorem 1. Consider the uncertain nonlinear multi-agent systems (1) with unknown control directions and dead-zone fault under Assumptions 1–3, and the intermediate control law (25), adaptive control laws (26), (28), (40), and (42), and adaptive fuzzy control law (39) are applied such that the system (1) can track the trajectory of the leader agent, and the tracking errors of all follower agents finally converge to a small neighborhood of zero.

Proof. We design the Lyapunov function V as:

$$V = V_1 + V_2 + \sum_{i=1}^n \frac{(\tilde{W}_{i1})^T \tilde{W}_{i1}}{2\gamma_{i1}} + \sum_{i=1}^n \frac{(\tilde{W}_{i2})^T \tilde{W}_{i2}}{2\gamma_{i2}} \tag{43}$$

In combination with (27), (28), (41), and (42), and considering $\dot{\tilde{W}}_{i1} = -\dot{\hat{W}}_{i1}$ and $\dot{\tilde{W}}_{i2} = -\dot{\hat{W}}_{i2}$, the time derivative of V is given as:

$$\begin{aligned} \dot{V} \leq & \sum_{i=1}^n (a_{i1} N(\kappa_{i1}) \dot{\kappa}_{i1} + \dot{\kappa}_{i1}) + \sum_{i=1}^n (\omega a_{i2} N(\kappa_{i2}) \dot{\kappa}_{i2} + \dot{\kappa}_{i2}) - \sum_{i=1}^n (\rho_{i1} z_{i1}^2 + \rho_{i2} z_{i2}^2) \\ & + \sum_{i=1}^n \eta_{i1} (\tilde{W}_{i1})^T \hat{W}_{i1} + \sum_{i=1}^n \eta_{i2} (\tilde{W}_{i2})^T \hat{W}_{i2} + \sum_{i=1}^n D_i \end{aligned} \tag{44}$$

where $D_i = \phi_i^2(t)/2 + 0.5579$.

Due to

$$(\tilde{W}_{i1})^T \hat{W}_{i1} \leq -\frac{(\tilde{W}_{i1})^T \tilde{W}_{i1}}{2} + \frac{(W_{i1})^T W_{i1}}{2} \tag{45}$$

$$(\tilde{W}_{i2})^T \hat{W}_{i2} \leq -\frac{(\tilde{W}_{i2})^T \tilde{W}_{i2}}{2} + \frac{(W_{i2})^T W_{i2}}{2} \tag{46}$$

By substituting (45) and (46) into (44), we obtain:

$$\begin{aligned} \dot{V} \leq & \sum_{i=1}^n (a_{i1} N(\kappa_{i1}) \dot{\kappa}_{i1} + \dot{\kappa}_{i1}) + \sum_{i=1}^n (\omega a_{i2} N(\kappa_{i2}) \dot{\kappa}_{i2} + \dot{\kappa}_{i2}) - \sum_{i=1}^n (\rho_{i1} z_{i1}^2 + \rho_{i2} z_{i2}^2) \\ & - \frac{1}{2} \sum_{i=1}^n \eta_{i1} (\tilde{W}_{i1})^T \tilde{W}_{i1} - \frac{1}{2} \sum_{i=1}^n \eta_{i2} (\tilde{W}_{i2})^T \tilde{W}_{i2} + d_0 \end{aligned} \tag{47}$$

where $d_0 = \frac{1}{2} \sum_{i=1}^n (\eta_{i1} (W_{i1})^T W_{i1} + \eta_{i2} (W_{i2})^T W_{i2} + 2D_i)$.

σ is chosen so that

$$\sigma = \min_{i=1, \dots, n} \{2\rho_{i1} / \lambda_{\max}(\mathcal{H}^{-1}), 2\rho_{i2}, \gamma_{i1}\eta_{i1}, \gamma_{i2}\eta_{i2}\} \tag{48}$$

where $\lambda_{\max}(\mathcal{H}^{-1})$ represents the maximum characteristic value of \mathcal{H}^{-1} .

Thus, (47) can be rewritten as:

$$\dot{V} \leq -\sigma V + \sum_{i=1}^n (a_{i1} N(\kappa_{i1}) \dot{\kappa}_{i1} + \dot{\kappa}_{i1}) + \sum_{i=1}^n (\omega a_{i2} N(\kappa_{i2}) \dot{\kappa}_{i2} + \dot{\kappa}_{i2}) + d_0 \tag{49}$$

Both sides of (50) are multiplied by $e^{\sigma t}$, and integrated over the interval $[0, t]$; then, we have:

$$V \leq e^{-\sigma t} \sum_{i=1}^n \int_0^t e^{\sigma \tau} (a_{i1} N(\kappa_{i1}) \dot{\kappa}_{i1} + \dot{\kappa}_{i1}) d\tau + e^{-\sigma t} \sum_{i=1}^n \int_0^t e^{\sigma \tau} (\omega a_{i2} N(\kappa_{i2}) \dot{\kappa}_{i2} + \dot{\kappa}_{i2}) d\tau + \left(V(0) - \frac{d_0}{\sigma} \right) e^{-\sigma t} + \frac{d_0}{\sigma} \tag{50}$$

According to the previous analysis, the unknown constants a_{i1} and ωa_{i2} are bounded. Considering Lemma 2, it becomes evident that $V(t)$, $\int_0^t e^{\sigma \tau} (a_{i1} N(\kappa_{i1}) \dot{\kappa}_{i1} + \dot{\kappa}_{i1}) d\tau$, $\int_0^t e^{\sigma \tau} (\omega a_{i2} N(\kappa_{i2}) \dot{\kappa}_{i2} + \dot{\kappa}_{i2}) d\tau$, and κ_{i1} and κ_{i2} for $i=1, \dots, n$, are bounded over the interval $[0, t]$. Let

$$C = \max_{i=1, \dots, n} \left(\sum_{i=1}^n \int_0^t e^{\sigma \tau} (a_{i1} N(\kappa_{i1}) \dot{\kappa}_{i1} + \dot{\kappa}_{i1}) d\tau + \sum_{i=1}^n \int_0^t e^{\sigma \tau} (\omega a_{i2} N(\kappa_{i2}) \dot{\kappa}_{i2} + \dot{\kappa}_{i2}) d\tau \right) \tag{51}$$

Then, (50) can be simplified as:

$$V \leq \left(V(0) + C - \frac{d_0}{\sigma} \right) e^{-\sigma t} + \frac{d_0}{\sigma} \tag{52}$$

Considering (14), (15), and (43), for $t \rightarrow \infty$, we obtain:

$$\lim_{t \rightarrow \infty} |\delta_{xi}| = \lim_{t \rightarrow \infty} |x_i - x_d| \leq \sqrt{\frac{2d_0}{\lambda_{\min}(\mathcal{H})\sigma}} \tag{53}$$

where $\lambda_{\min}(\mathcal{H})$ represents the minimum eigenvalue of \mathcal{H} .

By observing (48) and adjusting ρ_{i1} , ρ_{i2} , γ_{i1} , γ_{i2} , η_{i1} , and η_{i2} , σ is sufficiently large. According to (53), the increase in σ or decrease in d_0 ensures that the tracking error $|\delta_{xi}|$ converges to a small neighborhood of zero. The proof is completed. \square

The control block diagram of the system is given in Figure 1.

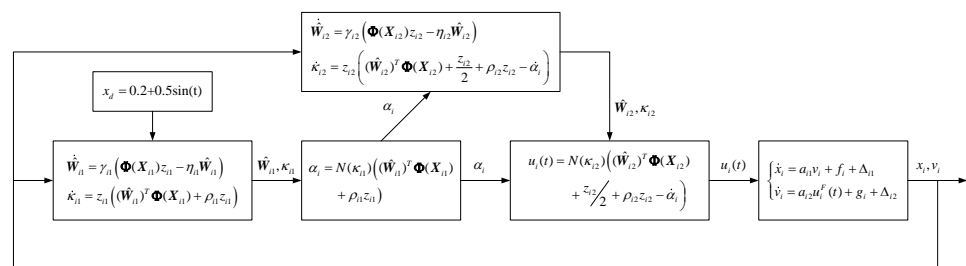


Figure 1. Block diagram of the control system.

Note 1. Noting (53), the tracking error $\lim_{t \rightarrow \infty} |\delta_{xi}| \leq \sqrt{2d_0/\lambda_{\min}(\mathcal{H})}\sigma$ can converge to a small neighborhood of zero by adjusting the parameters of d_0 and σ . We can decrease d_0 by decreasing η_{i1} and η_{i2} or increase σ by increasing ρ_{i1} , ρ_{i2} , γ_{i1} , and γ_{i2} . Nevertheless, the change in these parameters may cause the amplitude of the control signal to become larger. Therefore, to select the appropriate design parameters, a reasonable trade-off should be made between tracking performance and control signal amplitude.

4. Simulation Analysis

In this section, two simulation cases are provided to verify the effectiveness of theoretical analysis results.

Case 1. Consider the following uncertain nonlinear multi-agent system with unknown control directions and a dead-zone fault:

$$\begin{cases} \dot{x}_i(t) = 1.5v_i(t) + x_i(t)e^{-1.5x_i(t)} + 0.01\sin(t) \\ \dot{v}_i(t) = 2u^F(t) + 0.3\sin(x_i(t))\cos(v_i(t)) + 0.01\cos(t) \\ i = 1, 2, 3, 4 \end{cases} \tag{54}$$

The communication topology of the multi-agent systems is shown in Figure 2. Here, L_0 represents the leader agent, and $F_1, F_2, F_3,$ and F_4 represent the follower agents. The reference trajectory is given as $x_d = 0.2 + 0.5\sin(t)$.

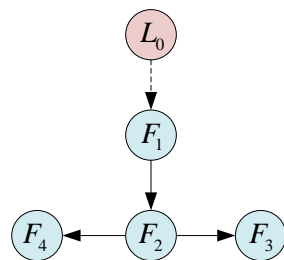


Figure 2. Communication topology of multi-agent systems.

According to Figure 1, we obtain the Laplacian matrix \mathcal{L} and adjacency matrix \mathcal{B} as:

$$\mathcal{L} = \begin{bmatrix} 0 & 0 & 0 & 0 \\ -1 & 1 & 0 & 0 \\ 0 & -1 & 1 & 0 \\ 0 & -1 & 0 & 1 \end{bmatrix}, \quad \mathcal{B} = \text{diag}\{1, 0, 0, 0\}$$

The initial conditions of four follower agents are set as $x_1(0) = 0.25(\text{m})$, $x_2(0) = 0.15(\text{m})$, $x_3(0) = 0.1(\text{m})$, $x_4(0) = 0.35(\text{m})$, and $v_1(0) = v_2(0) = v_3(0) = v_4(0) = 0(\text{m/s})$. The model of the actuator dead-zone fault is shown in (2), and the parameters are given as $\omega = 1.5$, $b_r = 0.3$, and $b_{il} = 0.1$.

The fuzzy logic systems are introduced to approximate the uncertain dynamics $F_{i1} = a_{i1}^2 z_{i1}/2 + f_i(x_i, v_i)$ and $F_{i2} = a_{i2}^2 z_{i2}/2 + g_i(x_i, v_i)$, where $i = 1, 2, 3, 4$. The selected membership functions as:

$$\begin{aligned} \mu_{F_{i1}^1} &= \exp(-0.5(x_i - 2.5)^2), \quad \mu_{F_{i2}^2} = \exp(-0.5(x_i - 1.5)^2), \quad \mu_{F_{i3}^3} = \exp(-0.5(x_i - 0.5)^2) \\ \mu_{F_{i4}^4} &= \exp(-0.5(x_i + 0.5)^2), \quad \mu_{F_{i5}^5} = \exp(-0.5(x_i + 1.5)^2), \quad \mu_{F_{i6}^6} = \exp(-0.5(x_i + 2.5)^2) \end{aligned}$$

In this paper, the Nussbaum gain function is selected as $N(\kappa) = \kappa^2 \cos(\kappa)$; the initial conditions of the adaptive control laws are set as $\kappa_{i1}(0) = \kappa_{i2}(0) = 0.01$ and $\hat{W}_{i1}(0) = \hat{W}_{i2}(0) = 0.01$; the other parameters are given as $\mathcal{G} = 0.01$, $\rho_{i1} = 50$, $\rho_{i2} = 10$, $\gamma_{i1} = \gamma_{i2} = 10$, $\eta_{i1} = \eta_{i2} = 20$, $\varepsilon_{i1}^* = \varepsilon_{i2}^* = 0.1$, and $i = 1, 2, 3, 4$; and the simulation time is $t = 30$ (s). The simulation results were obtained using the designed control law and are shown in Figures 3–9.

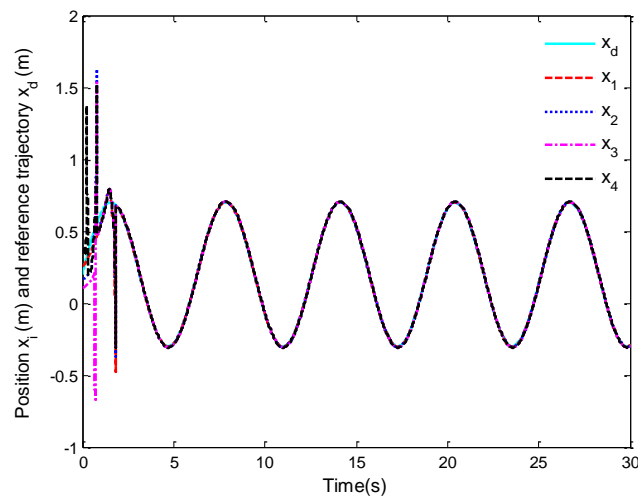


Figure 3. The curves of the agents' output x_i and the reference trajectory x_d .

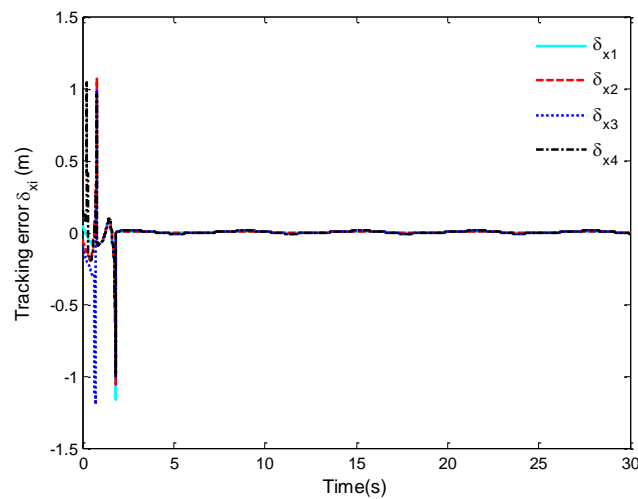


Figure 4. The curves of the position tracking error δ_{xi} .

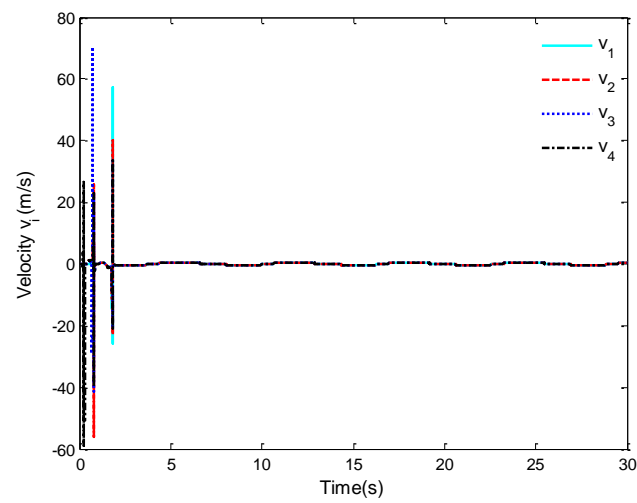


Figure 5. The curves of the agents' velocity v_i .

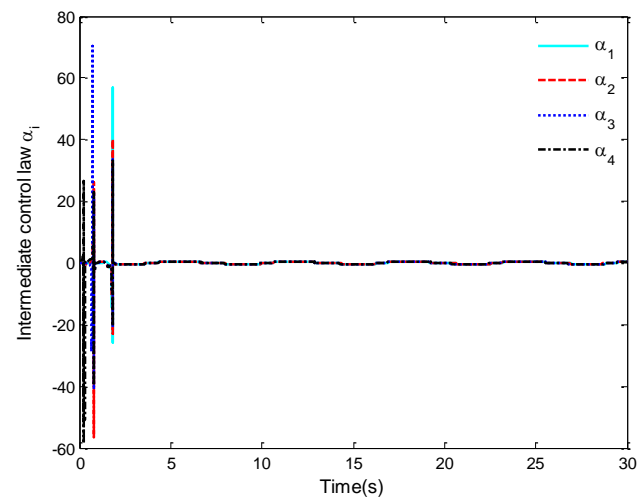


Figure 6. The curves of the intermediate control law α_i .

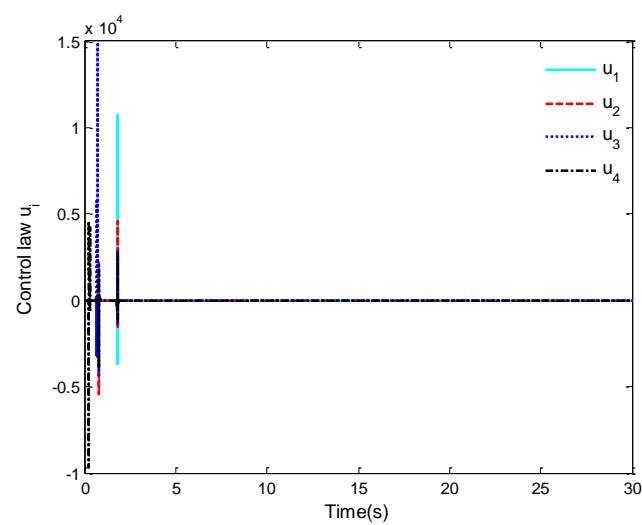


Figure 7. The curves of the control law u_i .

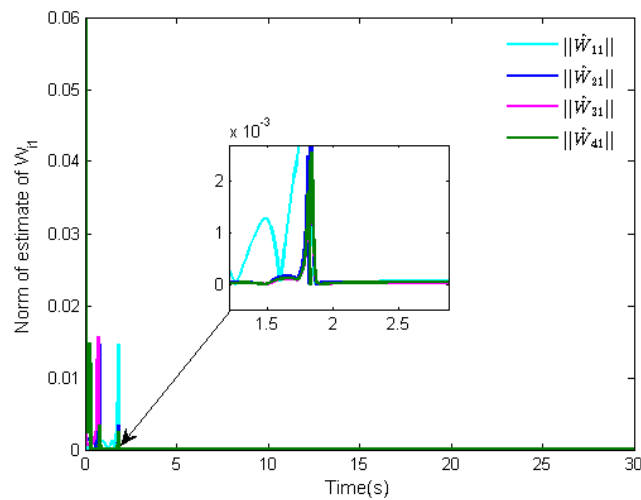


Figure 8. The curves of the norm for the adaptive control law \hat{W}_{i1} .

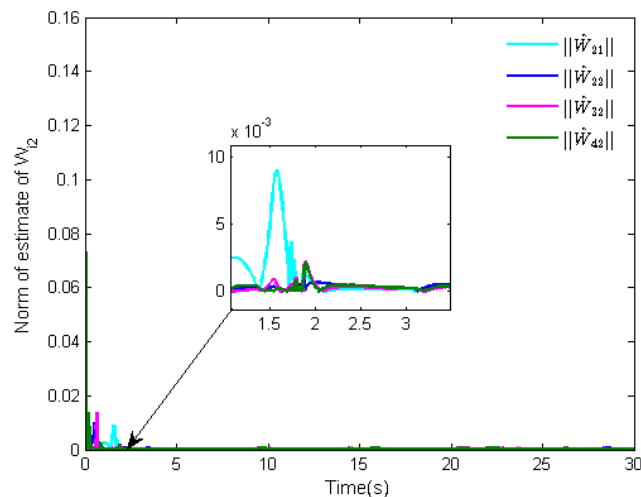


Figure 9. The curves of the norm for the adaptive control law \hat{W}_{i2} .

The curves of the position output x_i of four follower agents and the reference trajectory x_d of the leader agent are shown in Figure 3. It can be seen from Figure 3 that the position output of the four following agents can track the trajectory of the leader agent under the action of the designed control law. Notably, there is a large error within $t = 2(s)$ at the beginning of the simulation due to the existence of a dead-zone fault, but with continuous simulation, good tracking performance can finally be obtained. Figure 4 shows the curves of tracking error δ_{xi} of four following agents, and these tracking errors are seen to converge to a small neighborhood of zero in a short time, which also proves the effectiveness of the theoretical results from another angle. Figure 5 gives the curves of the velocity v_i of four follower agents. Moreover, the curves of the intermediate control laws α_i are given in Figure 6, and Figure 7 gives the curves of the adaptive fuzzy control law u_i . The curves of the norm for adaptive laws \hat{W}_{i1} and \hat{W}_{i2} are displayed as Figures 8 and 9, respectively.

Case 2. In this case, to illustrate the effectiveness of the proposed control law, a class of practical uncertain nonlinear multi-agent systems is considered [41]. Each follower agent is a model of ship steering with an unknown control direction, where we assume that

each follower agent is affected by the dead-zone fault. The multi-agent systems dynamic model is given as:

$$\begin{cases} \dot{x}_{i1} = x_{i2} + f_{i1}(x_{i1}) \\ \dot{x}_{i2} = B_i u_i^F + f_{i2}(x_{i2}) \\ i = 1, 2, 3, 4 \end{cases} \quad (55)$$

where we assume that the model of each ship steering is identical for simplicity. The uncertain, unknown nonlinear dynamics in the models of ship steering are $f_{i1}(x_{i1}) = 0$, $f_{i2}(x_{i2}) = -x_{i2}/21 - 0.3x_{i2}^3/21$, and $B_i = 0.23/21$. The communication topology is shown in Figure 1. The model of a dead-zone fault is described as (2).

In the practical simulation, the initial conditions are given as $x_1(0) = 0.3(\text{m})$, $x_2(0) = 0.2(\text{m})$, $x_3(0) = 0.1(\text{m})$, $x_4(0) = 0.05(\text{m})$, and $v_1(0) = v_2(0) = v_3(0) = v_4(0) = 0(\text{m/s})$. The fuzzy logic systems are introduced to approximate the uncertain dynamics $F_{i1} = a_{i1}^2 z_{i1}/2$ and $F_{i2} = a_{i2}^2 z_{i2}/2 + f_{i2}(x_{i2})$, $i = 1, 2, 3, 4$, where $a_{i1} = 1.0$ and $a_{i2} = 0.23/21$. The membership functions and other design parameters are selected as in Case 1. Applying the adaptive fuzzy control law designed in this paper yields the simulation results that are shown in Figures 10–16.

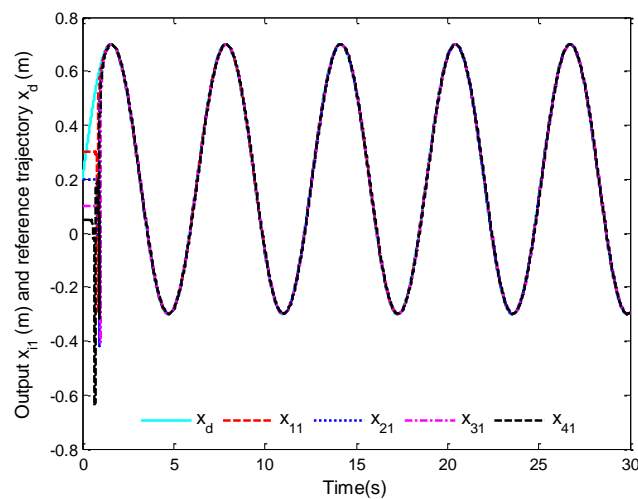


Figure 10. The curves of output x_{i1} and the reference trajectory x_d .

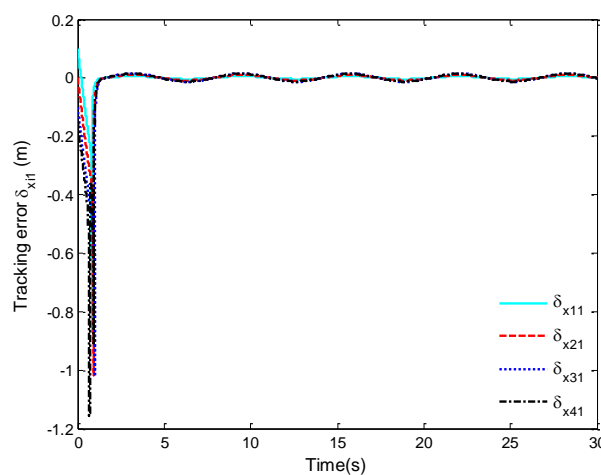


Figure 11. The curves of tracking error $\delta_{x_{it}}$.

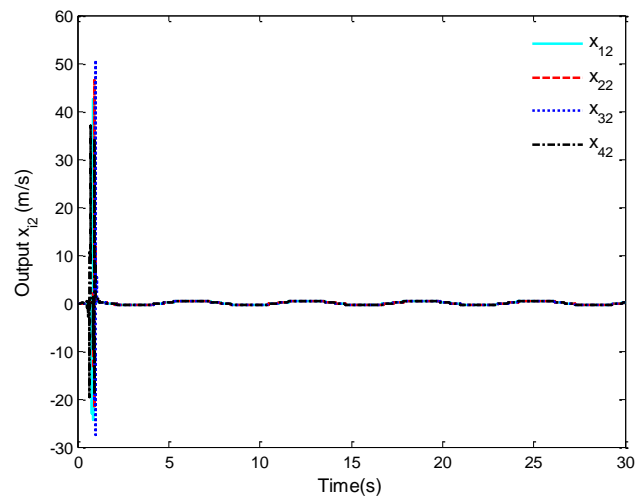


Figure 12. The curves of output x_{i2} .

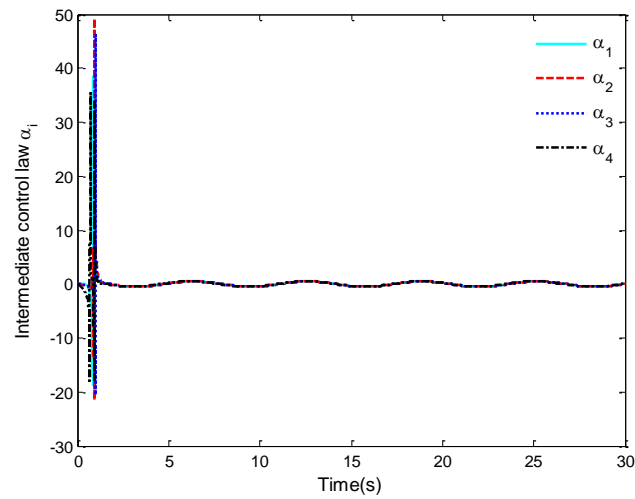


Figure 13. The curves of the intermediate control law α_i .

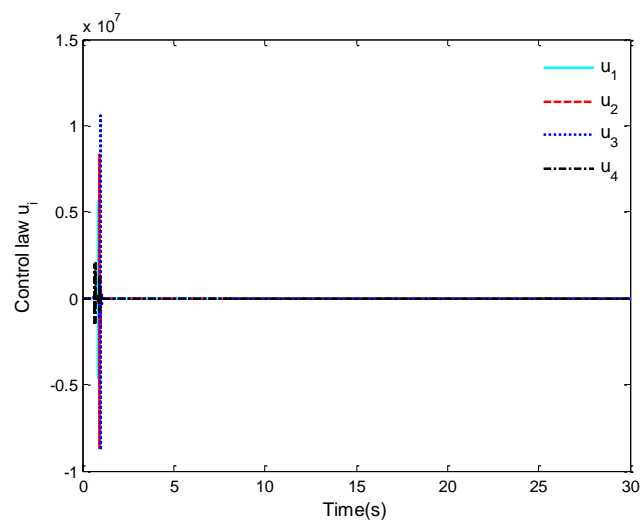


Figure 14. The curves of the control law u_i .

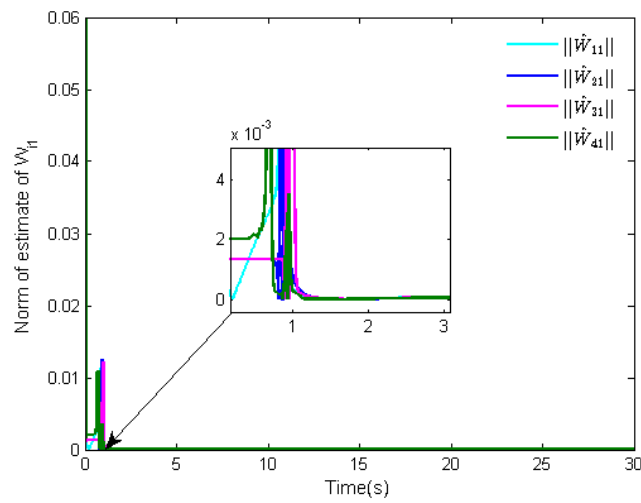


Figure 15. The curves of the norm for the adaptive control law \hat{W}_{11} .

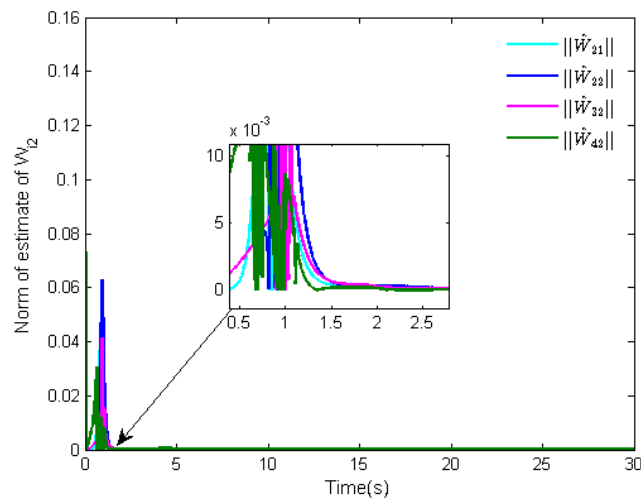


Figure 16. The curves of the norm for the adaptive control law \hat{W}_{12} .

The curves of the output x_{i1} of four ship steering systems and reference trajectory x_d are shown in Figure 10, and the tracking error curves are given in Figure 11. It is evident from the two figures that our presented adaptive fuzzy control law can solve the tracking problem of ship steering systems. Additionally, the tracking error of each ship steering system can converge to a small neighborhood of zero. The curves of output x_{i2} are shown in Figure 12. Furthermore, the curves for the intermediate control law α_i and final control law u_i are given in Figures 13 and 14, and Figures 15 and 16 display the curves of the norm for the adaptive control laws \hat{W}_{11} and \hat{W}_{12} .

To further clarify the effectiveness of the control law (Scheme 1) proposed in this paper, the results for the control law (Scheme 2) in [41] are displayed in Figures 17 and 18 for comparison. The parameters of Scheme 1 are the same as those in Case 2. For Scheme 2, the parameters are set as: $c_{11} = 25$, $c_{12} = 15$, $c_{21} = 12$, $c_{22} = 8.0$, $c_{31} = 11$, $c_{32} = 5.5$, $c_{41} = 10$, $c_{42} = 7.5$, $l_{12} = 10$, $l_{22} = 7.5$, $l_{32} = 6.5$, $l_{42} = 8.5$, $\kappa_1 = \kappa_2 = \kappa_3 = \kappa_4 = 30$, $r_{11} = r_{12} = 1.5$, $r_{21} = r_{22} = 3.5$, $r_{31} = r_{32} = 2.0$, $r_{41} = r_{42} = 4.0$, $\gamma_1 = \gamma_2 = \gamma_3 = \gamma_4 = 1.8$, and $\sigma_1 = \sigma_2 = \sigma_3 = \sigma_4 = 0.5$. The model of the actuator dead-zone fault is shown in (2), and the parameters are given as $\omega = 1.5$, $b_{ir} = 0.3$, and $b_{il} = 0.1$. The initial states are consistent

with Case 2. The comparison results of output x_{i1} are shown in Figure 17, and those of output x_{i2} are shown in Figure 18.

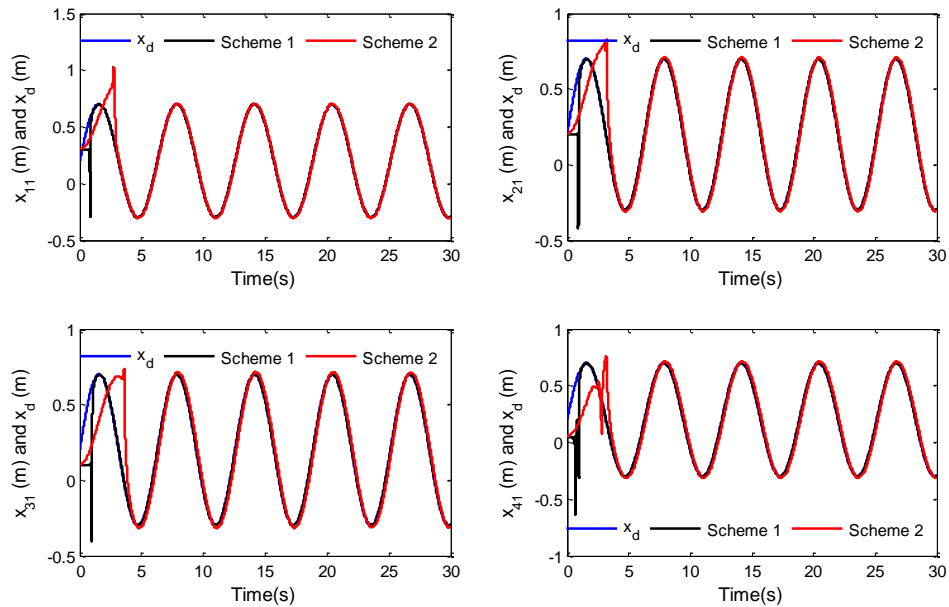


Figure 17. The comparison results of output x_{i1} .

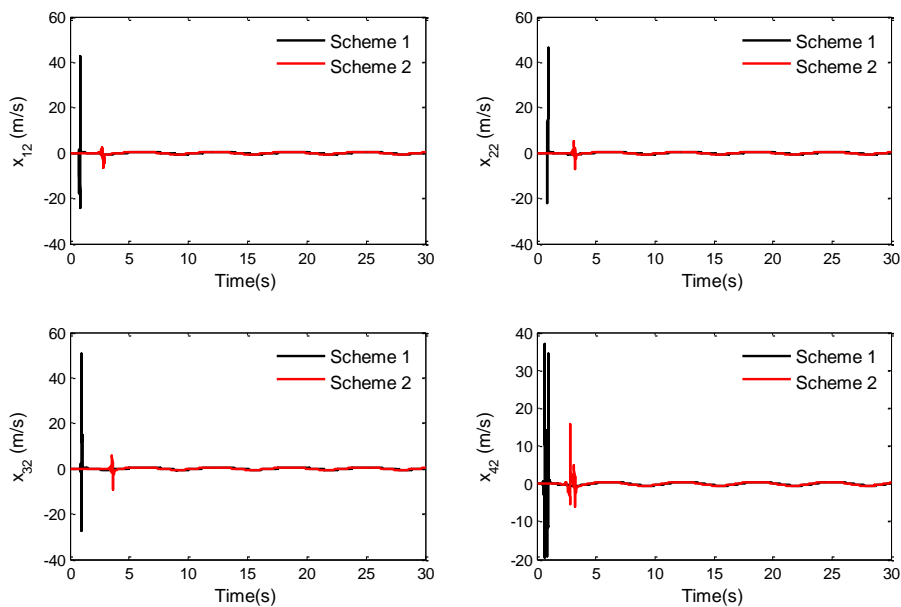


Figure 18. The comparison results of output x_{i2} .

It is observed from Figures 17 and 18 that the tracking control problem of the given system can be achieved using both Scheme 1 and Scheme 2. Although there is a certain overshoot, the system can obtain better performance under the action of Scheme 1. Furthermore, based on the application of the proposed adaptive control law (Scheme 1), the given multi-agent systems can achieve convergence in a relatively short time, and the tracking errors can converge to a small neighborhood of zero.

5. Conclusions

In this paper, the problem of consensus tracking control for a class of uncertain nonlinear multi-agent systems with unknown control directions and an actuator dead-zone fault is discussed. By introducing the fuzzy logic system, the unknown uncertain nonlinear dynamics in the analysis process are approximated. The application of the Nussbaum gain function technology solves the problem of control law design in the presence of an unknown control direction and a dead-zone fault. Finally, an adaptive fuzzy tracking control law is proposed. The simulation results show that the given uncertain nonlinear multi-agent systems can accurately track the trajectory of the leader agent and that the tracking errors finally converge to a small neighborhood of zero.

The adaptive fuzzy control law proposed in this paper has good control; therefore, it can solve the control problem of uncertain nonlinear multi-agent systems with unknown control directions and a dead-zone fault, and the tracking error can converge to a small neighborhood of zero by selecting the appropriate design parameters. However, the main limitation of this paper is that it does not consider more general cases, such as time delay, time-varying control gain, packet dropout and input constraint. Therefore, the authors will focus on these cases in future research.

Author Contributions: Data curation, X.Z.; Formal analysis, X.D.; Funding acquisition, X.D. and X.Z.; Investigation, X.D. and X.Z.; Methodology, X.D.; Software, X.Z.; Validation, X.D.; Visualization, X.Z.; Writing—original draft, X.D.; Writing—review and editing, X.D. and X.Z. All authors have read and agreed to the published version of the manuscript.

Funding: This work was partially supported by the Natural Science Research of Colleges and Universities of Anhui Province under grant KJ2020A0344, the Basic Ability Improvement Project for Young and Middle-Aged Teachers of Guangxi Province under grant 2021ky0857, and the Program for the Top Talents of Anhui Polytechnic University.

Institutional Review Board Statement: Not applicable.

Informed Consent Statement: Not applicable.

Data Availability Statement: Not applicable.

Conflicts of Interest: The authors declare no conflict of interest.

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