

# Joint Relay and Jammer Selection Improves the Physical Layer Security in the Face of CSI Feedback Delays

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**Abstract**—We enhance the physical layer security (PLS) of amplify-and-forward (AF) relaying networks with the aid of joint relay and jammer selection (JRJS), despite the deleterious effect of channel state information (CSI) feedback delays. Furthermore, we conceive a new outage-based characterization approach for the JRJS scheme. The traditional best relay selection (TBRS) is also considered as a benchmark. We first derive closed-form expressions of both the connection outage probability (COP) and the secrecy outage probability (SOP) for both the TBRS and JRJS schemes. Then, a reliable and secure connection probability (RSCP) is defined and analyzed for characterizing the effect of the correlation between the COP and the SOP introduced by the corporate source-relay link. The reliability-security ratio (RSR) is introduced for characterizing the relationship between the reliability and the security through asymptotic analysis. Moreover, the concept of effective secrecy throughput is defined as the product of the secrecy rate and of the RSCP for the sake of characterizing the overall efficiency of the system, as determined by the transmit SNR, the secrecy codeword rate, and the power sharing ratio between the relay and the jammer. The impact of the direct source-eavesdropper link and additional performance comparisons with respect to other related selection schemes are also included. Our numerical results show that the JRJS scheme outperforms the TBRS method both in terms of the RSCP and in terms of its effective secrecy throughput, but it is more sensitive to the feedback delays. Increasing the transmit signal-to-noise ratio (SNR) will not always improve the overall throughput. Moreover, the RSR results demonstrate that, upon reducing the CSI feedback delays, the reliability improves more substantially than the security degrades, implying an overall improvement in terms of the security-reliability tradeoff. Additionally, the secrecy throughput loss due to the second-hop feedback delay is more pronounced than that due to the first-hop one.

**Index Terms**—Effective secrecy throughput, feedback delay, physical layer security (PLS), relay and jammer selection, reliability and security.

## I. INTRODUCTION

WIRELESS communications systems are particularly vulnerable to security attacks because of the inherent openness of the transmission medium. Traditionally, the information privacy of wireless networks has been focused on the higher layers of the protocol stack employing cryptographically secure schemes. However, these methods typically assume a limited computing power for the eavesdroppers and exhibit inherent vulnerabilities in terms of the inevitable secret key distribution and management [1]. In recent years, physical layer security (PLS) has emerged as a promising technique of improving the confidentiality wireless communications, which exploits the time-varying properties of fading channels, instead of relying on conventional cryptosystems. The pivotal idea of PLS solutions is to exploit the dynamically fluctuating random nature of radio channels for maximizing the uncertainty concerning the source messages at the eavesdropper [2], [3].

To achieve this target, several PLS-enhancement approaches have been proposed in the literature, including secrecy-enhancing channel coding [4], secure on-off transmission designs [5], secrecy-improving beamforming (BF)/precoding, and artificial-noise-aided techniques relying on multiple antennas [6], as well as secure relay-assisted transmission techniques [7]. Specifically, apart from improving the reliability and coverage of wireless transmissions, user cooperation also has a great potential in terms of enhancing the wireless security against eavesdropping attacks. There has been a growing interest in improving the security of cooperative networks at the physical layer [8]–[14]. To explore the spatial diversity potential of the relaying networks and to boost the secrecy capacity (the difference between the channel capacity of the legitimate main link and that of the eavesdropping link), most of the existing work has been focused on secrecy-enhancing BF [8], [9], as well as on intelligent relay node/jammer node (RN/JN) selection, etc. Notably, given the availability of multiple relays, appropriately designed RN/JN selection is capable of achieving a significant security improvement for cooperative networks, which is emerging as a promising research topic. In particular, Zou *et al.* investigated both amplify-and-forward (AF)- and decode-and-forward (DF)-based optimal relay selection conceived for

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83 enhancing the PLS in cooperative wireless networks [10], [11],  
 84 where the global channel state information (CSI) of both the  
 85 main link and the eavesdropping link was assumed to be avail-  
 86 able. Similarly, jamming techniques, which impose artificial  
 87 interference on the eavesdropper, have also attracted substantial  
 88 attention [12]–[14]. More specifically, several sophisticated  
 89 joint relay and jammer selection (JRJS) schemes were proposed  
 90 in [12], where the beneficially selected relay increases the reli-  
 91 ability of the main link, whereas the carefully selected jammer  
 92 imposes interference on the eavesdropper and simultaneously  
 93 protects the legitimate destination from interference. In [13]  
 94 and [14], cooperative jamming has been studied in the context  
 95 of bidirectional scenarios, and efficient RN/JN selection criteria  
 96 have been developed for achieving improved secrecy rates with  
 97 the aid of multiple relays. Furthermore, more effective relaying  
 98 and jamming schemes, when taking the information leakage  
 99 of the source–eavesdropper link into consideration, have been  
 100 presented lately in [15] and [16].

101 Nevertheless, an idealized assumption of the previously re-  
 102 ported research on PLS is the availability of perfect channel  
 103 state information (CSI), which is regarded as a stumbling block  
 104 in the way of invoking practical secrecy-enhancing Wyner  
 105 coding, on–off design, BF/precoding, and RN/JN selection.  
 106 However, this idealized simplifying assumption is not realistic,  
 107 since practical channel estimation imposes CSI imperfections,  
 108 which are aggravated by the feedback delay, limited-rate feed-  
 109 back, and channel estimation errors (CEEs) [17]. Generally, the  
 110 related research has been focused on the issues of robust secure  
 111 BF design from an average secrecy-rate-based optimization  
 112 perspective for point-to-point multi-antenna aided channels and  
 113 relay channels [18], [19] supporting delay-tolerant systems.  
 114 For systems imposing stringent delay constraints, particularly  
 115 in imperfect CSI scenarios, perfect secrecy cannot always be  
 116 achieved. Hence, the secrecy-outage-based characterization of  
 117 systems is more appropriate, which provides a probabilistic  
 118 performance measure of secure communication. The concept  
 119 of secrecy outage was adopted in [20] for characterizing the  
 120 probability of having both reliable and secure transmission,  
 121 which, however, is inapplicable for the imperfect CSI case and  
 122 fails to distinguish a connection outage from the secrecy outage.  
 123 In [21], an alternative secrecy outage formulation is proposed  
 124 for characterizing the attainable security level and provided  
 125 a general framework for designing transmission schemes that  
 126 meet specific target security requirements. To quantify both the  
 127 reliability and security performance at both the legitimate and  
 128 eavesdropper nodes separately, two types of outages, namely,  
 129 the connection outage probability (COP) and the secrecy outage  
 130 probability (SOP) are introduced. Then, considering the impact  
 131 of time delay caused by the antenna selection process at the  
 132 legitimate receiver, Hu *et al.* [22] proposed a new secure  
 133 transmission scheme in the multiinput multioutput multieaves-  
 134 dropper wiretap channel. Much recently, considering the out-  
 135 dated CSI from the legitimate receiver, a new secure on–off  
 136 transmission scheme was proposed for enhancing the secrecy  
 137 throughput in [23].

138 Moreover, prior studies of the outage-based secure trans-  
 139 mission design are limited to single-antenna-assisted single-  
 140 hop systems and have not been considered for cooperative

relaying systems. Hence, the issues of secure transmissions 141  
 over cooperative relaying channels expressed in terms of the 142  
 SOP, COP, and secrecy throughput constitute an open problem. 143  
 On the other hand, apart from CEE, the CSI feedback delay 144  
 results in critical challenges for the PLS of cooperative relaying 145  
 systems, particularly when considering the specifics of RN/JN 146  
 selection. In [15], the effects of outdated CSI knowledge con- 147  
 cerning the legitimate links on the ergodic secrecy rate achieved 148  
 by the proposed secure transmission strategy in the context 149  
 of DF relaying is investigated. The impact of CSI feedback 150  
 delay on the secure relay and jammer selection conceived for 151  
 DF relaying was investigated in [24], albeit only in terms 152  
 of the SOP. In our previous study [25], we considered the 153  
 secure transmission design and the secrecy performance of an 154  
 opportunistic DF system relying on outdated CSI, where only a 155  
 single relay is invoked. Additionally, during the revision of this 156  
 work, we investigated the security performance for outdated AF 157  
 relay selection in [26]. Therefore, in this treatise, we extend 158  
 our investigations to the PLS of multiple AF relaying assisted 159  
 networks relying on RN/JN selection. 160

Explicitly, we focus our attention on the outage-based char- 161  
 acterization of secure transmissions in cooperative relay-aided 162  
 networks relying on realistic CSI feedback delay. To exploit the 163  
 multirelay induced diversity gain and the associated jamming 164  
 capabilities, joint AF relay node and jammer node selection 165  
 is employed by the relay–destination link. We assume that, in 166  
 line with the practical reality, the instantaneous eavesdropper’s 167  
 CSI is unavailable at the legitimate transmitter and that the 168  
 RN/JN selections are performed based on the outdated CSI of 169  
 the main links. Two types of cooperative strategies are invoked 170  
 by our cooperative network operating under secrecy constraints, 171  
 namely, the traditional best relay selection (TBRS) strategy and 172  
 the JRJS strategy. Specifically, the main contributions of this 173  
 paper can be summarized as follows. 174

- We develop an outage-based characterization for quan- 175  
 tifying both the reliability and security performance of 176  
 a two-hop AF relaying system. Specifically, in contrast 177  
 to [21] and [22], we propose the novel definition of 178  
 the reliable and secure connection probability (RSCP). 179  
 Explicitly, closed-form expressions of the COP, the SOP, 180  
 and the RSCP are derived for both the TBRS and for our 181  
 JRJS strategies. Numerical results demonstrate that the 182  
 JRJS scheme outperforms the TBRS scheme in terms of 183  
 its RSCP. 184
- We also introduce the reliability–security ratio (RSR) 185  
 for characterizing their direct relationship by a single 186  
 parameter through the asymptotic analysis of the COP and 187  
 the SOP in the high-SNR regime. We derive the RSR for 188  
 both the TBRS and JRJS strategies for investigating the 189  
 effect of secrecy codeword rate setting, as well as that 190  
 of the feedback delay and that of the power sharing ratio 191  
 between the relay and the jammer on the RSR. 192
- We then modify the definition of effective secrecy 193  
 throughput by multiplying the secrecy rate with the RSCP, 194  
 which results in an optimization problem of the trans- 195  
 mit signal-to-noise ratio (SNR), secrecy codeword rate, 196  
 and power sharing between the relay and the jammer. 197  
 198

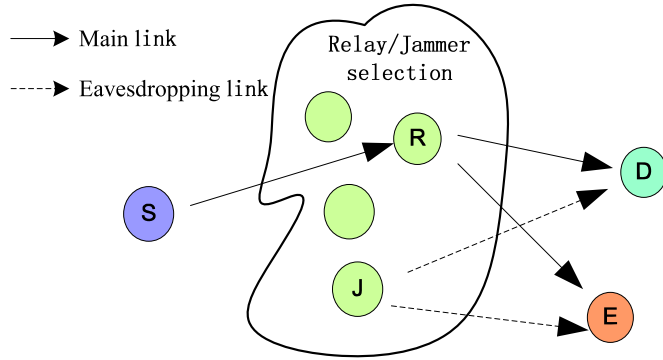


Fig. 1. Cooperative relaying network assisted by multiple relays in the presence of an eavesdropper.

199 It is shown that, compared with the TBRS strategy,  
 200 JRJS achieves a significantly higher effective secrecy  
 201 throughput, and the corresponding throughput loss is  
 202 more sensitive to feedback delays. The impact of the direct  
 203 source–eavesdropper link and additional throughput  
 204 performance comparisons with respect to other related  
 205 selection schemes are further discussed.

206 The remainder of this paper is organized as follows.  
 207 Section II introduces our system model and describes both  
 208 the TBRS and our JRJS strategies. In Sections III and IV,  
 209 we present the mathematical framework of our performance  
 210 analysis both for the TBRS strategy and for the JRJS strategy,  
 211 respectively, including the COP, the SOP, the RSCP, the RSR,  
 212 and the effective secrecy throughput. Our numerical results  
 213 and discussions are provided in Section V. Finally, Section VI  
 214 presents our concluding remarks.

## 215 II. SYSTEM MODEL

### 216 A. System Description

217 Consider a cooperative relaying network consisting of a  
 218 source  $S$ , a destination  $D$ ,  $K_r$  relays  $R_k$ ,  $k = 1, \dots, K_r$ , and  
 219 an eavesdropper  $E$ , as shown in Fig. 1, where all nodes are  
 220 equipped with a single transmit antenna (TA), except for the  
 221 source, which has  $N_t$  TAs. The cooperative relay architecture  
 222 in Fig. 1 is generally applicable to diverse practical wireless  
 223 systems in the presence of an eavesdropper, including the  
 224 family of wireless sensor networks (WSNs), mobile ad hoc  
 225 networks (MANETs), and the long-term evolution advanced  
 226 cellular systems [11].

227 To exploit the diversity potential of multiple relay nodes over  
 228 independently fading channels, AF relay/jammer selection is  
 229 employed. All relays operate in the half-duplex AF mode, and  
 230 data transmission is performed in two phases. More particu-  
 231 larly, during the broadcast phase, the source node transmits its  
 232 signal to a selected relay with the aid of BF, which is invoked  
 233 for forwarding the signal received from  $S$  to  $D$ . An inherent  
 234 assumption is that the transmit BF weights are based on the  
 235 CSI estimates quantified and fed back by the selected relay.  
 236 During the cooperative phase, a pair of appropriately selected  
 237 relays transmit toward  $D$  and  $E$ , respectively. A conventional

relay (denoted by  $R^*$ ) forwards the source’s message to the  
 238 destination. Another relay (denoted by  $J^*$ ) operates in the  
 239 “jammer mode” and imposes intentional interference upon  $E$  in  
 240 order to confuse it. However,  $D$  is unable to mitigate the artificial  
 241 interference emanating from the jammer node  $J^*$  due to its  
 242 critical secrecy constraints [12]. It should be noted that both  
 243 the process of RN/JN selection and the feedback of the transmit  
 244 BF weights from  $R^*$  to  $S$  may impose a time lag between the  
 245 data transmission and the channel estimation. These time delays  
 246 are denoted by  $T_{d_{SR}}$  and  $T_{d_{RD}}$ , respectively. Furthermore, we  
 247 assume that the BF and RN/JN selection process is based  
 248 on the perfectly estimated but outdated CSI. We employ the  
 249 first-order autoregressive outdated CSI model of [20], while  
 250 relying on the correlation coefficients of  $\rho_{SR} = J_0(2\pi f_d T_{d_{SR}})$   
 251 and  $\rho_{RD} = J_0(2\pi f_d T_{d_{RD}})$  for the two hops, where  $J_0(\cdot)$  is  
 252 the zero-order Bessel function of the first kind, and  $f_d$  is the  
 253 Doppler frequency.  
 254

A slow flat block Rayleigh fading environment is assumed,  
 255 where the channel remains static for the coherence interval (one  
 256 slot) and changes independently in different coherence inter-  
 257 vals, as denoted by  $h_{i,j} \sim \mathcal{CN}(0, \sigma_{i,j}^2)$ ,  $i, j \in \{S, R, J, D, E\}$ .  
 258 The direct communication links are assumed to be unavailable  
 259 due to the presence of obstructions between  $S$  and  $D$ , as well  
 260 as the eavesdropper.<sup>1</sup> This assumption follows the rationale of  
 261 [12] and has been routinely exploited in previous literature (see  
 262 [27] and [28] and the references therein), where the source  
 263 and relays belong to the same cluster, whereas the destination  
 264 and the eavesdropper are located in another. More specifically,  
 265 this assumption is particularly valid in networks with broadcast  
 266 and unicast transmission, where each terminal is a legitimate  
 267 receiver for one signal and acts as an eavesdropper for some  
 268 other signal. Therefore, the security concerns are only related  
 269 to the cooperative relay-aided channel. Furthermore, additive  
 270 white Gaussian noise (AWGN) is assumed with zero mean  
 271 and unit variance  $N_0$ . Let  $P_i$  be the transmit power of node  
 272  $i$ , and the instantaneous SNR of the  $i \rightarrow j$  link is given by  
 273  $\gamma_{i,j} = P_i |h_{i,j}|^2 / N_0$ .  
 274

We employ the constant-rate Wyner coding scheme for con-  
 275 structing wiretap codes of [2] to meet the PLS requirements  
 276 due to the fact that the accurate global CSI is not available.  
 277 Let  $\mathbb{C}(R_0, R_s, N)$  denote the set of all possible Wyner codes  
 278 of length  $N$ , where  $R_0$  is the codeword transmission rate, and  
 279  $R_s$  is the confidential information rate ( $R_0 > R_s$ ). The positive  
 280 rate difference  $R_e = R_0 - R_s$  is the cost of providing secrecy  
 281 against the eavesdropper. A confidential message is encoded  
 282 into a codeword at  $S$  and then transmitted to  $D$ .  
 283

### 284 B. Secure Transmission

In the broadcast phase,  $S$  transmits its BF signal  $s(t)$  to the  
 285 selected relay  $R^*$ , where the relay selection is performed  
 286 before data transmission commences, and the selection cri-  
 287 terion will be detailed later in the context of the cooper-  
 288 ative phase. The transmit BF vector  $\mathbf{w}(t|T_d)$  is calculated  
 289 using the perfectly estimated but outdated CSI given by  
 290

<sup>1</sup>The case when the  $S \rightarrow E$  link is introduced will be investigated separately in Section VI.

291  $\mathbf{w}(t|T_{d_{SR}}) = \mathbf{h}_{SR^*}^H(t - T_d)/|\mathbf{h}_{SR^*}(t - T_{d_{SR}})|$  [29], where we  
 292 have  $\mathbf{h}_{SR^*}(t) = [h_{SR^*,1}(t), \dots, h_{SR^*,N_t}(t)]^T$ , and the signal  
 293 received by the relay  $R^*$  can be written as

$$y_{R^*}(t) = \sqrt{P_s} \mathbf{w}(t|T_d) \mathbf{h}_{SR^*}(t) s(t) + n_{SR^*}(t) \quad (1)$$

294 where  $n_{SR^*}(t)$  is the AWGN at the relay. Then, we can  
 295 define the received SNR at the relay node as  $\gamma_{SR} =$   
 296  $P_S |\mathbf{w}(t|T_{d_{SR}}) \mathbf{h}_{SR^*}(t)|^2 / N_0$ .

297 In the cooperative phase, we consider two RN/JN selection  
 298 schemes performed by  $D$ : relay selection without jamming  
 299 and JRJS.

300 1) *Traditional Best Relay Selection*: The first category of so-  
 301 lutions does not involve a jamming process, and therefore, only  
 302 a conventional relay accesses the channel during the second  
 303 phase of the protocol. The relay selection process is performed  
 304 based on the highest instantaneous SNR of the second hop,  
 305 which is formulated as

$$R^* = \arg \max_{R_k \in \mathcal{R}} \left\{ \frac{\tilde{\gamma}_{R_k D}}{\mathbb{E}[\gamma_{R_k E}]} = \frac{P_R |\tilde{h}_{R_k D}(t - T_d)|^2}{N_0 \mathbb{E}[\gamma_{R_k E}]} \right\} \quad (2)$$

306 where  $\tilde{\gamma}_{R_k D}$  is the instantaneous SNR in the relay selection  
 307 process, and  $\mathbb{E}[\gamma_{R_k E}]$  denotes the average SNR at  $E$ . We can  
 308 model  $\gamma_{R_k D}$  and  $\tilde{\gamma}_{R_k D}$  as two gamma distributed random  
 309 variables having the correlation factor of  $\rho_{RD}^2$ .

310 During the second phase, the received signal  $y_{R^*}(t)$   
 311 is multiplied by a time-variant AF-relay gain  $G$  and  
 312 retransmitted to  $D$ , where we have  $G =$   
 313  $\sqrt{P_R / (P_S |\mathbf{w}_{\text{opt}}(t|T_{d_{SR}}) \mathbf{h}_{SR^*}(t)|^2 + N_0)}$ . After further math-  
 314 ematical manipulations, the mutual information (MI) between  
 315  $S$  and  $D$ , as well as the eavesdropper, can be written as

$$I_D^{\text{TBRS}} = \frac{1}{2} \log(1 + \gamma_D^{\text{TBRS}}) = \frac{1}{2} \log \left( 1 + \frac{\gamma_{SR} \gamma_{R^* D}}{\gamma_{SR} + \gamma_{R^* D} + 1} \right) \quad (3)$$

$$I_E^{\text{TBRS}} = \frac{1}{2} \log(1 + \gamma_E^{\text{TBRS}}) = \frac{1}{2} \log \left( 1 + \frac{\gamma_{SR} \gamma_{R^* E}}{\gamma_{SR} + \gamma_{R^* E} + 1} \right). \quad (4)$$

316 2) *Joint Relay and Jammer Selection*: Similarly, consider-  
 317 ing the unavailability of the instantaneous CSI regarding the  
 318 eavesdropper, we adopt a suboptimal RN/JN selection metric  
 319 conditioned on the outdated CSI as

$$R^* = \arg \max_{R_k \in \mathcal{R}} \left\{ \frac{\tilde{\gamma}_{R_k D}}{\mathbb{E}[\gamma_{R_k E}]} \right\} \quad (5)$$

$$J^* = \arg \min_{R_k \in \mathcal{R} - R^*} \left\{ \frac{\tilde{\gamma}_{R_k D}}{\mathbb{E}[\gamma_{R_k E}]} \right\}$$

320 where  $J^*$  is selected for minimizing the interference imposed  
 321 on  $D$ .

322 It should be noted that, to have the same transmit power as  
 323 that of the TBRS case, we assume that  $P_{R^*} + P_{J^*} = P_R$  for  
 324 our JRJS strategy and introduce  $\lambda = P_{R^*} / (P_{R^*} + P_{J^*})$  as the

ratio of the relay's transmit power to the total power required  
 by the active relay and jammer. 326

In the cooperative phase,  $R^*$  will also amplify the received  
 signal  $y_{R^*}(t)$  by  $G$  and forward it to  $D$ . At the same time, the  
 jammer  $J^*$  will generate intentional interference to confuse  $E$ ,  
 which will also cause interference at  $D$ . Consequently, the MI  
 between the terminals is given by 331

$$I_D^{\text{JRJS}} = \frac{1}{2} \log(1 + \gamma_D^{\text{JRJS}}) = \frac{1}{2} \log \left( 1 + \frac{\gamma_{SR} \frac{\gamma_{R^* D}}{\gamma_{J^* D} + 1}}{\gamma_{SR} + \frac{\gamma_{R^* D}}{\gamma_{J^* D} + 1} + 1} \right) \quad (6)$$

$$I_E^{\text{JRJS}} = \frac{1}{2} \log(1 + \gamma_E^{\text{JRJS}}) = \frac{1}{2} \log \left( 1 + \frac{\gamma_{SR} \frac{\gamma_{RE}}{\gamma_{JE} + 1}}{\gamma_{SR} + \frac{\gamma_{RE}}{\gamma_{JE} + 1} + 1} \right). \quad (7)$$

*Remark 1*: Generally, the optimal RN/JN selection scheme  
 should take into account the global SNR knowledge set  
 $\{\gamma_{SR}, \gamma_{RD}, \gamma_{RE}\}$ . However, given the potentially excessive  
 implementational complexity overhead of the optimal selection  
 schemes and the unavailability of the global CSI, we employ  
 suboptimal selection schemes as in [12].<sup>2</sup> Furthermore, it is  
 commonly assumed that the average SNR of the eavesdropper  
 is available at the transmitter, which seems, somehow, not  
 reasonable. However, as stated in most of the literature, such as  
 [12]–[22], [24]–[28], and [30], provided that the eavesdropper  
 belongs to the network, which is also the case in our paper,  
 the related assumption might still be deemed reasonably. Addi-  
 tionally, as in [8], [11], [12], and [24], for mathematical conve-  
 nience, we assume that the relaying channels are independent  
 and identically distributed and that we have  $\mathbb{E}[\gamma_{SR_k}] = \bar{\gamma}_{SR}$ ,  
 $\mathbb{E}[\gamma_{R_k D}] = \bar{\gamma}_{RD}$ , and  $\mathbb{E}[\gamma_{R_k E}] = \bar{\gamma}_{RE}$ . The distances between  
 the relays are assumed to be much smaller than the distances  
 between relays and source/destination/eavesdropper; hence, the  
 corresponding path losses among the different relays are ap-  
 proximately the same. This assumption is reasonable both for  
 WSNs and for MANETs associated with a symmetric clustered  
 relay configuration, and it may be also satisfied as valid by  
 classic cellular systems in a statistical sense [11]. 354

### III. SECURE TRANSMISSION WITHOUT JAMMING 355

Here, we endeavor to characterize both the reliability and  
 security performance comprehensively of the TBRS scheme.  
 We first derive closed-form expressions for both the COP and  
 the SOP. Then, the RSR is introduced through the asymptotic  
 analysis of the COP and the SOP. Furthermore, we propose  
 the novel definition of the RSCP and the effective secrecy  
 throughput. 362

<sup>2</sup>To further alleviate the cooperation-related overhead, the selection criterion is based on the  $R \rightarrow D$  link, since the second hop plays a dominant role in determining the received SNR, because the first hop corresponds to a multiple-input-single-output channel with the aid of multiple antennas, and hence, it is more likely to be better than the second hop. The optimal selection based on both hops is beyond the scope of this work.

## 363 A. COP and SOP

364 When the perfect instantaneous CSI of the eavesdropper's  
 365 channel and even the legitimate users' channel is unavailable,  
 366 alternative definitions of the outage probability may be adopted  
 367 for the statistical characterization of the attainable secrecy  
 368 performance, particularly for delay-limited applications. Based  
 369 on [31, Def. 2], perfect secrecy cannot be achieved, when we  
 370 have  $R_e < I_E$ , where  $I_E$  denotes the MI between the source  
 371 and the eavesdropper. Encountering this event is termed as a  
 372 secrecy outage. Furthermore, the destination is unable to flaw-  
 373 lessly decode the received codewords when  $R_0 > I_D$ , which is  
 374 termed as a connection outage. The grade of reliability and the  
 375 grade of security maintained by a transmission scheme may be  
 376 then quantified by the COP and the SOP, respectively.

377 We continue by presenting our preliminary results versus the  
 378 point-to-point SNRs. Let us denote the cumulative distribution  
 379 function (CDF) and the probability density function (PDF) of a  
 380 random variable  $X$  by  $F_X(x)$  and  $f_X(x)$ , respectively. On one  
 381 hand, the PDF of  $\gamma_{SR}$  using [29, eq. (15)] is given by

$$f_{\gamma_{SR}}(x) = \sum_{n=0}^{N_t-1} \binom{N_t-1}{n} \frac{\rho_{SR}^{2(N_t-1-n)} (\bar{\gamma}_{SR} (1 - \rho_{SR}^2))^n}{\bar{\gamma}_{SR}^{N_t} (N_t - 1 - n)!} \times x^{N_t-1-n} e^{-\frac{x}{\bar{\gamma}_{SR}}} \quad (8)$$

382 whereas its CDF is given by

$$F_{\gamma_{SR}}(x) = 1 - \sum_{n=0}^{N_t-1} \sum_{m=0}^{N_t-1-n} \binom{N_t-1}{n} \times \frac{\rho_{SR}^{2(N_t-1-n)} (1 - \rho_{SR}^2)^n}{m! \bar{\gamma}_{SR}^m} x^m e^{-\frac{x}{\bar{\gamma}_{SR}}}. \quad (9)$$

383 On the other hand, for the instantaneous SNR of the  $R \rightarrow$   
 384  $D$  hop, according to the principles of concomitants or induced  
 385 order statistics, the CDF of  $\gamma_{R^*D}$  can be derived as in [32]

$$F_{\gamma_{R^*D}}(y) = K_r \sum_{k=0}^{K_r-1} (-1)^k \binom{K_r-1}{k} \frac{1 - e^{-\frac{-(k+1)y}{k(1-\rho_{RD}^2)+1} \bar{\gamma}_{RD}}}{k+1}. \quad (10)$$

386 Thus, the COP of the TBRS strategy is given by

$$P_{co}^{\text{TBRS}}(R_0) = \Pr [I_D^{\text{TBRS}} < R_0] = F_{\gamma_D^{\text{TBRS}}}(\gamma_{th}^D) \quad (11)$$

387 where we have  $\gamma_{th}^D = 2^{2R_0} - 1$ , and the CDF of  $\gamma_D^{\text{TBRS}}$  can be  
 388 calculated as

$$F_{\gamma_D^{\text{TBRS}}}(x) = 1 - \int_0^\infty \left[ 1 - F_{\gamma_{R^*D}}\left(\frac{xz + x(x+1)}{z}\right) \right] f_{\gamma_{SR^*}}(z+x) dz. \quad (12)$$

389 Consequently, by substituting (8) and (10) into (12) and using  
 390 [33, eq. (3.471.9)], we arrive at a closed-form expression for

$F_{\gamma_D^{\text{TBRS}}}(x)$  as

391

$$F_{\gamma_D^{\text{TBRS}}}(x) = 1 - 2 \sum_{n=0}^{N_t-1} \sum_{k=0}^{K_r-1} \sum_{m=0}^{N_t-1-n} (-1)^k K_r \binom{N_t-1}{n} \times \binom{K_r-1}{k} \binom{N_t-1-n}{m} \times \frac{\rho_{SR}^{2(N_t-1-n)} (1 - \rho_{SR}^2)^n x^{N_t-1-n-m}}{(N_t-1-n)! (k+1) \bar{\gamma}_{SR}^{N_t-n}} \times \left[ \frac{\bar{\gamma}_{SR} x(x+1)}{\omega_k \bar{\gamma}_{RD}} \right]^{\frac{m+1}{2}} \times e^{-\left(\frac{\bar{\gamma}_{SR} + \omega_k \bar{\gamma}_{RD}}{\omega_k \bar{\gamma}_{SR} \bar{\gamma}_{RD}}\right) x} K_{m+1} \left( 2 \sqrt{\frac{x(x+1)}{\omega_k \bar{\gamma}_{SR} \bar{\gamma}_{RD}}} \right) \quad (13)$$

where we have  $\omega_k = (k(1 - \rho_{RD}^2) + 1)/(k+1)$ . Then, by  
 substituting  $x = \gamma_{th}^D$  into (13), we obtain  $P_{co}^{\text{TBRS}}$ .

Furthermore, the SOP of the TBRS strategy may be expressed as

$$P_{so}^{\text{TBRS}}(R_0, R_s) = \Pr [J_E^{\text{TBRS}} > R_0 - R_s] = 1 - F_{\gamma_E^{\text{TBRS}}}(\gamma_{th}^E) \quad (14)$$

where we have  $\gamma_{th}^E = 2^{2(R_0 - R_s)} - 1$ . Similarly, we may calcu-  
 late the CDF of  $\gamma_E^{\text{TBRS}}$  in (14) as

396

$$F_{\gamma_E^{\text{TBRS}}}(x) = 1 - 2 \sum_{n=0}^{N_t-1} \sum_{m=0}^{N_t-1-n} \binom{N_t-1}{n} \binom{N_t-1-n}{m} \times \frac{\rho_{SR}^{2(N_t-1-n)} (1 - \rho_{SR}^2)^n x^{N_t-1-n-m}}{(N_t-1-n)! \bar{\gamma}_{SR}^{N_t-n}} \times \left[ \frac{\bar{\gamma}_{SR} x(x+1)}{\bar{\gamma}_{RE}} \right]^{\frac{m+1}{2}} \times e^{-\left(\frac{\bar{\gamma}_{SR} + \bar{\gamma}_{RE}}{\bar{\gamma}_{SR} \bar{\gamma}_{RE}}\right) x} K_{m+1} \left( 2 \sqrt{\frac{x(x+1)}{\bar{\gamma}_{SR} \bar{\gamma}_{RE}}} \right). \quad (15)$$

Then, by substituting  $x = \gamma_{th}^E$  into (15), we can derive  $P_{so}^{\text{TBRS}}$ .

The COP and the SOP in (11) and (14) characterize the at-  
 tainable reliability and security performance, respectively, and  
 can be regarded as the detailed requirements of accurate system  
 design. From the definition of COP and SOP, it is clear that  
 the reliability of the main link can be improved by increasing  
 the transmit SNR (or decreasing its data rate) to reduce the  
 COP, which unfortunately increases the risk of eavesdropping.  
 Thus, a tradeoff between reliability and security may be struck,  
 despite the fact that closed-form expressions cannot be obtained  
 as in [11]. Furthermore, we denote the minimal reliability and  
 security requirements by  $\nu$  and  $\delta$ , where the feasible range of  
 the reliability constraint is  $0 < \nu < 1$ . Bearing in mind that  
 the COP is a monotonously increasing function of  $R_0$ , the  
 corresponding threshold of the codeword transmission rate is  
 $R_0^{th} = \arg\{P_{co}^{\text{TBRS}}(R_0) = \nu\}$ , which leads to a lower bound of  
 the SOP, when we have  $(R_0 - R_s) \rightarrow R_0^{th}$ . Thus, the feasible  
 range of  $\delta$  is  $P_{so}^{\text{TBRS}}(R_0^{th}, 0) < \delta < 1$ . The preceding analysis  
 indicates that, given a reliability constraint  $\nu$ , the lower bound  
 of the security constraint is determined.

416

### 417 B. Reliability–Security Ratio

418 Here, we will focus our attention on the asymptotic analysis  
419 of the COP and the SOP in the high-SNR regime. Then, inspired  
420 by [25], we introduce the concept of the RSR for characterizing  
421 the direct relationship between reliability and security.

422 *Proposition 1:* Based on the asymptotic probabilities of  $P_{co}$   
423 and  $P_{so}$  at high SNRs,<sup>3</sup> the RSR is defined as

$$P_{co}(R_0) = \Lambda [1 - P_{so}(R_0, R_s)] \quad (16)$$

424 where  $\Lambda = \lim_{\eta \rightarrow \infty} P_{co}/(1 - P_{so})$ , which represents the im-  
425 provement in COP upon decreasing the SOP. More specifically,  
426 since the reduction of the SOP/COP must be followed by an  
427 improvement of COP/SOP, a lower  $\Lambda$  implies that, when the  
428 security is reduced, the reliability is improved, and *vice versa*.  
429 Thus, for the TBRS scheme studied earlier, the RSR is derived  
430 as (17), shown at the bottom of the page.

431 *Proof:* The proof is given in Appendix B.

432 *Remark 2:* It can be seen from the preceding expression  
433 that the factor  $\Lambda$  is independent of the transmit SNR, but  
434 directly depends on the channel gains, the rate pair  $(R_0, R_s)$ ,  
435 and the number of TAs and relays. For a given  $R_s$ , reducing  
436  $R_0$  to enhance the reliability may erode the security, because  
437  $(R_0 - R_s)$  is also reduced. Conversely, increasing  $R_0$  provides  
438 more redundancy for protecting the security of the information,  
439 but simultaneously, the reliability is reduced. Hence, the RSR  
440 analysis underlines an important point of view concerning how  
441 to balance the reliability versus security tradeoff by adjusting  
442  $(R_0, R_s)$ . Furthermore, as long as a CSI feedback delay exists,  
443 the RSR has an intimate relationship with  $\rho_{SR}$  and  $\rho_{RD}$ . It is  
444 clear that the value of  $\Lambda^{\text{TBRS}}$  decreases as  $\rho_{RD}$  increases, which  
445 is due to the fact that the relay selection process only improves  
446 the reliability of the legitimate user. On the other hand, since  
447 we always have the conclusion that  $\sum_{k=0}^{K_r-1} (-1)^k \binom{K_r-1}{k} (K_r/k - 1) < 1$ , when  $\sigma_{RD}^2$  and  $\sigma_{RE}^2$  are comparable,  
449  $\Lambda^{\text{TBRS}}$  will be reduced as  $\rho_{SR}$  increases. This observation  
450 implies that, although both  $P_{co}$  and  $(1 - P_{so})$  are reduced  
451 when the first-hop CSI becomes better, the improvement of

<sup>3</sup>Assume equal power allocation between  $S$  and the relay, yielding  $P_S = P_R = P$ , and define  $\eta = P/N_0$  as the transmit SNR [24].

the reliability is more substantial than the security loss, as  $\rho_{SR}$   
increases.

### C. Effective Secrecy Throughput

It should be noted that the COP and SOP metrics ignore the  
correlation between these two outage events. More specifically,  
in contrast to the point-to-point transmission case, since the  
 $S \rightarrow R$  link's SNR included in the MI expressions of (3) and  
(4), the secrecy outage and the connection outage are definitely  
not independent of each other. Therefore, it might be of limited  
benefit in evaluating the reliability or the security separately.  
We note furthermore that, although another metric referred to  
as the secrecy throughput was introduced as the product of the  
successful decoding probability and of the secrecy rate [21],  
[22], this definition ignores the fact that a reliable transmission  
may be insecure, and the SOP is not taken into consideration.  
Hence, this metric is unable to holistically characterize the  
efficiency of our scheme, while capable of achieving both re-  
liable and secure transmission. Therefore, here, we redefine the  
effective secrecy throughput as the probability of a successful  
transmission (reliable and secure) multiplied by the secrecy  
rate, namely, as  $\varsigma = R_s P_{R\&S}$ , where the RSCP is defined as

$$P_{R\&S} = \Pr\{I_D > R_0, I_E < R_0 - R_s\}. \quad (18)$$

Upon substituting the expressions of  $I_D$  and  $I_E$  in (3) and (4)  
into (18), we can rewrite  $P_{R\&S}$  for the TBRS strategy in (19),  
shown at the bottom of the page.

Finally, using the corresponding CDFs and PDFs of (8)–(10)  
from our previous analysis, we can obtain  $P_{R\&S}^{\text{TBRS}}$  in (20),  
shown at the bottom of the next page, as well as the secrecy  
throughput.

Furthermore, considering the asymptotic result for RSCP at  
high SNRs in (20) by applying the approximation  $K_v(x) \approx$   
 $(v-1)!/2(x/2)^v$  and closing the highest terms of  $\eta$  after  
invoking the McLaurin series representation for the exponential  
function, the asymptotic effective secrecy throughput can be  
approximated as

*Remark 3:* Given the definition of COP, SOP, and the secrecy  
throughput result of (21), shown at the bottom of the next page,  
it can be shown that, for a fixed  $R_s$ , if  $R_0$  is too small, although

$$\Lambda^{\text{TBRS}} = \frac{\left[ (1 - \rho_{SR}^2)^{N_t-1} + \sum_{k=0}^{K_r-1} (-1)^k \binom{K_r-1}{k} \frac{K_r \sigma_{SR}^2}{[k(1 - \rho_{RD}^2) + 1] \sigma_{RD}^2} \right] (2^{2R_0} - 1)}{\left[ N_t (1 - \rho_{SR}^2)^{N_t-1} + \sigma_{SR}^2 / \sigma_{RE}^2 \right] (2^{2(R_0 - R_s)} - 1)} \quad (17)$$

$$\begin{aligned} P_{R\&S}^{\text{TBRS}} &= \Pr \left\{ \left\{ \gamma_{SR} > \gamma_{th}^D, \gamma_{R^*D} > \frac{\gamma_{th}^D \gamma_{SR} + \gamma_{SR}}{\gamma_{SR} - \gamma_{th}^D} \right\} \cap \left[ \left\{ \gamma_{SR} > \gamma_{th}^E, \gamma_{R^*E} < \frac{\gamma_{th}^E \gamma_{SR} + \gamma_{SR}}{\gamma_{SR} - \gamma_{th}^E} \right\} \cup \left\{ \gamma_{SR} < \gamma_{th}^E \right\} \right] \right\} \\ &= \Pr \left\{ \gamma_{SR} > \gamma_{th}^D, \gamma_{R^*D} > \gamma_{th}^D + \frac{\gamma_{th}^D (\gamma_{th}^D + 1)}{\gamma_{SR} - \gamma_{th}^D}, \gamma_{R^*E} < \gamma_{th}^E + \frac{\gamma_{th}^E (\gamma_{th}^E + 1)}{\gamma_{SR} - \gamma_{th}^E} \right\} \end{aligned} \quad (19)$$

489  $P_{RS}$  may be high (i.e., close to 1), the value of  $\varsigma$  remains small.  
 490 By contrast, if  $R_0$  is too large, the value of  $P_{co}$  is close to 1,  
 491 and therefore,  $\varsigma$  will also become small. This observation is  
 492 also suitable for  $R_s$ . Thus, as pointed out in the RSR analysis,  
 493 it is elusive to improve both the reliability and the security  
 494 simultaneously, but both of them are equally crucial in terms  
 495 of the effective secrecy throughput, which depends on the rate  
 496 pair  $(R_0, R_s)$ .

497 Additionally, (21) also reveals that increasing the SNR would  
 498 drastically reduce the effective secrecy throughput. For high  
 499 transmit SNRs, a high reliability can indeed be perfectly guar-  
 500 anteed, but at the same time, the grade of the security is severely  
 501 degraded. However, the probability of a reliable and simultane-  
 502 ously secure transmission will tend toward zero. Hence, we may  
 503 conclude that there exists an optimal SNR, which achieves the  
 504 maximal secrecy throughput.

505 In conclusion, adopting the appropriate code rate pair and  
 506 transmit SNR is crucial for achieving the maximum effective  
 507 secrecy throughput, which can be formulated as

$$\begin{aligned} \max_{R_0, R_s, \eta} \quad & \varsigma(R_0, R_s) = R_s P_{R\&S}^{\text{TBRS}} \\ \text{s.t.} \quad & P_{co} \leq \nu, P_{so} \leq \delta, 0 < R_s < R_0 \end{aligned} \quad (22)$$

508 where  $\nu$  and  $\delta$  denote the system's reliability and security  
 509 requirements. Unfortunately, it is quite a challenge to find  
 510 the closed-form optimal solution to this problem due to the  
 511 complexity of the expressions. Although suboptimal solutions  
 512 can be found numerically (with the aid of gradient-based search  
 513 techniques), the secrecy throughput optimization problem and  
 514 the corresponding complexity analysis and performance com-  
 515 parisons are beyond the scope of this work.

#### 516 IV. SECURE TRANSMISSION WITH JAMMING

517 Here, we consider the extension of the aforementioned relay  
 518 selection approaches to systems additionally invoking relay-

aided jamming. JRJS is based on the outdated but perfectly  
 estimated CSI, and the details have been presented in Section II.  
 We would also like to investigate the security performance  
 from an outage-based perspective. The COP, SOP, RSCP, and  
 effective secrecy throughput will be included.

#### A. COP and SOP

It is plausible that the main differences between the JRJS and  
 TBRS schemes are determined by the instantaneous SNR of the  
 $R \rightarrow D$  hop, where, now, a jammer is included. Based on our  
 preliminary results detailed for the point-to-point SNRs in (8)  
 and (10), we now focus our attention on the statistical analysis  
 of the SNR, including  $J^*$ . As stated for the JRJS scheme in  
 Section II,  $J^*$  corresponds to the lowest  $\tilde{\gamma}_{R_k D}$  and is selected  
 from the set  $\{\mathcal{R} - R^*\}$ . Recalling that  $R^*$  is the best relay  
 of the second hop, we have  $\tilde{\gamma}_{J^* D} = \min_{R_k \in \mathcal{R} - R^*} \{\tilde{\gamma}_{R_k D}\} \triangleq$   
 $\min_{R_k \in \mathcal{R}} \{\tilde{\gamma}_{R_k D}\}$  for  $K_r > 1$ . Using the induced order statis-  
 tics, the corresponding CDF of  $\gamma_{R^* D}$  is presented in (10),  
 whereas the PDF of  $\gamma_{J^* D}$  can be formulated as

$$f_{\gamma_{J^* D}}(x) = \frac{K_r \exp\left(\frac{-K_r x}{[(K_r - 1)(1 - \rho_{RD}^2) + 1]\tilde{\gamma}_{JD}}\right)}{[(K_r - 1)(1 - \rho_{RD}^2) + 1]\tilde{\gamma}_{JD}}. \quad (23)$$

Although the relay and jammer selection processes are not  
 entirely disjoint, we may exploit the assumption that  $\gamma_{R^* D}$  and  
 $\gamma_{J^* D}$  are independent of each other, which is valid when the  
 number of relays is sufficiently high, as justified in [24]. Let us  
 define the signal-to-interference-plus-noise ratio of the second  
 hop as  $\xi_D = \gamma_{R^* D} / (\gamma_{J^* D} + 1)$ , using (10) and (23), whose  
 CDF can be formulated as

$$F_{\xi_D}(x) = 1 - K_r \sum_{k=0}^{K_r - 1} (-1)^k \binom{K_r - 1}{k} \frac{\varphi_k e^{-\frac{x}{\tilde{\gamma}_{RD} \omega_k}}}{(k + 1)(x + \varphi_k)} \quad (24)$$

where we have  $\varphi_k = \lambda K_r \omega_k / [(K_r - 1)(1 - \rho_{RD}^2) + 1](1 - \lambda)$ .

$$\begin{aligned} P_{R\&S}^{\text{TBRS}} &= \int_{\gamma_{th}^D}^{\infty} \left[ 1 - F_{\gamma_{R^* D}} \left( \gamma_{th}^D + \frac{\gamma_{th}^D (\gamma_{th}^D + 1)}{x - \gamma_{th}^D} \right) \right] F_{\gamma_{R^* E}} \left( \gamma_{th}^E + \frac{\gamma_{th}^E (\gamma_{th}^E + 1)}{x - \gamma_{th}^E} \right) f_{\gamma_{SR^*}}(x) dx \\ &\approx 2 \sum_{n=0}^{N_t - 1} \sum_{k=0}^{K_r - 1} \sum_{m=0}^{N_t - 1 - n} (-1)^k \binom{K_r - 1}{k} \binom{N_t - 1}{n} \binom{N_t - 1 - n}{m} \frac{K_r \rho_{SR}^{2(N_t - 1 - n)} (1 - \rho_{SR}^2)^n (\gamma_{th}^D)^{N_t - 1 - n - m}}{(N_t - 1 - n)! (k + 1) \tilde{\gamma}_{SR}^{N_t - n - (m + 1)/2}} \\ &\quad \times \exp \left[ - \left( \frac{\gamma_{th}^D}{\tilde{\gamma}_{SR}} + \frac{\gamma_{th}^D}{\omega_k \tilde{\gamma}_{RD}} \right) \right] \left[ \left( \frac{\gamma_{th}^D (\gamma_{th}^D + 1)}{\omega_k \tilde{\gamma}_{RD}} \right)^{\frac{m + 1}{2}} K_{m + 1} \left( 2 \sqrt{\frac{\gamma_{th}^D (\gamma_{th}^D + 1)}{\omega_k \tilde{\gamma}_{SR} \tilde{\gamma}_{RD}}} \right) \right. \\ &\quad \left. - \exp \left( \frac{-\gamma_{th}^E}{\tilde{\gamma}_{RE}} \right) \left( \frac{\gamma_{th}^D (\gamma_{th}^D + 1)}{\omega_k \tilde{\gamma}_{RD}} + \frac{\gamma_{th}^E (\gamma_{th}^E + 1)}{\tilde{\gamma}_{RE} + \gamma_{th}^D - \gamma_{th}^E} \right)^{\frac{m + 1}{2}} K_{m + 1} \left( 2 \sqrt{\frac{\gamma_{th}^D (\gamma_{th}^D + 1)}{\omega_k \tilde{\gamma}_{SR} \tilde{\gamma}_{RD}} + \frac{\gamma_{th}^E (\gamma_{th}^E + 1)}{\tilde{\gamma}_{SR} (\tilde{\gamma}_{RE} + \gamma_{th}^D - \gamma_{th}^E)}} \right) \right] \end{aligned} \quad (20)$$

$$\zeta^{\text{TBRS}}(R_0, R_s, \eta) = R_s \left\{ 1 - \left[ \frac{N_t (1 - \rho_{SR}^2)^{N_t - 1}}{\sigma_{SR}^2} + \sum_{k=0}^{K_r - 1} \frac{K_r (-1)^k}{[k (1 - \rho_{RD}^2) + 1] \sigma_{RD}^2} \binom{K_r - 1}{k} \right] \times \frac{2^{2R_0} - 1}{\eta} \right\} \frac{2^{2(R_0 - R_s)} - 1}{\sigma_{RE}^2 \eta} \quad (21)$$

545 As far as the eavesdropper is concerned,  $\gamma_{R^*E}$  and  $\gamma_{J^*E}$   
546 are independent and exponentially distributed. Furthermore, for  
547  $\xi_E = \gamma_{R^*E}/(\gamma_{J^*E} + 1)$ , we have

$$F_{\xi_E}(x) = 1 - \frac{\phi}{x + \phi} e^{-\frac{x}{\phi}} \quad (25)$$

548 where  $\phi = \lambda/(1 - \lambda)$ . According to the definition of COP and  
549 SOP in Section III-A, we can obtain the following closed-form  
550 approximations of the COP and the SOP.<sup>4</sup>

551 *Lemma 1:* The COP and the SOP of the JRJS strategy  
552 associated with feedback delays are approximated by

$$\begin{aligned} P_{\text{co}}^{\text{JRJS}}(R_0) &\approx 1 - \sum_{n=0}^{N_t-1} \sum_{k=0}^{K_r-1} \sum_{m=0}^{N_t-1-n} \binom{N_t-1}{n} \\ &\times \binom{K_r-1}{k} \binom{N_t-1-n}{m} \\ &\times \frac{(-1)^k (K_r+1) \rho_{SR}^{2(N_t-1-n)} (1 - \rho_{SR}^2)^n}{(N_t-1-n)! (k+1) \bar{\gamma}_{SR}^{N_t-n}} \\ &\times \frac{\Gamma(m+2) \hat{\phi}_k (\gamma_{th}^D)^{N_t-n} (\gamma_{th}^D + 1)^{m+1}}{(\gamma_{th}^D + \hat{\phi}_k)^{m+2}} \\ &\times \exp \left[ -\frac{\gamma_{th}^D (\hat{\phi}_k - 1)}{\bar{\gamma}_{SR} (\gamma_{th}^D + \hat{\phi}_k)} \right] \\ &\times \Gamma \left( -m-1, \frac{\gamma_{th}^D (\gamma_{th}^D + 1)}{\bar{\gamma}_{SR} (\gamma_{th}^D + \hat{\phi}_k)} \right) \end{aligned} \quad (26)$$

553 where  $\hat{\phi}_k = K_r \lambda \omega_k \eta \sigma_{RD}^2 / [(K_r - 1)(1 - \rho_{RD}^2) + 1](1 -$   
554  $\lambda) \eta \sigma_{RD}^2 + K_r)$ , and

$$\begin{aligned} P_{\text{so}}^{\text{JRJS}}(R_0, R_s) &\approx \sum_{n=0}^{N_t-1} \sum_{m=0}^{N_t-1-n} \binom{N_t-1}{n} \\ &\times \frac{\rho_{SR}^{2(N_t-1-n)} (1 - \rho_{SR}^2)^n}{m! \bar{\gamma}_{SR}^m} \\ &\times \frac{(2\gamma_{th}^E)^m \phi}{(2\gamma_{th}^E + \phi)} \exp \left[ -\left( \frac{2\gamma_{th}^E}{\bar{\gamma}_{SR}} + \frac{2\gamma_{th}^E}{\bar{\gamma}_{RE}} \right) \right]. \end{aligned} \quad (27)$$

555 *Proof:* The proof is given in Appendix B.

556 The feasible range of the reliability constraint is similar to  
557 that of the TBRS strategy, and hence, it is omitted here.

### 558 B. Reliability–Security Ratio

559 *Lemma 2:* Recalling the definition in Section III, the RSR  
560 for the JRJS strategy may be expressed in (28), shown at the  
561 bottom of the page.

<sup>4</sup>When we have  $\lambda \rightarrow 1$ , (24) will degenerate into the TBRS case seen in (10). The performance analysis of the JRJS will be presented separately in the following, since several approximations have to be included.

562 It can be seen from the previous expression that, in contrast  
563 to the analysis of the TBRS strategy operating without jam-  
564 ming, for a fixed SNR threshold, the CDF of the second-hop  
565 SNR will converge to a nonzero limit. We also find that this  
566 limit is determined by the power sharing ratio between the  
567 relay and the jammer. Furthermore, according to the analy-  
568 sis of the TBRS strategy, for  $\eta \rightarrow \infty$ , we have  $F_{\gamma_{SR^*}}(x) \rightarrow$   
569 0. Thus, by exploiting the tight upper bound that  $\gamma_D^{\text{TBRS}} \leq$   
570  $\min\{\gamma_{SR}, \gamma_{R^*D}\}$  and  $\gamma_E^{\text{TBRS}} \leq \min\{\gamma_{SR}, \gamma_{R^*E}\}$ , we have  
571  $P_{\text{co}}^{\text{JRJS}, \infty} \rightarrow F_{\gamma_{\xi_D}}(\gamma_{th}^D)$  and  $1 - P_{\text{so}}^{\text{JRJS}, \infty} \rightarrow F_{\gamma_{\xi_E}}(\gamma_{th}^E)$ . Finally,  
572 substituting the corresponding results into (16), we arrive at the  
573 RSR of the JRJS strategy.

*Remark 4:* It can be seen from the RSR expression of (28)  
574 again that the rate-pair setting  $(R_0, R_s)$  has an inconsistent  
575 influence on the RSR, and hence, we have to carefully adjust  $R_0$   
576 and  $R_s$  to balance the reliability versus security performance.  
577 Let us now focus our attention on the differences between the  
578 JRJS scheme and the TBRS arrangement.

579 First, we may find that the power sharing ratio  $\lambda$  between  
580 the relay and the jammer plays a very important role. The  
581 optimization of  $\lambda$  will be investigated from an effective secrecy  
582 throughput optimization point of view in the following.

583 Second, it is plausible that, in contrast to the behavior of the  
584 TBRS strategy,  $\Lambda^{\text{JRJS}}$  of (28) is only related to the delay of the  
585 second hop, but it is still a monotonically decreasing function of  
586  $\rho_{RD}$ . This implies that the improvement of the channel quality  
587 of the JRJS will achieve a more pronounced COP improvement  
588 than the associated SOP improvement. Furthermore, recalling  
589 that the RSR is considered in the high-SNR region, it has no  
590 dependence on the first hop quality. This is due to the fact that  
591 if the first-hop channel quality is sufficiently high for ensuring  
592 a successful transmission, the asymptotic CDFs of  $\xi_D$  and  $\xi_E$   
593 in (29) and (30) associated with  $\eta \rightarrow \infty$  will converge to a  
594 nonzero limit at high SNRs, which ultimately dominates the  
595 COP and the SOP.

### 597 C. Effective Secrecy Throughput

598 Before proceeding to the effective secrecy throughput analy-  
599 sis, we also have to investigate the RSCP.

600 *Lemma 3:* The RSCP of our JRJS strategy may be approxi-  
601 mated as in (31), shown at the bottom of the next page, where  
602 we have  $\theta_{1,k} = (\gamma_{th}^D (\gamma_{th}^D + 1)) / (\gamma_{th}^D + \hat{\phi}_k)$ ,  $\theta_2 = \gamma_{th}^D - \gamma_{th}^E +$   
603  $(\gamma_{th}^E (\gamma_{th}^E + 1)) / (\gamma_{th}^E + \hat{\phi})$ , and  $\hat{\phi} = \lambda \eta \sigma_{RE}^2 / ((1 - \lambda) \eta \sigma_{RE}^2 + 1)$ .

604 *Proof:* The proof is given in Appendix C.

605 Apart from the rate pair  $(R_0, R_s)$ , the aforementioned  $P_{R\&S}^{\text{JRJS}}$   
606 of (31) is also a function of the power sharing ratio  $\lambda$  between  
607 the selected relay and the jammer.

608 Given the complexity of the RSCP expression, it is quite  
609 a challenge to find a closed-form result for maximizing the

$$\Lambda^{\text{JRJS}} = \frac{(2^{2R_0} - 1)}{(2^{2(R_0 - R_s)} - 1)} \sum_{k=0}^{K_r-1} \binom{K_r-1}{k} \frac{(-1)^k K_r [(K_r - 1)(1 - \rho_{RD}^2) + 1] [(\lambda^{-1} - 1)(2^{2(R_0 - R_s)} - 1) + 1]}{[(K_r - 1)(1 - \rho_{RD}^2) + 1] (k+1)(\lambda^{-1} - 1)(2^{2R_0} - 1) + K_r [k(1 - \rho_{RD}^2) + 1]} \quad (28)$$



610 effective secrecy throughput that  $\max_{0 < \lambda < 1} \varsigma = R_s P_{R\&S}^{\text{JRJS}}$ . Al-  
 611 ternatively, we can focus on the asymptotic analysis in the high-  
 612 SNR region and try to find a general closed-form solution for  $\lambda$ .  
 613 Specifically, when we have  $\eta \rightarrow \infty$ ,  $P_{R\&S}^{\text{JRJS}}$  will be dominated  
 614 by the channel quality of the second hop; hence, we have

$$P_{R\&S}^{\text{JRJS},\infty}(R_0, R_s, \lambda) \approx \Pr \{ \xi_D > \gamma_{th}^D, \xi_E < \gamma_{th}^E \} \\ = [1 - F_{\xi_D}(\gamma_{th}^D)] F_{\xi_E}(\gamma_{th}^E) \quad (32)$$

615 where the approximation is based on the fact that, in contrast to  
 616 both  $F_{\xi_D}(\gamma_{th}^D)$  and  $F_{\xi_E}(\gamma_{th}^E)$ , which converge to a nonzero limit  
 617 regardless of  $\eta$ , the first hop's  $F_{\gamma_{SR}}(x)$  will tend to zero, and  
 618 hence, it can be neglected. Substituting the asymptotic results  
 619 of (29) and (30) into (33), we can obtain  $P_{R\&S}^{\text{JRJS},\infty}$ . In contrast to  
 620 the TBRS case operating without jamming, as the SNR tends to  
 621  $\infty$ , the RSCP will tend to a nonzero value and, upon increasing  
 622 the transmit SNR beyond a certain limit, will no longer increase  
 623 the effective secrecy throughput.

624 Then, based on (32), we arrive at the approximated optimal  
 625 value  $\lambda_{\text{opt}}$ , which is the solution of the following equation:

$$\frac{\partial P_{R\&S}^{\text{JRJS},\infty}(R_0, R_s, \lambda)}{\partial \lambda} = 0. \quad (33)$$

626 Then, by exploiting the approximation of  $[k(1 - \rho_{RD}^2) + 1]/$   
 627  $(k + 1) \approx 1 - \rho_{RD}^2$  in (29) for a large  $\rho_{RD}$  (practically, the CSI  
 628 delay is small, and  $\rho_{RD} \rightarrow 1$ ), we have

$$\lambda_{\text{subopt}} = \frac{\sqrt{[(K_r - 1)(1 - \rho_{RD}^2) + 1] \gamma_{th}}}{\sqrt{[(K_r - 1)(1 - \rho_{RD}^2) + 1] \gamma_{th} + \sqrt{K_r(1 - \rho_{RD}^2)}}} \quad (34)$$

629 where  $\gamma_{th} = (2^{2R_0} - 1)(2^{2(R_0 - R_s)} - 1)$ . It is clear that this  
 630 value is determined by the number of relays and  $(R_0, R_s)$ .

## 631 V. NUMERICAL RESULTS

632 Both our numerical and Monte Carlo simulation results are  
 633 presented here for verifying the theoretical PLS performance  
 634 analysis of the multiple-relay-aided network under CSI feed-

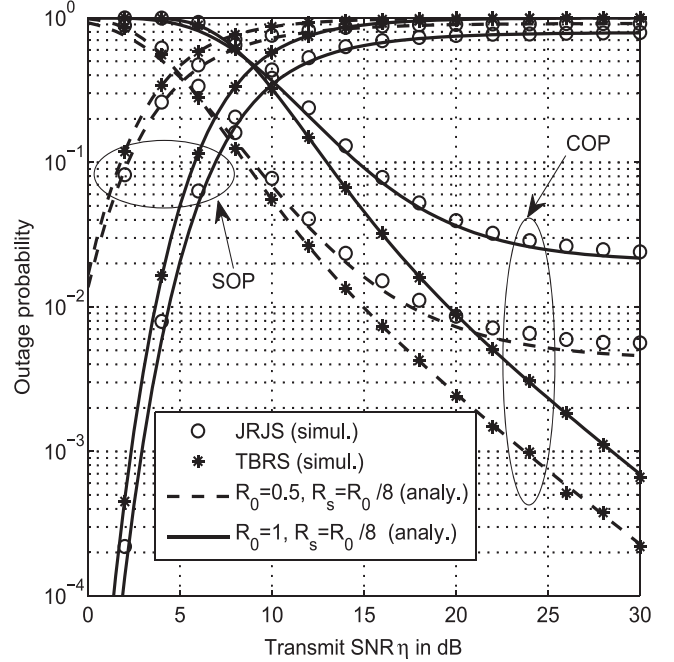


Fig. 2. COP and SOP versus transmit SNR for the TBRS and JRJS strategies in conjunction with different rate pairs, for  $N_t = K_r = 3$ ,  $f_d T_d = 0.1$ , and  $\lambda = 1/10$ .

back delays. Explicitly, the COP, SOP, RSCP, and RSR are  
 635 validated for both the TBRS and JRJS strategies. Furthermore,  
 636 the effects of feedback delays and system parameters (including  
 637 the transmission rate pair  $(R_0, R_s)$  and the power sharing ratio  
 638  $\lambda$  between the relay and the jammer) on the achievable effective  
 639 secrecy throughput are evaluated. The Rayleigh fading model  
 640 is employed for characterizing all communication links in our  
 641 system. Additionally, we set the total power to  $P = 1$  and  
 642  $\sigma_{SR}^2 = \sigma_{RD}^2 = \sigma_{RE}^2 = 1$ , and used  $T_{dSR} = T_{dRD} = T_d$ .  
 643

644 Fig. 2 plots the COP and the SOP versus the transmit SNR for  
 645 both the TBRS and JRJS strategies in conjunction with different  
 646 rate pairs. The analytical lines are plotted by using (11) and (14)  
 647 for the TBRS strategy and by using (26) and (27) for the JRJS

$$P_{R\&S}^{\text{JRJS}}(R_0, R_s, \lambda) \approx \sum_{n=0}^{N_t-1} \sum_{k=0}^{K_r-1} \sum_{m=0}^{N_t-1-n} (-1)^k \binom{N_t-1}{n} \binom{K_r-1}{k} \binom{N_t-1-n}{m} \\ \times \frac{K_r \rho_{SR}^{2(N_t-1-n)} (1 - \rho_{SR}^2)^n \hat{\varphi}_k(\gamma_{th}^D)^{N_t-1-n-m}}{(N_t-1-n)! (k+1) \bar{\gamma}_{SR}^{N_t-n} (\gamma_{th}^D + \hat{\varphi}_k) e^{\frac{\gamma_{th}^D}{\bar{\gamma}_{SR}} + \frac{\gamma_{th}^D}{\bar{\gamma}_{RD} \omega_k}}} \\ \times \left\{ \theta_{1,k}^{m+1} e^{\frac{\theta_1}{\bar{\gamma}_{SR}}} \Gamma(m+2) \Gamma\left(-m-1, \frac{\theta_{1,k}}{\bar{\gamma}_{SR}}\right) - \frac{\hat{\phi} e^{-\gamma_{th}^E / \bar{\gamma}_{RE}}}{(\gamma_{th}^E + \phi) (\theta_{1,k} - \theta_2)} \Gamma(m+3) \right. \\ \times \left[ \theta_2^{m+2} e^{\frac{\theta_2}{\bar{\gamma}_{SR}}} \Gamma\left(-m-2, \frac{\theta_2}{\bar{\gamma}_{SR}}\right) - \theta_{1,k}^{m+2} e^{\frac{\theta_1}{\bar{\gamma}_{SR}}} \Gamma\left(-m-2, \frac{\theta_{1,k}}{\bar{\gamma}_{SR}}\right) \right] + \Gamma(m+2) (\gamma_{th}^D - \gamma_{th}^E) \\ \left. \times \left[ \theta_2^{m+1} e^{\frac{\theta_2}{\bar{\gamma}_{SR}}} \Gamma\left(-m-1, \frac{\theta_2}{\bar{\gamma}_{SR}}\right) - \theta_{1,k}^{m+1} e^{\frac{\theta_1}{\bar{\gamma}_{SR}}} \Gamma\left(-m-1, \frac{\theta_{1,k}}{\bar{\gamma}_{SR}}\right) \right] \right\} \quad (31)$$

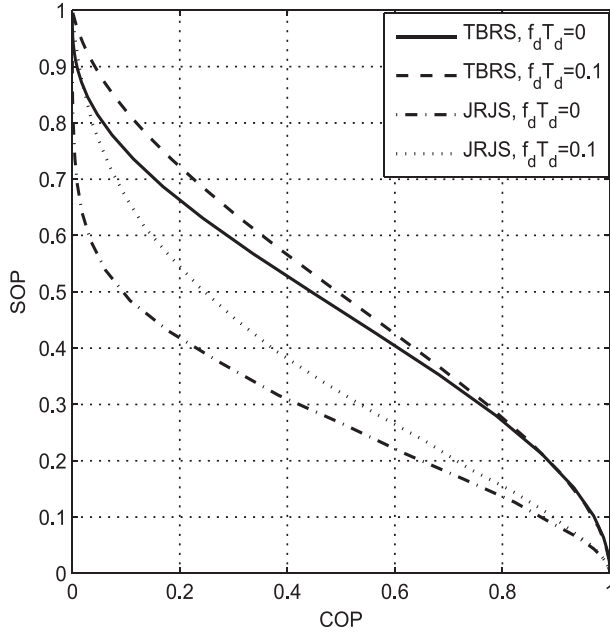


Fig. 3. SOP versus COP for the TBRS and JRJS strategies with different feedback delays for  $N_t = K_r = 3$ ,  $R_s = R_0/8$ , and  $\lambda = 1/10$ .

648 case, respectively. It can be clearly seen from the figure that the  
649 analytical and simulated outage probability curves match well,  
650 which confirms the accuracy of the mathematical analysis. As  
651 expected, compared with the TBRS strategy, the SOP of the  
652 JRJS strategy is much better, whereas the COP is worse. We  
653 can also find that both the COP and the SOP will converge to an  
654 outage floor at high SNRs for the JRJS strategy. The reason for  
655 this is that the jammer also imposes interference on the destina-  
656 tion and the interference inflicted increases with the SNR. Thus,  
657 the designers have to take into account the tradeoff between  
658 the reliability and the security and the interference imposed on  
659  $D$ , particularly when considering the JRJS strategy. Moreover,  
660 we can observe in Fig. 2 that increasing the transmission rate  
661 decreases the COP and increases the SOP.

662 Fig. 3 further characterizes the SOP versus COP for both the  
663 TBRS and JRJS strategies based on the numerical results in  
664 Fig. 2, which shows the tradeoff between the reliability and the  
665 security. It can be seen from the figure that the SOP decreases as  
666 the COP increases, and for a specific COP, the SOP of the JRJS  
667 scheme is strictly lower than that of TBRS. This confirms that  
668 the JRJS scheme performs better than the conventional TBRS  
669 scheme. Furthermore, the CSI feedback delay will also degrade  
670 the system tradeoff performance.

671 Fig. 4 illustrates the RSCP versus transmit SNR for the  
672 TBRS strategy in the context of different network configura-  
673 tions, including different rate pairs, different number of relays,  
674 and both perfect and outdated CSI feedback scenarios. The  
675 analytical lines are plotted by using the approximation in (20).  
676 We may conclude from the figure that the rate-pair setting  
677 ( $R_0, R_s$ ) determines both the reliability and security transmis-  
678 sion performance. These curves also show that the RSCP is a  
679 concave function of the transmit SNR, whereas the continued  
680 boosting of the SNR would only decrease the probability of  
681 a successful transmission. We can observe from Fig. 4 that,

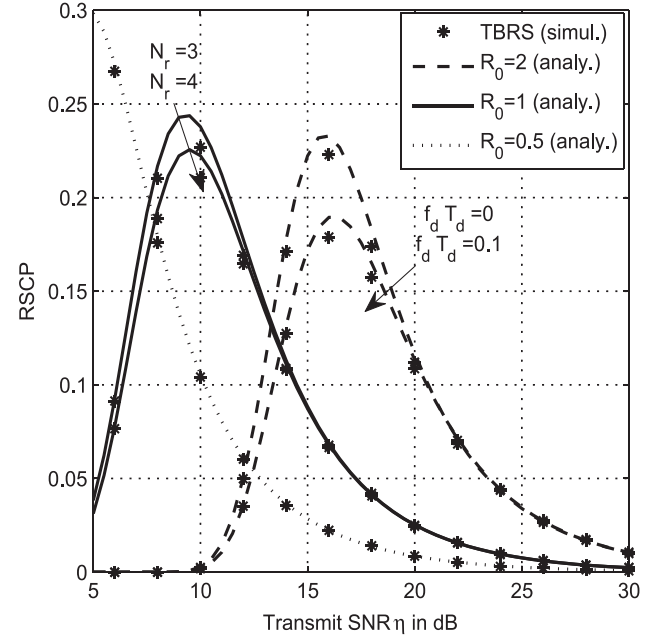


Fig. 4. RSCP versus transmit SNR for the TBRS strategy with different rate pairs for  $N_t = K_r = 3$ ,  $f_d T_d = 0.1$ .

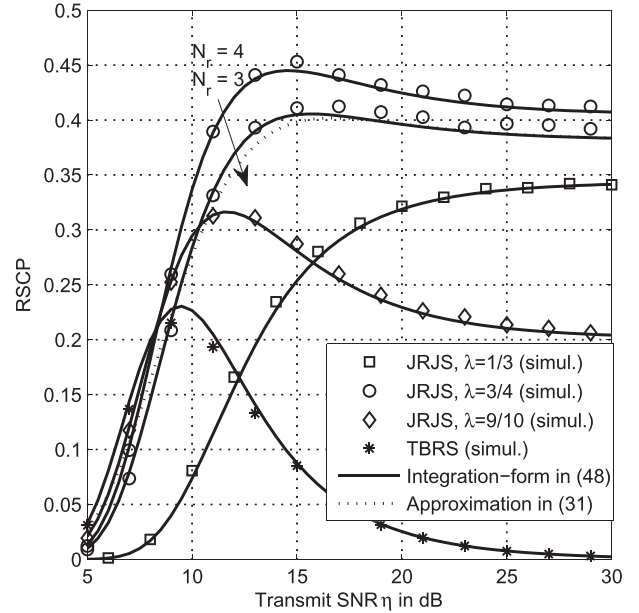


Fig. 5. RSCP versus transmit SNR for the JRJS strategy for different power sharing ratios  $\lambda$  and for  $N_t = K_r = 3$ ,  $f_d T_d = 0.1$ , and  $R_0 = 1$ ,  $R_s = R_0/8$ .

for a high transmit SNR, total reliability can be guaranteed,  
682 whereas the associated grade of security is severely eroded.  
683 Furthermore, increasing the number of relays and decreasing  
684 the feedback delay will improve both the reliability and security  
685 performance. 686

The RSCP of the JRJS strategy is presented in Fig. 5 for  
687 different power sharing ratios between relaying and jamming.  
688 Both the integration form (45) and the approximated closed  
689 form in (31) match well with the Monte Carlo simulations.  
690 The performance of the TBRS strategy is also included for  
691 comparison. The JRJS scheme outperforms the TBRS operating  
692

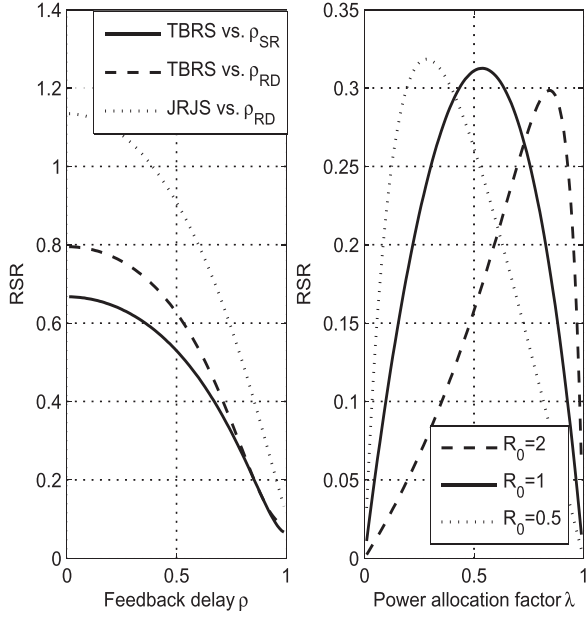


Fig. 6. RSR versus feedback delay coefficient ( $R_0 = 1$ ,  $R_s = R_0/8$ ,  $\lambda = 3/4$ ) and power sharing ratio  $\lambda$  ( $R_s = R_0/8$ ,  $\rho_{SR} = \rho_{RD} = 0.9$ ) for the TBRS and JRJS strategies, with  $N_t = K_r = 3$ .

693 without jamming under the scenario considered when encoun-  
 694 tering comparable relay–destination and relay–eavesdropper  
 695 channels. For some extreme configurations (when the relay–  
 696 eavesdropper links are comparatively weak), this statement  
 697 may not hold, but this scenario is beyond the scope of this  
 698 paper. The maximum RSCP appears at about  $\eta = 15$  dB  
 699 for the JRJS strategy using  $\lambda = 3/4$ , whereas it is  $\eta = 10$  dB  
 700 for the TBRS strategy. Furthermore, as expected, increasing the  
 701 number of available relays and jamming nodes will always be  
 702 able to improve the reliability and security performance. How-  
 703 ever, the continued boosting of the jammer’s power (decreasing  
 704  $\lambda$ ) will not always improve the overall performance, because  
 705 the interference improves initially the security, but then, it starts  
 706 to reduce the reliability as  $\lambda$  decreases. This further motivates  
 707 the designer to carefully take into account the power sharing  
 708 between relaying and jamming. The effect of the rate-pair  
 709 setting on the security and reliability of the JRJS strategy is  
 710 neglected here, which follows a similar trend to that of the  
 711 TBRS strategy.

712 Fig. 6 characterizes the RSR versus feedback delay and  
 713 power sharing ratio for both TBRS and JRJS, in which the  
 714 RSR curves are plotted by using (17) and (28), respectively.  
 715 The first illustration shows that the RSR decreases as the delay  
 716 coefficients ( $\rho_{SR}$  and  $\rho_{RD}$ ), which confirms that the im-  
 717 provement of reliability becomes more pronounced than the  
 718 reduction of the security as the feedback delay decreases.  
 719 This observation implies an improvement in terms of the  
 720 security–reliability tradeoff. In addition, the RSR versus  $\rho_{RD}$   
 721 is larger than that of  $\rho_{SR}$ , which indicates that the impact of the  
 722 second-hop CSI feedback delay is more prominent. The other  
 723 illustration in the right demonstrates that the RSR is a concave  
 724 function of the power sharing ratio, which reflects the tradeoff  
 725 between the reliability and the security struck by adjusting  $\lambda$ .

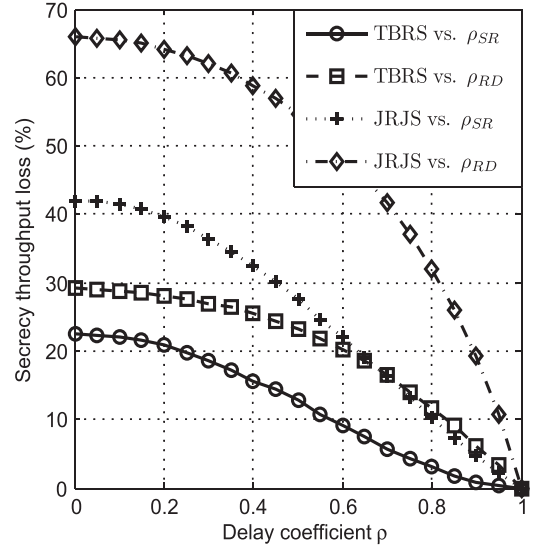


Fig. 7. Percentage secrecy throughput loss versus delay coefficients with  $N_t = K_r = 3$ ,  $R_0 = 1$ ,  $R_s = R_0/8$ ,  $\lambda = 3/4$ , and  $\eta = 10$  dB.

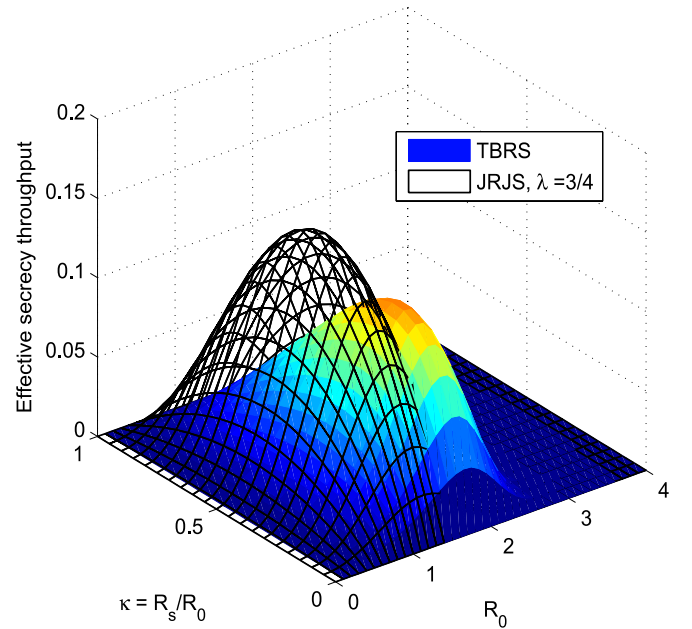


Fig. 8. Secrecy throughput versus  $R_0$  and  $\kappa = R_s/R_0$  for both the TBRS and JRJS strategies with  $N_t = K_r = 3$ ,  $f_d T_d = 0.1$ , and  $\eta = 15$  dB.

To further evaluate the effect of feedback delays on the  
 726 secrecy performance, Fig. 7 plots the resultant percentage of 727  
 728 secrecy throughput loss versus the delay, which is defined as

$$\text{Loss} = \frac{S_{\text{no-delay}} - S_{\text{delay}}}{S_{\text{no-delay}}}. \quad (35)$$

It can be seen from the figure that, compared with the TBRS  
 729 scheme, JRJS is more sensitive to the feedback delays. Further-  
 730 more, recalling that increasing the delay coefficient  $\rho_{SR}$  of the  
 731 first hop improves the reliability, but at the same time also helps  
 732 the eavesdropper, it is not surprising that the secrecy throughput  
 733 loss due to the second-hop feedback delay is more pronounced. 734

Fig. 8 illustrates the achievable effective secrecy throughput  
 735 for both the TBRS and JRJS strategies versus the codeword 736

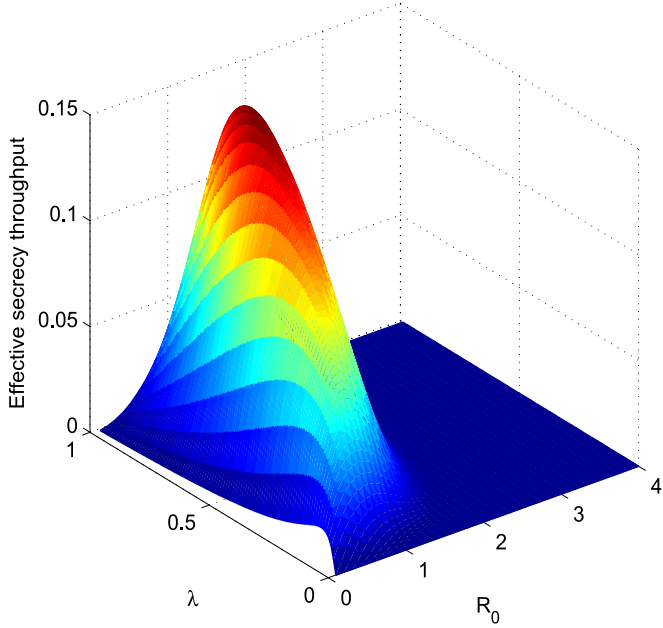


Fig. 9. Secrecy throughput versus  $R_0$  and  $\lambda$  for the JRJS strategy with  $N_t = K_r = 3$ ,  $f_d T_d = 0.1$ ,  $\eta = 15$  dB, and  $R_s/R_0 = 1/8$ .

737 transmission rate  $R_0$  and the secrecy code ratio  $\kappa = R_s/R_0$   
 738 with no outage constraints ( $v = \delta = 1$ ). The values of the  
 739 effective secrecy throughput are plotted by using  $\zeta = R_s P_{R\&S}$ .  
 740 We can observe in Fig. 8 that, subject to a fixed code rate  
 741 ratio  $\kappa$ , the effective secrecy throughput increases to a peak  
 742 value as  $R_0$  reaches its optimal value and then decreases. This  
 743 phenomenon can be explained as follows. At a low transmission  
 744 rate, although the COP increases with  $R_0$ , which has a negative  
 745 effect on the effective secrecy throughput, both the secrecy  
 746 rate and the SOP performance will benefit. However, after  
 747 reaching the optimal  $R_0$ , the effective secrecy throughput drops  
 748 since the main link cannot afford a reliable transmission, and  
 749 the resultant COP increase becomes dominant. On the other  
 750 hand, subject to a fixed  $R_0$  (which results in a constant COP),  
 751 the effective secrecy throughput is also a concave function  
 752 of  $\kappa$ , and increasing the code rate ratio ultimately results  
 753 in an increased secrecy information rate at the cost of an  
 754 increased SOP.

755 The achievable effective secrecy throughput for the JRJS  
 756 strategy is also presented in Fig. 8, and similar conclusions and  
 757 trends can be observed to that of the TBRS case. Additionally,  
 758 the comparison of the two strategies indicates that the JRJS  
 759 scheme attains a higher effective secrecy throughput than the  
 760 TBRS scheme operating without jamming, even if no power  
 761 sharing optimization has been employed.

762 Fig. 9 further illustrates the impact of power sharing between  
 763 the relay and the jammer on the achievable effective secrecy  
 764 throughput of the JRJS strategy versus  $R_0$  in the absence of  
 765 outage constraints. Given a fixed code rate pair  $(R_0, R_s)$ , the  
 766 effective secrecy throughput follows the trend of the RSCP,  
 767 which is a concave function of  $\lambda$ , as shown in Fig. 6. The  
 768 interference introduced by the jammer initially improves both  
 769 the reliability and the security as  $\lambda$  increases, but this trend is  
 770 reversed beyond a certain point.

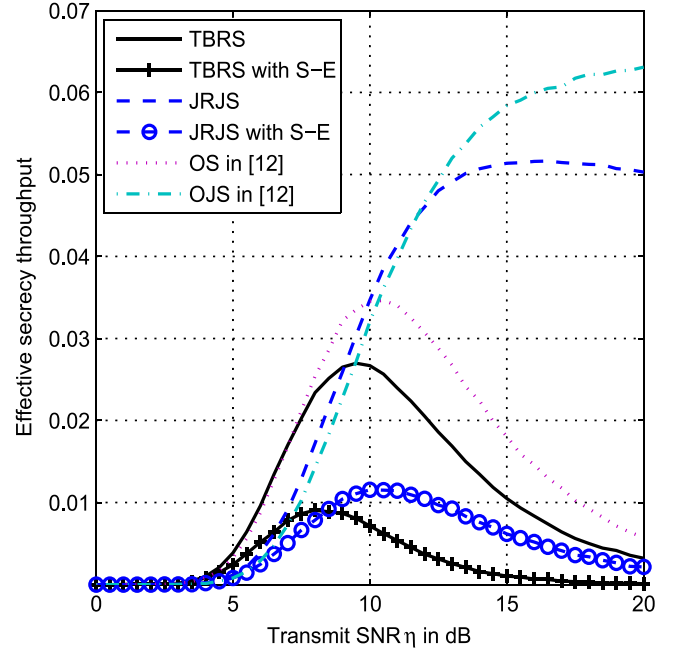


Fig. 10. Comparisons for different strategies with and without the S-E link, for  $N_t = K_r = 3$ ,  $R_0 = 1$ ,  $R_s = R_0/8$ ,  $f_d T_d = 0.1$ , and  $\lambda = 3/4$ .

## VI. DISCUSSION

771

### A. Impact of the S-E Link

772

We note that the introduction of the S-E link, i.e., the  
 773 information leakage in the first phase, is very critical to the  
 774 security. There are also some research studies focusing on  
 775 the corresponding secure transmission design and performance  
 776 evaluation for cooperative networks with the S-E link, such  
 777 as [15] and [16]. Here, we assume that the eavesdropper can  
 778 receive information directly from the source in the first phase.  
 779 Thus, following the steps in the prior sections, for the TBRS  
 780 and JRJS schemes, it is clear that the SNR experienced at the  
 781 eavesdropper should be rewritten as

$$\tilde{\gamma}_E^\tau = \gamma_{SE} + \gamma_E^\tau \quad (36)$$

where  $\gamma_{SE} = P_s |\mathbf{w}_{\text{opt}}(t|T_{dSR}) \mathbf{h}_{SE}(t)|^2 / N_0$  follows the ex-  
 783 ponential distribution with the average value  $\bar{\gamma}_{SE}$ ,  $\tau =$   
 784  $\{\text{TBRS, JRJS}\}$ , and  $\gamma_E^\tau$  has been defined in (4) and (7).  
 785

Then, the corresponding SOP, RSCP, and effective secrecy  
 786 throughput have to be reconsidered. Unfortunately, to the best  
 787 of our knowledge, it is a mathematically intractable problem  
 788 to obtain closed-form results for the related performance eval-  
 789 uations. Therefore, we resorted to numerical simulations for  
 790 further investigating the impact of the S-E link. Fig. 10 com-  
 791 pares the effective secrecy throughput of the TBRS and JRJS  
 792 schemes both with and without considering the direct S-E  
 793 link. It becomes clear that the information leakage in the first  
 794 phase will lead to a severe security performance degradation,  
 795 particularly for the JRJS scheme, which will no longer be  
 796 capable of maintaining a steady throughput at high SNRs. The  
 797 reason for this trend is that increasing the transmit SNR will  
 798 help the eavesdropper in the presence of the direct S-E link.  
 799

## 800 B. Comparisons

801 Here, based on the outdated CSI assumption, we provide per-  
802 formance comparisons with a range of other schemes advocated  
803 in [12] with the aid of the proposed outage-based characteriza-  
804 tion. Fig. 10 also incorporates our effective secrecy throughput  
805 performance comparison, where the optimal selection (OS)  
806 regime and the optimal selection combined with jamming (OSJ)  
807 were proposed in [12]. They are formulated as

$$808 \text{ OS : } R^* = \arg \max_{R_k \in \mathcal{R}} \left\{ \frac{\tilde{\gamma}_{R_k D}}{\tilde{\gamma}_{R_k E}} \right\} \quad (37)$$

$$809 \text{ OSJ : } \left\{ \begin{array}{l} R^* = \arg \max_{R_k \in \mathcal{R}} \left\{ \frac{\tilde{\gamma}_{R_k D}}{\tilde{\gamma}_{R_k E}} \right\} \\ J^* = \arg \min_{R_k \in \mathcal{R} - R^*} \left\{ \frac{\tilde{\gamma}_{R_k D}}{\tilde{\gamma}_{R_k E}} \right\} \end{array} \right\} \quad (38)$$

810 where  $\tilde{\gamma}_{R_k E}$  is the delayed version of the instantaneous CSI of  
811 the R–E link. It should be noted that this constitutes an entirely  
812 new performance characterization of these schemes from the  
813 perspective of the effective secrecy throughput. It is shown in  
814 Fig. 1 that the selection combined with jamming outperforms  
815 the corresponding nonjamming techniques at high SNRs, albeit  
816 this trend may no longer prevail at low SNRs. In comparison,  
817 compared with those selections relying on the average SNRs of  
818 the R–E link, the optimal selections relying on the idealized  
819 simplifying assumptions of having global CSI (OS and OSJ  
820 schemes) knowledge can only achieve throughput gains at high  
821 SNRs due to the inevitable feedback delay.

820

## VII. CONCLUSION

821 An outage-based characterization of cooperative relay net-  
822 works has been provided in the face of CSI feedback delays.  
823 Two types of relaying strategies were considered, namely, the  
824 TBRS strategy and the JRJS strategy. Closed-form expressions  
825 of the COP, the SOP, and the RSCP, as well as of the RSR,  
826 were derived. The RSR results demonstrated that the reliability  
827 is improved more substantially than the security performance  
828 when the CSI feedback delays are reduced. Furthermore, we  
829 presented a modified effective secrecy throughput definition  
830 and demonstrated that the JRJS strategy achieves a significant  
831 effective secrecy throughput gain over the TBRS strategy. The  
832 transmit SNR, the secrecy codeword rate setting, and the power  
833 sharing ratio between the relay and jammer nodes play impor-  
834 tant roles in striking a balance between the reliability and the  
835 security in terms of the secrecy throughput. The impact of the  
836 direct S–E link and the performance comparisons with other  
837 selection schemes were also included. Additionally, our results  
838 demonstrate that JRJS is more sensitive to the feedback delays  
839 and that the secrecy throughput loss due to the second-hop  
840 feedback delay is more pronounced than that due to the first-  
841 hop one.

842

## APPENDIX A

843

### PROOF OF PROPOSITION 1

844 To simplify the asymptotic performance analysis, (3) can be  
845 expressed in a more mathematically tractable form by the com-  
846 monly used tight upper bound of  $\gamma_D^{\text{TBRS}} \leq \min\{\gamma_{SR}, \gamma_{R^*D}\}$

and  $\gamma_E^{\text{TBRS}} \leq \min\{\gamma_{SR}, \gamma_{R^*E}\}$ . When we have  $\eta \rightarrow \infty$ , based  
847 on the CDFs in (9) and (10) and closing the smallest order terms  
848 of  $x/\eta$ , we have 849

$$\begin{aligned} F_{\gamma_{SR}}(x) &\rightarrow 1 - \left[ \sum_{n=0}^{N_t-1} \binom{N_t-1}{n} \rho_{SR}^{2(N_t-1-n)} (1 - \rho_{SR}^2)^n \right. \\ &\quad \left. + \sum_{n=0}^{N_t-2} \binom{N_t-1}{n} \times \rho_{SR}^{2(N_t-1-n)} \right. \\ &\quad \left. \times (1 - \rho_{SR}^2)^n \frac{x}{\tilde{\gamma}_{SR}} + \mathcal{O}\left(\frac{x}{\tilde{\gamma}_{SR}}\right) \right] \\ &\quad \times \left[ 1 - \frac{x}{\tilde{\gamma}_{SR}} + \mathcal{O}\left(\frac{x}{\tilde{\gamma}_{SR}}\right) \right] \\ &= 1 - \left[ 1 + (1 - (1 - \rho_{SR}^2)^{N_t-1}) \frac{x}{\tilde{\gamma}_{SR}} + \mathcal{O}\left(\frac{x}{\tilde{\gamma}_{SR}}\right) \right] \\ &\quad \times \left[ 1 - \frac{x}{\tilde{\gamma}_{SR}} + \mathcal{O}\left(\frac{x}{\tilde{\gamma}_{SR}}\right) \right] \\ &= (1 - \rho_{SR}^2)^{N_t-1} \frac{x}{\tilde{\gamma}_{SR}} + \mathcal{O}\left(\frac{x}{\tilde{\gamma}_{SR}}\right) \quad (39) \end{aligned}$$

where  $\mathcal{O}(x)$  denotes the high-order infinitely small contribu-  
850 tions as a function of  $x$ , and 851

$$\begin{aligned} F_{\gamma_{R^*D}}(x) &\rightarrow 1 - \sum_{k=0}^{K_r-1} (-1)^k \frac{K_r}{k+1} \binom{K_r-1}{k} \\ &\quad \times \left[ 1 - \frac{k+1}{k(1 - \rho_{RD}^2) + 1} \frac{x}{\tilde{\gamma}_{RD}} + \mathcal{O}\left(\frac{x}{\tilde{\gamma}_{RD}}\right) \right] \\ &= \sum_{k=0}^{K_r-1} (-1)^k \binom{K_r-1}{k} \frac{K_r}{k(1 - \rho_{RD}^2) + 1} \\ &\quad \times \frac{x}{\tilde{\gamma}_{RD}} + \mathcal{O}\left(\frac{x}{\tilde{\gamma}_{RD}}\right). \quad (40) \end{aligned}$$

Then, applying the upper bound of the receiver SNR, we may  
852 rewrite the COP and the SOP of the TBRS strategy at high  
853 SNRs as 854

$$\begin{aligned} P_{\text{co}}^{\text{TBRS},\infty} &= 1 - (1 - F_{\gamma_{SR^*}}(\gamma_{th}^D)) (1 - F_{\gamma_{R^*D}}(\gamma_{th}^D)) \\ &= \left[ \frac{(1 - \rho_{SR}^2)^{N_t-1}}{\sigma_{SR}^2} + \sum_{k=0}^{K_r-1} (-1)^k \right. \\ &\quad \left. \times \binom{K_r-1}{k} \frac{K_r}{[k(1 - \rho_{RD}^2) + 1] \sigma_{RD}^2} \right] \frac{2^{2R_0} - 1}{\eta} \quad (41) \end{aligned}$$

and according to the fact that  $\gamma_{R^*E}$  is exponentially distributed,  
855 we have 856

$$\begin{aligned} 1 - P_{\text{so}}^{\text{TBRS},\infty} &= 1 - (1 - F_{\gamma_{SR^*}}(\gamma_{th}^E)) (1 - F_{\gamma_{R^*E}}(\gamma_{th}^E)) \\ &= \left[ \frac{(1 - \rho_{SR}^2)^{N_t-1}}{\sigma_{SR}^2} + \frac{1}{\sigma_{RE}^2} \right] \frac{2^{2(R_0 - R_s)} - 1}{\eta}. \quad (42) \end{aligned}$$

Finally, substituting (41) and (42) into the definition of RSR  
857 in (16), we can obtain (17). 858



859 APPENDIX B  
860 PROOF OF LEMMA 1

861 According to the description of COP and SOP, replacing  
862  $F_{\gamma_{R^*D}}(x)$  and  $F_{\gamma_{R^*E}}(x)$  by  $F_{\xi_D}(x)$  and  $F_{\xi_E}(x)$  in (12) and (14)  
863 will involve a mathematically intractable integration of the form

$$\Upsilon(a, b, \mu, \nu) = \int_0^{\infty} \frac{z^a}{z+b} \exp\left(-\mu z - \frac{\nu}{z}\right) dz \quad (43)$$

864 which, to the best of our knowledge, does not have a closed-  
865 form solution. Alternatively, bearing in mind that the preceding  
866 integration has a great matter with  $\xi_D$ , we now focus our  
867 attention on the approximation of  $\xi_D$ . Based on the PDF  
868 results in (23), it may be seen that  $\gamma_{J^*D}$  obeys an exponential  
869 distribution. Then, we can approximate  $\hat{\gamma}_{J^*D} = \gamma_{J^*D} + 1$  by  
870 the exponential distribution as well, with an average value  
871 of  $\mathbb{E}\{\hat{\gamma}_{J^*D}\} = ((K_r - 1)(1 - \rho_{RD}^2) + 1)\bar{\gamma}_{RD} + K_r)/K_r$  by  
872 assuming that the AWGN term "1" is part of the stochastic  
873 mean terms. The approximation based on this method provides  
874 a very accurate analysis, and the accuracy of this method is  
875 verified by the numerical results of [34]. Thus, the CDF of  
876  $\hat{\xi}_D = \gamma_{R^*D}/\hat{\gamma}_{J^*D}$  can be derived as

$$F_{\hat{\xi}_D}(x) = \sum_{k=0}^{K_r-1} (-1)^k \binom{K_r-1}{k} \frac{K_r}{k+1} \frac{x}{x+\hat{\varphi}_k} \quad (44)$$

877 where  $\hat{\varphi}_k = \mathbb{E}\{\gamma_{R^*D}\}/\mathbb{E}\{\hat{\gamma}_{J^*D}\}$ .

878 Then, substituting (44) into (11), we have

$$\begin{aligned} & F_{\gamma_{D}^{\text{JRJS}}}(x) \\ & \approx \sum_{n=0}^{N_t-1} \sum_{k=0}^{K_r-1} \sum_{m=0}^{N_t-1-n} \binom{N_t-1}{n} \binom{K_r-1}{k} \binom{N_t-1-n}{m} \\ & \times \frac{(-1)^k K_r \rho_{SR}^{2(N_t-1-n)} (1-\rho_{SR}^2)^n \varphi_k x^{N_t-1-n-m} e^{-\frac{x}{\bar{\gamma}_{SR}}}}{(N_t-1-n)!(k+1)\bar{\gamma}_{SR}^{N_t-n}(x+\varphi_k)} \\ & \times \int_0^{\infty} \frac{z^{m+1}}{z+\frac{x(x+1)}{x+\varphi_k}} \exp\left(-\frac{z}{\bar{\gamma}_{SR}}\right) dz. \end{aligned} \quad (45)$$

879 Using [33, eq. (3.383.10)], we can obtain the CDF of  $\gamma_D^{\text{JRJS}}$  as

$$\begin{aligned} F_{\gamma_D^{\text{JRJS}}}(x) & \approx 1 - \sum_{n=0}^{N_t-1} \sum_{k=0}^{K_r-1} \sum_{m=0}^{N_t-1-n} \binom{N_t-1}{n} \\ & \times \binom{K_r-1}{k} \binom{N_t-1-n}{m} \\ & \times \frac{(-1)^k (K_r+1) \rho_{SR}^{2(N_t-1-n)} (1-\rho_{SR}^2)^n}{(N_t-1-n)!(k+1)\bar{\gamma}_{SR}^{N_t-n}} \\ & \times \frac{\Gamma(m+2) \hat{\varphi}_k x^{N_t-n} (x+1)^{m+1}}{(x+\hat{\varphi}_k)^{m+2}} \\ & \times \exp\left[-\frac{x(\hat{\varphi}_k-1)}{\bar{\gamma}_{SR}(x+\hat{\varphi}_k)}\right] \\ & \times \Gamma\left(-m-1, \frac{x(x+1)}{\bar{\gamma}_{SR}(x+\hat{\varphi}_k)}\right). \end{aligned} \quad (46)$$

880 Finally, substituting  $x = \gamma_{th}^D$  into (46), we obtain  $P_{\text{co}}^{\text{JRJS}}$ .

As far as the SOP is considered, we exploit the commonly  
used tight upper bound of  $\gamma_E^{\text{JRJS}} \geq (1/2) \min\{\gamma_{SR}, \xi_E\}$  to  
calculate it, which may be rewritten as

$$\begin{aligned} P_{\text{so}}^{\text{JRJS}} & \approx \Pr\left\{\frac{1}{2} \min\{\gamma_{SR}, \xi_E\} > \gamma_{th}^E\right\} \\ & = [1 - F_{\gamma_{SR}}(2\gamma_{th}^E)] [1 - F_{\xi_E}(2\gamma_{th}^E)]. \end{aligned} \quad (47)$$

Substituting (9) and (25) into (47), we obtain  $P_{\text{so}}^{\text{JRJS}}$ .

APPENDIX C  
PROOF OF LEMMA 3

According to the definition of the RSCP in (18), we can  
calculate it by

$$\begin{aligned} P_{RS}^{\text{JRJS}} & = \int_0^{\infty} \left[1 - F_{\xi_D}\left(\gamma_{th}^D + \frac{\gamma_{th}^D(\gamma_{th}^D+1)}{z}\right)\right] \\ & \times F_{\xi_E}\left(\gamma_{th}^E + \frac{\gamma_{th}^E(\gamma_{th}^E+1)}{z+\gamma_{th}^D-\gamma_{th}^E}\right) f_{\gamma_{SR^*}}(z+\gamma_{th}^D) dz. \end{aligned} \quad (48)$$

To make the integration mathematically tractable, we invoke  
a simple approximation for  $F_{\xi_E}(x)$  by treating the AWGN term  
"1" in  $\xi_E = \gamma_{R^*E}/(\gamma_{J^*E}+1)$  as part of the stochastic mean  
terms. Hence, we have

$$F_{\xi_E}(x) = \frac{x}{x+\hat{\phi}} \quad (49)$$

where  $\hat{\phi} = \lambda\eta\sigma_{RE}^2/((1-\lambda)\eta\sigma_{RE}^2+1)$ .

Then, replacing the corresponding CDFs of the second hop  
with  $F_{\hat{\xi}_D}(x)$  and  $F_{\hat{\xi}_E}(x)$  in (26), the integration can be derived as

$$\begin{aligned} P_{RS}^{\text{JRJS}} & \approx 1 - \sum_{n=0}^{N_t-1} \sum_{k=0}^{K_r-1} \sum_{m=0}^{N_t-1-n} (-1)^k \binom{N_t-1}{n} \\ & \times \binom{K_r-1}{k} \binom{N_t-1-n}{m} \\ & \times \frac{(K_r+1) \rho_{SR}^{2(N_t-1-n)} (1-\rho_{SR}^2)^n}{(N_t-1-n)!(k+1)\bar{\gamma}_{SR}^{N_t-n}} \\ & \times \frac{\hat{\varphi}_k (\gamma_{th}^D)^{N_t-1-n-m}}{\gamma_{th}^D + \hat{\varphi}_k} \exp\left(-\frac{\gamma_{th}^D}{\bar{\gamma}_{SR}} - \frac{\gamma_{th}^D}{\omega_k \bar{\gamma}_{RD}}\right) \\ & \times \int_0^{\infty} e^{-\frac{z}{\bar{\gamma}_{SR}}} z^{m+1} \left[ \frac{1}{z+\theta_{1,k}} - \frac{\hat{\phi} (z+\gamma_{th}^D-\gamma_{th}^E) e^{-\frac{\gamma_{th}^E}{\bar{\gamma}_{RE}}}}{(\gamma_{th}^E+\hat{\phi})(\theta_{1,k}-\theta_2)} \right] \\ & \times \left( \frac{1}{z+\theta_2} - \frac{1}{z+\theta_{1,k}} \right) dz \end{aligned} \quad (50)$$

where  $\hat{\varphi}_k$  and  $\hat{\phi}$  are introduced by relying on the similar approx-  
imation as in Appendix B. Then, using [33, eq. (3.383.10)], we  
obtain  $P_{R\&S}^{\text{JRJS}}$ .

900 [1] B. Schneier, "Cryptographic design vulnerabilities," *Computer*, vol. 31,  
901 no. 9, pp. 29–33, Sep. 1998.

902 [2] A. D. Wyner, "The wire-tap channel," *Bell Syst. Techn. J.*, vol. 54, no. 8,  
903 pp. 1355–1387, Oct. 1975.

904 [3] I. Csiszar and J. Korner, "Broadcast channels with confidential messages,"  
905 *IEEE Trans. Inf. Theory*, vol. IT-24, no. 3, pp. 339–348, May 1978.

906 [4] W. K. Harrison, J. Almeida, M. R. Bloch, S. W. McLaughlin, and  
907 J. Barros, "Coding for secrecy: An overview of error-control coding techni-  
908 ques for physical-layer security," *IEEE Signal Process. Mag.*, vol. 30,  
909 no. 5, pp. 41–50, Sep. 2013.

910 [5] P. K. Gopala, L. Lai, and H. E. Gamal, "On the secrecy capacity of  
911 fading channels," *IEEE Trans. Inf. Theory*, vol. 54, no. 10, pp. 4687–4698,  
912 Oct. 2008.

913 [6] Y. W. P. Hong, P. C. Lan, and C. C. J. Kuo, "Enhancing physical-layer  
914 secrecy in multi-antenna wireless systems: An overview of signal process-  
915 ing approaches," *IEEE Signal Process. Mag.*, vol. 30, no. 5, pp. 29–40,  
916 Sep. 2013.

917 [7] R. Bassily *et al.*, "Cooperative security at the physical layer: A summary  
918 of recent advances," *IEEE Signal Process. Mag.*, vol. 30, no. 5, pp. 16–28,  
919 Sep. 2013.

920 [8] L. Dong, Z. Han, A. P. Petropulu, and H. V. Poor, "Improving wire-  
921 less physical layer security via cooperating relays," *IEEE Trans. Signal  
922 Process.*, vol. 58, no. 3, pp. 1875–1888, Mar. 2010.

923 [9] J. Huang and A. L. Swindlehurst, "Cooperative jamming for secure com-  
924 munications in MIMO relay networks," *IEEE Trans. Signal Process.*,  
925 vol. 59, no. 10, pp. 4871–4884, Oct. 2011.

926 [10] Y. Zou, X. Wang, and W. Shen, "Optimal relay selection for physical-layer  
927 security in cooperative wireless networks," *IEEE J. Sel. Areas Commun.*,  
928 vol. 31, no. 10, pp. 2099–2111, Oct. 2013.

929 [11] Y. Zou, X. Wang, W. Shen, and L. Hanzo, "Security versus reliability  
930 analysis of opportunistic relaying," *IEEE Trans. Veh. Technol.*, vol. 63,  
931 no. 6, pp. 2653–2661, Jul. 2014.

932 [12] I. Krikidis, J. S. Thompson, and S. McLaughlin, "Relay selection  
933 for secure cooperative networks with jamming," *IEEE Trans. Wireless  
934 Commun.*, vol. 8, no. 10, pp. 5003–5011, Oct. 2009.

935 [13] J. Chen, R. Zhang, L. Song, Z. Han, and B. Jiao, "Joint relay and jammer  
936 selection for secure two-way relay networks," *IEEE Trans. Inf. Forensic  
937 Security*, vol. 7, no. 1, pp. 310–320, Feb. 2012.

938 [14] Z. Ding, M. Xu, J. Lu, and F. Liu, "Improving wireless security  
939 for bidirectional communication scenarios," *IEEE Trans. Veh. Technol.*,  
940 vol. 61, no. 6, pp. 2842–2848, Jul. 2012.

941 [15] C. Wang, H. M. Wang, and X. G. Xia, "Hybrid opportunistic relay-  
942 ing and jamming with power allocation for secure cooperative net-  
943 works," *IEEE Trans. Wireless Commun.*, vol. 14, no. 2, pp. 589–605,  
944 Feb. 2015.

945 [16] H. Deng, H. M. Wang, W. Guo, and W. Wang, "Secrecy transmission with  
946 a helper: To relay or to jam," *IEEE Trans. Inf. Forensic Security*, vol. 10,  
947 no. 2, pp. 293–307, Feb. 2015.

948 [17] B. He, X. Zhou, and T. D. Abhayapala, "Wireless physical layer secu-  
949 rity with imperfect channel state information: A survey," *ZTE Commun.*,  
950 vol. 11, no. 3, pp. 11–19, Sep. 2013.

951 [18] A. Mukherjee and A. L. Swindlehurst, "Robust beamforming for security  
952 in MIMO wiretap channels with imperfect CSI," *IEEE Trans. Signal  
953 Process.*, vol. 59, no. 1, pp. 351–361, Jan. 2011.

954 [19] J. Zhang and M. C. Gursoy, "Relay beamforming strategies for physical-  
955 layer security," in *Proc. CISS*, Princeton, NJ, USA, Mar. 2010, pp. 1–6.

956 [20] M. Bloch, J. Barros, M. R. D. Rodrigues, and S. W. McLaughlin, "Wire-  
957 less information-theoretic security," *IEEE Trans. Inf. Theory*, vol. 54,  
958 no. 6, pp. 2515–2534, Jun. 2008.

959 [21] X. Zhou, M. R. McKay, B. Maham, and A. Hjørungnes, "Rethink-  
960 ing the secrecy outage formulation: A secure transmission design  
961 perspective," *IEEE Commun. Lett.*, vol. 15, no. 3, pp. 302–304,  
962 Mar. 2011.

963 [22] J. Hu, Y. Cai, N. Yang, and W. Yang, "A new secure transmission scheme  
964 with outdated antenna selection," *IEEE Trans. Inf. Forensics Security*,  
965 to be published.

966 [23] J. Hu, W. Yang, N. Yang, X. Zhou, and Y. Cai, "On-off-based secure trans-  
967 mission design with outdated channel state information," *IEEE Trans.  
968 Veh. Technol.*, to be published.

969 [24] N. E. Wu and H. J. Li, "Effect of feedback delay on secure cooperative  
970 networks with joint relay and jammer selection," *IEEE Wireless Commun.  
971 Lett.*, vol. 2, no. 4, pp. 415–418, Aug. 2013.

972 [25] X. Guan, Y. Cai, and Y. Yang, "Secure transmission design and perfor-  
973 mance analysis for cooperation exploring outdated CSI," *IEEE Commun.  
974 Lett.*, vol. 18, no. 9, pp. 1637–1640, Sep. 2014.

[26] L. Wang, S. Xu, W. Yang, W. Yang, and Y. Cai, "Security performance  
975 of multiple antennas multiple relaying networks with outdated relay  
976 selection," in *Proc. WCSP*, Hefei, China, Oct. 2014, pp. 1–6. 977

[27] J. Huang and A. L. Swindlehurst, "Buffer-aided relaying for two-hop  
978 secure communication," *IEEE Trans. Wireless Commun.*, vol. 14, no. 1,  
979 pp. 152–164, Jan. 2015.

[28] S. I. Kim, I. M. Kim, and J. Heo, "Secure transmission for multiuser relay  
981 networks," *IEEE Trans. Wireless Commun.*, vol. 14, no. 7, pp. 3724–3737,  
982 Jul. 2015.

[29] Y. Ma, D. Zhang, A. Leith, and Z. Wang, "Error performance of transmit  
984 beamforming with delayed and limited feedback," *IEEE Trans. Wireless  
985 Commun.*, vol. 8, no. 3, pp. 1164–1170, Mar. 2009. 986

[30] Z. Rezki, A. Khisti, and M. S. Alouini, "Ergodic secret message capac-  
987 ity of the wirechannel with finite-rate feedback," *IEEE Trans. Wireless  
988 Commun.*, vol. 13, no. 6, pp. 3364–3379, Jun. 2014. 989

[31] X. Tang, R. Liu, P. Spasojevic, and H. V. Poor, "On the throughput of  
990 secure hybrid-ARQ protocols for Gaussian block-fading channels," *IEEE  
991 Trans. Inf. Theory*, vol. 55, no. 4, pp. 1575–1591, Apr. 2009. 992

[32] H. A. Suraweera, M. Soysa, C. Tellambura, and H. K. Garg, "Performance  
993 analysis of partial relay selection with feedback delay," *IEEE Signal  
994 Process. Lett.*, vol. 17, no. 6, pp. 531–534, Jun. 2010. 995

[33] I. S. Gradshteyn and I. M. Ryzhik, *Table of Integrals, Series and Products*,  
996 6th ed. San Diego, CA, USA: Academic, 2000. 997

[34] S. Kim and J. Heo, "Outage probability of interference-limited amplify-  
998 and-forward relaying with partial relay selection," in *Proc. IEEE VTC*,  
999 Yokohama, Japan, May 2011, pp. 1–5. 1000



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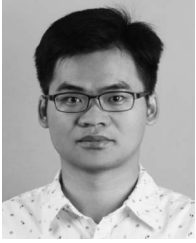
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## AUTHOR QUERIES

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AQ1 = RV was expanded as “random variable.” Please check if appropriate. Otherwise, please make the necessary changes.

AQ2 = Equations (29) and (30) are missing in the document. Please check.

AQ3 = Please provide publication update in Ref [22].

AQ4 = Please provide publication update in Ref [23].

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# Joint Relay and Jammer Selection Improves the Physical Layer Security in the Face of CSI Feedback Delays

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**Abstract**—We enhance the physical layer security (PLS) of amplify-and-forward (AF) relaying networks with the aid of joint relay and jammer selection (JRJS), despite the deleterious effect of channel state information (CSI) feedback delays. Furthermore, we conceive a new outage-based characterization approach for the JRJS scheme. The traditional best relay selection (TBRS) is also considered as a benchmark. We first derive closed-form expressions of both the connection outage probability (COP) and the secrecy outage probability (SOP) for both the TBRS and JRJS schemes. Then, a reliable and secure connection probability (RSCP) is defined and analyzed for characterizing the effect of the correlation between the COP and the SOP introduced by the corporate source–relay link. The reliability–security ratio (RSR) is introduced for characterizing the relationship between the reliability and the security through asymptotic analysis. Moreover, the concept of effective secrecy throughput is defined as the product of the secrecy rate and of the RSCP for the sake of characterizing the overall efficiency of the system, as determined by the transmit SNR, the secrecy codeword rate, and the power sharing ratio between the relay and the jammer. The impact of the direct source–eavesdropper link and additional performance comparisons with respect to other related selection schemes are also included. Our numerical results show that the JRJS scheme outperforms the TBRS method both in terms of the RSCP and in terms of its effective secrecy throughput, but it is more sensitive to the feedback delays. Increasing the transmit signal-to-noise ratio (SNR) will not always improve the overall throughput. Moreover, the RSR results demonstrate that, upon reducing the CSI feedback delays, the reliability improves more substantially than the security degrades, implying an overall improvement in terms of the security–reliability tradeoff. Additionally, the secrecy throughput loss due to the second-hop feedback delay is more pronounced than that due to the first-hop one.

**Index Terms**—Effective secrecy throughput, feedback delay, physical layer security (PLS), relay and jammer selection, reliability and security.

## I. INTRODUCTION

WIRELESS communications systems are particularly vulnerable to security attacks because of the inherent openness of the transmission medium. Traditionally, the information privacy of wireless networks has been focused on the higher layers of the protocol stack employing cryptographically secure schemes. However, these methods typically assume a limited computing power for the eavesdroppers and exhibit inherent vulnerabilities in terms of the inevitable secret key distribution and management [1]. In recent years, physical layer security (PLS) has emerged as a promising technique of improving the confidentiality wireless communications, which exploits the time-varying properties of fading channels, instead of relying on conventional cryptosystems. The pivotal idea of PLS solutions is to exploit the dynamically fluctuating random nature of radio channels for maximizing the uncertainty concerning the source messages at the eavesdropper [2], [3].

To achieve this target, several PLS-enhancement approaches have been proposed in the literature, including secrecy-enhancing channel coding [4], secure on–off transmission designs [5], secrecy-improving beamforming (BF)/precoding, and artificial-noise-aided techniques relying on multiple antennas [6], as well as secure relay-assisted transmission techniques [7]. Specifically, apart from improving the reliability and coverage of wireless transmissions, user cooperation also has a great potential in terms of enhancing the wireless security against eavesdropping attacks. There has been a growing interest in improving the security of cooperative networks at the physical layer [8]–[14]. To explore the spatial diversity potential of the relaying networks and to boost the secrecy capacity (the difference between the channel capacity of the legitimate main link and that of the eavesdropping link), most of the existing work has been focused on secrecy-enhancing BF [8], [9], as well as on intelligent relay node/jammer node (RN/JN) selection, etc. Notably, given the availability of multiple relays, appropriately designed RN/JN selection is capable of achieving a significant security improvement for cooperative networks, which is emerging as a promising research topic. In particular, Zou *et al.* investigated both amplify-and-forward (AF)- and decode-and-forward (DF)-based optimal relay selection conceived for

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83 enhancing the PLS in cooperative wireless networks [10], [11],  
 84 where the global channel state information (CSI) of both the  
 85 main link and the eavesdropping link was assumed to be avail-  
 86 able. Similarly, jamming techniques, which impose artificial  
 87 interference on the eavesdropper, have also attracted substantial  
 88 attention [12]–[14]. More specifically, several sophisticated  
 89 joint relay and jammer selection (JRJS) schemes were proposed  
 90 in [12], where the beneficially selected relay increases the reli-  
 91 ability of the main link, whereas the carefully selected jammer  
 92 imposes interference on the eavesdropper and simultaneously  
 93 protects the legitimate destination from interference. In [13]  
 94 and [14], cooperative jamming has been studied in the context  
 95 of bidirectional scenarios, and efficient RN/JN selection criteria  
 96 have been developed for achieving improved secrecy rates with  
 97 the aid of multiple relays. Furthermore, more effective relaying  
 98 and jamming schemes, when taking the information leakage  
 99 of the source–eavesdropper link into consideration, have been  
 100 presented lately in [15] and [16].

101 Nevertheless, an idealized assumption of the previously re-  
 102 ported research on PLS is the availability of perfect channel  
 103 state information (CSI), which is regarded as a stumbling block  
 104 in the way of invoking practical secrecy-enhancing Wyner  
 105 coding, on–off design, BF/precoding, and RN/JN selection.  
 106 However, this idealized simplifying assumption is not realistic,  
 107 since practical channel estimation imposes CSI imperfections,  
 108 which are aggravated by the feedback delay, limited-rate feed-  
 109 back, and channel estimation errors (CEEs) [17]. Generally, the  
 110 related research has been focused on the issues of robust secure  
 111 BF design from an average secrecy-rate-based optimization  
 112 perspective for point-to-point multi-antenna aided channels and  
 113 relay channels [18], [19] supporting delay-tolerant systems.  
 114 For systems imposing stringent delay constraints, particularly  
 115 in imperfect CSI scenarios, perfect secrecy cannot always be  
 116 achieved. Hence, the secrecy-outage-based characterization of  
 117 systems is more appropriate, which provides a probabilistic  
 118 performance measure of secure communication. The concept  
 119 of secrecy outage was adopted in [20] for characterizing the  
 120 probability of having both reliable and secure transmission,  
 121 which, however, is inapplicable for the imperfect CSI case and  
 122 fails to distinguish a connection outage from the secrecy outage.  
 123 In [21], an alternative secrecy outage formulation is proposed  
 124 for characterizing the attainable security level and provided  
 125 a general framework for designing transmission schemes that  
 126 meet specific target security requirements. To quantify both the  
 127 reliability and security performance at both the legitimate and  
 128 eavesdropper nodes separately, two types of outages, namely,  
 129 the connection outage probability (COP) and the secrecy outage  
 130 probability (SOP) are introduced. Then, considering the impact  
 131 of time delay caused by the antenna selection process at the  
 132 legitimate receiver, Hu *et al.* [22] proposed a new secure  
 133 transmission scheme in the multi-input multi-output multieaves-  
 134 dropper wiretap channel. Much recently, considering the out-  
 135 dated CSI from the legitimate receiver, a new secure on–off  
 136 transmission scheme was proposed for enhancing the secrecy  
 137 throughput in [23].

138 Moreover, prior studies of the outage-based secure trans-  
 139 mission design are limited to single-antenna-assisted single-  
 140 hop systems and have not been considered for cooperative

relaying systems. Hence, the issues of secure transmissions 141  
 over cooperative relaying channels expressed in terms of the 142  
 SOP, COP, and secrecy throughput constitute an open problem. 143  
 On the other hand, apart from CEE, the CSI feedback delay 144  
 results in critical challenges for the PLS of cooperative relaying 145  
 systems, particularly when considering the specifics of RN/JN 146  
 selection. In [15], the effects of outdated CSI knowledge con- 147  
 cerning the legitimate links on the ergodic secrecy rate achieved 148  
 by the proposed secure transmission strategy in the context 149  
 of DF relaying is investigated. The impact of CSI feedback 150  
 delay on the secure relay and jammer selection conceived for 151  
 DF relaying was investigated in [24], albeit only in terms 152  
 of the SOP. In our previous study [25], we considered the 153  
 secure transmission design and the secrecy performance of an 154  
 opportunistic DF system relying on outdated CSI, where only a 155  
 single relay is invoked. Additionally, during the revision of this 156  
 work, we investigated the security performance for outdated AF 157  
 relay selection in [26]. Therefore, in this treatise, we extend 158  
 our investigations to the PLS of multiple AF relaying assisted 159  
 networks relying on RN/JN selection. 160

Explicitly, we focus our attention on the outage-based char- 161  
 acterization of secure transmissions in cooperative relay-aided 162  
 networks relying on realistic CSI feedback delay. To exploit the 163  
 multirelay induced diversity gain and the associated jamming 164  
 capabilities, joint AF relay node and jammer node selection 165  
 is employed by the relay–destination link. We assume that, in 166  
 line with the practical reality, the instantaneous eavesdropper’s 167  
 CSI is unavailable at the legitimate transmitter and that the 168  
 RN/JN selections are performed based on the outdated CSI of 169  
 the main links. Two types of cooperative strategies are invoked 170  
 by our cooperative network operating under secrecy constraints, 171  
 namely, the traditional best relay selection (TBRS) strategy and 172  
 the JRJS strategy. Specifically, the main contributions of this 173  
 paper can be summarized as follows. 174

- We develop an outage-based characterization for quan- 175  
 tifying both the reliability and security performance of 176  
 a two-hop AF relaying system. Specifically, in contrast 177  
 to [21] and [22], we propose the novel definition of 178  
 the reliable and secure connection probability (RSCP). 179  
 Explicitly, closed-form expressions of the COP, the SOP, 180  
 and the RSCP are derived for both the TBRS and for our 181  
 JRJS strategies. Numerical results demonstrate that the 182  
 JRJS scheme outperforms the TBRS scheme in terms of 183  
 its RSCP. 184
- We also introduce the reliability–security ratio (RSR) 185  
 for characterizing their direct relationship by a single 186  
 parameter through the asymptotic analysis of the COP and 187  
 the SOP in the high-SNR regime. We derive the RSR for 188  
 both the TBRS and JRJS strategies for investigating the 189  
 effect of secrecy codeword rate setting, as well as that 190  
 of the feedback delay and that of the power sharing ratio 191  
 between the relay and the jammer on the RSR. 192
- We then modify the definition of effective secrecy 193  
 throughput by multiplying the secrecy rate with the RSCP, 194  
 which results in an optimization problem of the trans- 195  
 mit signal-to-noise ratio (SNR), secrecy codeword rate, 196  
 and power sharing between the relay and the jammer. 197  
 198

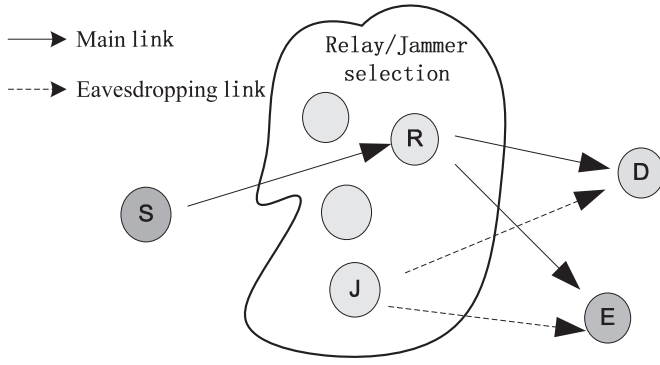


Fig. 1. Cooperative relaying network assisted by multiple relays in the presence of an eavesdropper.

199 It is shown that, compared with the TBRS strategy,  
 200 JRJS achieves a significantly higher effective secrecy  
 201 throughput, and the corresponding throughput loss is  
 202 more sensitive to feedback delays. The impact of the direct  
 203 source–eavesdropper link and additional throughput  
 204 performance comparisons with respect to other related  
 205 selection schemes are further discussed.

206 The remainder of this paper is organized as follows.  
 207 Section II introduces our system model and describes both  
 208 the TBRS and our JRJS strategies. In Sections III and IV,  
 209 we present the mathematical framework of our performance  
 210 analysis both for the TBRS strategy and for the JRJS strategy,  
 211 respectively, including the COP, the SOP, the RSCP, the RSR,  
 212 and the effective secrecy throughput. Our numerical results  
 213 and discussions are provided in Section V. Finally, Section VI  
 214 presents our concluding remarks.

## 215 II. SYSTEM MODEL

### 216 A. System Description

217 Consider a cooperative relaying network consisting of a  
 218 source  $S$ , a destination  $D$ ,  $K_r$  relays  $R_k$ ,  $k = 1, \dots, K_r$ , and  
 219 an eavesdropper  $E$ , as shown in Fig. 1, where all nodes are  
 220 equipped with a single transmit antenna (TA), except for the  
 221 source, which has  $N_t$  TAs. The cooperative relay architecture  
 222 in Fig. 1 is generally applicable to diverse practical wireless  
 223 systems in the presence of an eavesdropper, including the  
 224 family of wireless sensor networks (WSNs), mobile ad hoc  
 225 networks (MANETs), and the long-term evolution advanced  
 226 cellular systems [11].

227 To exploit the diversity potential of multiple relay nodes over  
 228 independently fading channels, AF relay/jammer selection is  
 229 employed. All relays operate in the half-duplex AF mode, and  
 230 data transmission is performed in two phases. More particu-  
 231 larly, during the broadcast phase, the source node transmits its  
 232 signal to a selected relay with the aid of BF, which is invoked  
 233 for forwarding the signal received from  $S$  to  $D$ . An inherent  
 234 assumption is that the transmit BF weights are based on the  
 235 CSI estimates quantified and fed back by the selected relay.  
 236 During the cooperative phase, a pair of appropriately selected  
 237 relays transmit toward  $D$  and  $E$ , respectively. A conventional

relay (denoted by  $R^*$ ) forwards the source’s message to the  
 238 destination. Another relay (denoted by  $J^*$ ) operates in the  
 239 “jammer mode” and imposes intentional interference upon  $E$  in  
 240 order to confuse it. However,  $D$  is unable to mitigate the artificial  
 241 interference emanating from the jammer node  $J^*$  due to its  
 242 critical secrecy constraints [12]. It should be noted that both  
 243 the process of RN/JN selection and the feedback of the transmit  
 244 BF weights from  $R^*$  to  $S$  may impose a time lag between the  
 245 data transmission and the channel estimation. These time delays  
 246 are denoted by  $T_{d_{SR}}$  and  $T_{d_{RD}}$ , respectively. Furthermore, we  
 247 assume that the BF and RN/JN selection process is based  
 248 on the perfectly estimated but outdated CSI. We employ the  
 249 first-order autoregressive outdated CSI model of [20], while  
 250 relying on the correlation coefficients of  $\rho_{SR} = J_0(2\pi f_d T_{d_{SR}})$   
 251 and  $\rho_{RD} = J_0(2\pi f_d T_{d_{RD}})$  for the two hops, where  $J_0(\cdot)$  is  
 252 the zero-order Bessel function of the first kind, and  $f_d$  is the  
 253 Doppler frequency.  
 254

A slow flat block Rayleigh fading environment is assumed,  
 255 where the channel remains static for the coherence interval (one  
 256 slot) and changes independently in different coherence inter-  
 257 vals, as denoted by  $h_{i,j} \sim \mathcal{CN}(0, \sigma_{i,j}^2)$ ,  $i, j \in \{S, R, J, D, E\}$ .  
 258 The direct communication links are assumed to be unavailable  
 259 due to the presence of obstructions between  $S$  and  $D$ , as well  
 260 as the eavesdropper.<sup>1</sup> This assumption follows the rationale of  
 261 [12] and has been routinely exploited in previous literature (see  
 262 [27] and [28] and the references therein), where the source  
 263 and relays belong to the same cluster, whereas the destination  
 264 and the eavesdropper are located in another. More specifically,  
 265 this assumption is particularly valid in networks with broadcast  
 266 and unicast transmission, where each terminal is a legitimate  
 267 receiver for one signal and acts as an eavesdropper for some  
 268 other signal. Therefore, the security concerns are only related  
 269 to the cooperative relay-aided channel. Furthermore, additive  
 270 white Gaussian noise (AWGN) is assumed with zero mean  
 271 and unit variance  $N_0$ . Let  $P_i$  be the transmit power of node  
 272  $i$ , and the instantaneous SNR of the  $i \rightarrow j$  link is given by  
 273  $\gamma_{i,j} = P_i |h_{i,j}|^2 / N_0$ .  
 274

We employ the constant-rate Wyner coding scheme for con-  
 275 structing wiretap codes of [2] to meet the PLS requirements  
 276 due to the fact that the accurate global CSI is not available.  
 277 Let  $\mathbb{C}(R_0, R_s, N)$  denote the set of all possible Wyner codes  
 278 of length  $N$ , where  $R_0$  is the codeword transmission rate, and  
 279  $R_s$  is the confidential information rate ( $R_0 > R_s$ ). The positive  
 280 rate difference  $R_e = R_0 - R_s$  is the cost of providing secrecy  
 281 against the eavesdropper. A confidential message is encoded  
 282 into a codeword at  $S$  and then transmitted to  $D$ .  
 283

### 284 B. Secure Transmission

In the broadcast phase,  $S$  transmits its BF signal  $s(t)$  to the  
 285 selected relay  $R^*$ , where the relay selection is performed  
 286 before data transmission commences, and the selection cri-  
 287 terion will be detailed later in the context of the cooper-  
 288 ative phase. The transmit BF vector  $\mathbf{w}(t|T_d)$  is calculated  
 289 using the perfectly estimated but outdated CSI given by  
 290

<sup>1</sup>The case when the  $S \rightarrow E$  link is introduced will be investigated separately in Section VI.

291  $\mathbf{w}(t|T_{d_{SR}}) = \mathbf{h}_{SR^*}^H(t - T_d)/|\mathbf{h}_{SR^*}(t - T_{d_{SR}})|$  [29], where we  
 292 have  $\mathbf{h}_{SR^*}(t) = [h_{SR^*,1}(t), \dots, h_{SR^*,N_t}(t)]^T$ , and the signal  
 293 received by the relay  $R^*$  can be written as

$$y_{R^*}(t) = \sqrt{P_s} \mathbf{w}(t|T_d) \mathbf{h}_{SR^*}(t) s(t) + n_{SR^*}(t) \quad (1)$$

294 where  $n_{SR^*}(t)$  is the AWGN at the relay. Then, we can  
 295 define the received SNR at the relay node as  $\gamma_{SR} =$   
 296  $P_S |\mathbf{w}(t|T_{d_{SR}}) \mathbf{h}_{SR^*}(t)|^2 / N_0$ .

297 In the cooperative phase, we consider two RN/JN selection  
 298 schemes performed by  $D$ : relay selection without jamming  
 299 and JRJS.

300 1) *Traditional Best Relay Selection*: The first category of so-  
 301 lutions does not involve a jamming process, and therefore, only  
 302 a conventional relay accesses the channel during the second  
 303 phase of the protocol. The relay selection process is performed  
 304 based on the highest instantaneous SNR of the second hop,  
 305 which is formulated as

$$R^* = \arg \max_{R_k \in \mathcal{R}} \left\{ \frac{\tilde{\gamma}_{R_k D}}{\mathbb{E}[\gamma_{R_k E}]} = \frac{P_R |\tilde{h}_{R_k D}(t - T_d)|^2}{N_0 \mathbb{E}[\gamma_{R_k E}]} \right\} \quad (2)$$

306 where  $\tilde{\gamma}_{R_k D}$  is the instantaneous SNR in the relay selection  
 307 process, and  $\mathbb{E}[\gamma_{R_k E}]$  denotes the average SNR at  $E$ . We can  
 308 model  $\gamma_{R_k D}$  and  $\tilde{\gamma}_{R_k D}$  as two gamma distributed random  
 309 variables having the correlation factor of  $\rho_{RD}^2$ .

310 During the second phase, the received signal  $y_{R^*}(t)$   
 311 is multiplied by a time-variant AF-relay gain  $G$  and  
 312 retransmitted to  $D$ , where we have  $G =$   
 313  $\sqrt{P_R / (P_S |\mathbf{w}_{\text{opt}}(t|T_{d_{SR}}) \mathbf{h}_{SR^*}(t)|^2 + N_0)}$ . After further math-  
 314 ematical manipulations, the mutual information (MI) between  
 315  $S$  and  $D$ , as well as the eavesdropper, can be written as

$$I_D^{\text{TBRS}} = \frac{1}{2} \log(1 + \gamma_D^{\text{TBRS}}) = \frac{1}{2} \log \left( 1 + \frac{\gamma_{SR} \gamma_{R^* D}}{\gamma_{SR} + \gamma_{R^* D} + 1} \right) \quad (3)$$

$$I_E^{\text{TBRS}} = \frac{1}{2} \log(1 + \gamma_E^{\text{TBRS}}) = \frac{1}{2} \log \left( 1 + \frac{\gamma_{SR} \gamma_{R^* E}}{\gamma_{SR} + \gamma_{R^* E} + 1} \right). \quad (4)$$

316 2) *Joint Relay and Jammer Selection*: Similarly, consider-  
 317 ing the unavailability of the instantaneous CSI regarding the  
 318 eavesdropper, we adopt a suboptimal RN/JN selection metric  
 319 conditioned on the outdated CSI as

$$R^* = \arg \max_{R_k \in \mathcal{R}} \left\{ \frac{\tilde{\gamma}_{R_k D}}{\mathbb{E}[\gamma_{R_k E}]} \right\} \quad (5)$$

$$J^* = \arg \min_{R_k \in \mathcal{R} - R^*} \left\{ \frac{\tilde{\gamma}_{R_k D}}{\mathbb{E}[\gamma_{R_k E}]} \right\}$$

320 where  $J^*$  is selected for minimizing the interference imposed  
 321 on  $D$ .

322 It should be noted that, to have the same transmit power as  
 323 that of the TBRS case, we assume that  $P_{R^*} + P_{J^*} = P_R$  for  
 324 our JRJS strategy and introduce  $\lambda = P_{R^*} / (P_{R^*} + P_{J^*})$  as the

ratio of the relay's transmit power to the total power required  
 by the active relay and jammer. 326

In the cooperative phase,  $R^*$  will also amplify the received  
 signal  $y_{R^*}(t)$  by  $G$  and forward it to  $D$ . At the same time, the  
 jammer  $J^*$  will generate intentional interference to confuse  $E$ ,  
 which will also cause interference at  $D$ . Consequently, the MI  
 between the terminals is given by 331

$$I_D^{\text{JRJS}} = \frac{1}{2} \log(1 + \gamma_D^{\text{JRJS}}) = \frac{1}{2} \log \left( 1 + \frac{\gamma_{SR} \frac{\gamma_{R^* D}}{\gamma_{J^* D} + 1}}{\gamma_{SR} + \frac{\gamma_{R^* D}}{\gamma_{J^* D} + 1} + 1} \right) \quad (6)$$

$$I_E^{\text{JRJS}} = \frac{1}{2} \log(1 + \gamma_E^{\text{JRJS}}) = \frac{1}{2} \log \left( 1 + \frac{\gamma_{SR} \frac{\gamma_{RE}}{\gamma_{JE} + 1}}{\gamma_{SR} + \frac{\gamma_{RE}}{\gamma_{JE} + 1} + 1} \right). \quad (7)$$

*Remark 1*: Generally, the optimal RN/JN selection scheme  
 should take into account the global SNR knowledge set  
 $\{\gamma_{SR}, \gamma_{RD}, \gamma_{RE}\}$ . However, given the potentially excessive  
 implementational complexity overhead of the optimal selection  
 schemes and the unavailability of the global CSI, we employ  
 suboptimal selection schemes as in [12].<sup>2</sup> Furthermore, it is  
 commonly assumed that the average SNR of the eavesdropper  
 is available at the transmitter, which seems, somehow, not  
 reasonable. However, as stated in most of the literature, such as  
 [12]–[22], [24]–[28], and [30], provided that the eavesdropper  
 belongs to the network, which is also the case in our paper,  
 the related assumption might still be deemed reasonable. Addi-  
 tionally, as in [8], [11], [12], and [24], for mathematical conve-  
 nience, we assume that the relaying channels are independent  
 and identically distributed and that we have  $\mathbb{E}[\gamma_{SR_k}] = \bar{\gamma}_{SR}$ ,  
 $\mathbb{E}[\gamma_{R_k D}] = \bar{\gamma}_{RD}$ , and  $\mathbb{E}[\gamma_{R_k E}] = \bar{\gamma}_{RE}$ . The distances between  
 the relays are assumed to be much smaller than the distances  
 between relays and source/destination/eavesdropper; hence, the  
 corresponding path losses among the different relays are ap-  
 proximately the same. This assumption is reasonable both for  
 WSNs and for MANETs associated with a symmetric clustered  
 relay configuration, and it may be also satisfied as valid by  
 classic cellular systems in a statistical sense [11]. 354

### III. SECURE TRANSMISSION WITHOUT JAMMING 355

Here, we endeavor to characterize both the reliability and  
 security performance comprehensively of the TBRS scheme.  
 We first derive closed-form expressions for both the COP and  
 the SOP. Then, the RSR is introduced through the asymptotic  
 analysis of the COP and the SOP. Furthermore, we propose  
 the novel definition of the RSCP and the effective secrecy  
 throughput. 362

<sup>2</sup>To further alleviate the cooperation-related overhead, the selection criterion is based on the  $R \rightarrow D$  link, since the second hop plays a dominant role in determining the received SNR, because the first hop corresponds to a multiple-input-single-output channel with the aid of multiple antennas, and hence, it is more likely to be better than the second hop. The optimal selection based on both hops is beyond the scope of this work.

## 363 A. COP and SOP

364 When the perfect instantaneous CSI of the eavesdropper's  
 365 channel and even the legitimate users' channel is unavailable,  
 366 alternative definitions of the outage probability may be adopted  
 367 for the statistical characterization of the attainable secrecy  
 368 performance, particularly for delay-limited applications. Based  
 369 on [31, Def. 2], perfect secrecy cannot be achieved, when we  
 370 have  $R_e < I_E$ , where  $I_E$  denotes the MI between the source  
 371 and the eavesdropper. Encountering this event is termed as a  
 372 secrecy outage. Furthermore, the destination is unable to flaw-  
 373 lessly decode the received codewords when  $R_0 > I_D$ , which is  
 374 termed as a connection outage. The grade of reliability and the  
 375 grade of security maintained by a transmission scheme may be  
 376 then quantified by the COP and the SOP, respectively.

377 We continue by presenting our preliminary results versus the  
 378 point-to-point SNRs. Let us denote the cumulative distribution  
 379 function (CDF) and the probability density function (PDF) of a  
 380 random variable  $X$  by  $F_X(x)$  and  $f_X(x)$ , respectively. On one  
 381 hand, the PDF of  $\gamma_{SR}$  using [29, eq. (15)] is given by

$$f_{\gamma_{SR}}(x) = \sum_{n=0}^{N_t-1} \binom{N_t-1}{n} \frac{\rho_{SR}^{2(N_t-1-n)} (\bar{\gamma}_{SR} (1 - \rho_{SR}^2))^n}{\bar{\gamma}_{SR}^{N_t} (N_t - 1 - n)!} \times x^{N_t-1-n} e^{-\frac{x}{\bar{\gamma}_{SR}}} \quad (8)$$

382 whereas its CDF is given by

$$F_{\gamma_{SR}}(x) = 1 - \sum_{n=0}^{N_t-1} \sum_{m=0}^{N_t-1-n} \binom{N_t-1}{n} \times \frac{\rho_{SR}^{2(N_t-1-n)} (1 - \rho_{SR}^2)^n}{m! \bar{\gamma}_{SR}^m} x^m e^{-\frac{x}{\bar{\gamma}_{SR}}}. \quad (9)$$

383 On the other hand, for the instantaneous SNR of the  $R \rightarrow$   
 384  $D$  hop, according to the principles of concomitants or induced  
 385 order statistics, the CDF of  $\gamma_{R^*D}$  can be derived as in [32]

$$F_{\gamma_{R^*D}}(y) = K_r \sum_{k=0}^{K_r-1} (-1)^k \binom{K_r-1}{k} \frac{1 - e^{-\frac{-(k+1)y}{k(1-\rho_{RD}^2)+1} \bar{\gamma}_{RD}}}{k+1}. \quad (10)$$

386 Thus, the COP of the TBRS strategy is given by

$$P_{co}^{TBRS}(R_0) = \Pr [I_D^{TBRS} < R_0] = F_{\gamma_D^{TBRS}}(\gamma_{th}^D) \quad (11)$$

387 where we have  $\gamma_{th}^D = 2^{2R_0} - 1$ , and the CDF of  $\gamma_D^{TBRS}$  can be  
 388 calculated as

$$F_{\gamma_D^{TBRS}}(x) = 1 - \int_0^\infty \left[ 1 - F_{\gamma_{R^*D}}\left(\frac{xz + x(x+1)}{z}\right) \right] f_{\gamma_{SR^*}}(z+x) dz. \quad (12)$$

389 Consequently, by substituting (8) and (10) into (12) and using  
 390 [33, eq. (3.471.9)], we arrive at a closed-form expression for

$F_{\gamma_D^{TBRS}}(x)$  as

391

$$F_{\gamma_D^{TBRS}}(x) = 1 - 2 \sum_{n=0}^{N_t-1} \sum_{k=0}^{K_r-1} \sum_{m=0}^{N_t-1-n} (-1)^k K_r \binom{N_t-1}{n} \times \binom{K_r-1}{k} \binom{N_t-1-n}{m} \times \frac{\rho_{SR}^{2(N_t-1-n)} (1 - \rho_{SR}^2)^n x^{N_t-1-n-m}}{(N_t-1-n)! (k+1) \bar{\gamma}_{SR}^{N_t-n}} \times \left[ \frac{\bar{\gamma}_{SR} x(x+1)}{\omega_k \bar{\gamma}_{RD}} \right]^{\frac{m+1}{2}} \times e^{-\left(\frac{\bar{\gamma}_{SR} + \omega_k \bar{\gamma}_{RD}}{\omega_k \bar{\gamma}_{SR} \bar{\gamma}_{RD}}\right) x} K_{m+1} \left( 2 \sqrt{\frac{x(x+1)}{\omega_k \bar{\gamma}_{SR} \bar{\gamma}_{RD}}} \right) \quad (13)$$

where we have  $\omega_k = (k(1 - \rho_{RD}^2) + 1)/(k+1)$ . Then, by  
 substituting  $x = \gamma_{th}^D$  into (13), we obtain  $P_{co}^{TBRS}$ .

393 Furthermore, the SOP of the TBRS strategy may be expressed as 394

$$P_{so}^{TBRS}(R_0, R_s) = \Pr [I_E^{TBRS} > R_0 - R_s] = 1 - F_{\gamma_E^{TBRS}}(\gamma_{th}^E) \quad (14)$$

where we have  $\gamma_{th}^E = 2^{2(R_0 - R_s)} - 1$ . Similarly, we may calcu-  
 late the CDF of  $\gamma_E^{TBRS}$  in (14) as 396

$$F_{\gamma_E^{TBRS}}(x) = 1 - 2 \sum_{n=0}^{N_t-1} \sum_{m=0}^{N_t-1-n} \binom{N_t-1}{n} \binom{N_t-1-n}{m} \times \frac{\rho_{SR}^{2(N_t-1-n)} (1 - \rho_{SR}^2)^n x^{N_t-1-n-m}}{(N_t-1-n)! \bar{\gamma}_{SR}^{N_t-n}} \times \left[ \frac{\bar{\gamma}_{SR} x(x+1)}{\bar{\gamma}_{RE}} \right]^{\frac{m+1}{2}} \times e^{-\left(\frac{\bar{\gamma}_{SR} + \bar{\gamma}_{RE}}{\bar{\gamma}_{SR} \bar{\gamma}_{RE}}\right) x} K_{m+1} \left( 2 \sqrt{\frac{x(x+1)}{\bar{\gamma}_{SR} \bar{\gamma}_{RE}}} \right). \quad (15)$$

Then, by substituting  $x = \gamma_{th}^E$  into (15), we can derive  $P_{so}^{TBRS}$ .

397 The COP and the SOP in (11) and (14) characterize the at-  
 398 tainable reliability and security performance, respectively, and  
 399 can be regarded as the detailed requirements of accurate system  
 400 design. From the definition of COP and SOP, it is clear that  
 401 the reliability of the main link can be improved by increasing  
 402 the transmit SNR (or decreasing its data rate) to reduce the  
 403 COP, which unfortunately increases the risk of eavesdropping.  
 404 Thus, a tradeoff between reliability and security may be struck,  
 405 despite the fact that closed-form expressions cannot be obtained  
 406 as in [11]. Furthermore, we denote the minimal reliability and  
 407 security requirements by  $\nu$  and  $\delta$ , where the feasible range of  
 408 the reliability constraint is  $0 < \nu < 1$ . Bearing in mind that  
 409 the COP is a monotonously increasing function of  $R_0$ , the  
 410 corresponding threshold of the codeword transmission rate is  
 411  $R_0^{th} = \arg\{P_{co}^{TBRS}(R_0) = \nu\}$ , which leads to a lower bound of  
 412 the SOP, when we have  $(R_0 - R_s) \rightarrow R_0^{th}$ . Thus, the feasible  
 413 range of  $\delta$  is  $P_{so}^{TBRS}(R_0^{th}, 0) < \delta < 1$ . The preceding analysis  
 414 indicates that, given a reliability constraint  $\nu$ , the lower bound  
 415 of the security constraint is determined. 416

### 417 B. Reliability–Security Ratio

418 Here, we will focus our attention on the asymptotic analysis  
419 of the COP and the SOP in the high-SNR regime. Then, inspired  
420 by [25], we introduce the concept of the RSR for characterizing  
421 the direct relationship between reliability and security.

422 *Proposition 1:* Based on the asymptotic probabilities of  $P_{co}$   
423 and  $P_{so}$  at high SNRs,<sup>3</sup> the RSR is defined as

$$P_{co}(R_0) = \Lambda [1 - P_{so}(R_0, R_s)] \quad (16)$$

424 where  $\Lambda = \lim_{\eta \rightarrow \infty} P_{co}/(1 - P_{so})$ , which represents the im-  
425 provement in COP upon decreasing the SOP. More specifically,  
426 since the reduction of the SOP/COP must be followed by an  
427 improvement of COP/SOP, a lower  $\Lambda$  implies that, when the  
428 security is reduced, the reliability is improved, and *vice versa*.  
429 Thus, for the TBRS scheme studied earlier, the RSR is derived  
430 as (17), shown at the bottom of the page.

431 *Proof:* The proof is given in Appendix B.

432 *Remark 2:* It can be seen from the preceding expression  
433 that the factor  $\Lambda$  is independent of the transmit SNR, but  
434 directly depends on the channel gains, the rate pair  $(R_0, R_s)$ ,  
435 and the number of TAs and relays. For a given  $R_s$ , reducing  
436  $R_0$  to enhance the reliability may erode the security, because  
437  $(R_0 - R_s)$  is also reduced. Conversely, increasing  $R_0$  provides  
438 more redundancy for protecting the security of the information,  
439 but simultaneously, the reliability is reduced. Hence, the RSR  
440 analysis underlines an important point of view concerning how  
441 to balance the reliability versus security tradeoff by adjusting  
442  $(R_0, R_s)$ . Furthermore, as long as a CSI feedback delay exists,  
443 the RSR has an intimate relationship with  $\rho_{SR}$  and  $\rho_{RD}$ . It is  
444 clear that the value of  $\Lambda^{\text{TBRS}}$  decreases as  $\rho_{RD}$  increases, which  
445 is due to the fact that the relay selection process only improves  
446 the reliability of the legitimate user. On the other hand, since  
447 we always have the conclusion that  $\sum_{k=0}^{K_r-1} (-1)^k \binom{K_r-1}{k} (K_r /$   
448  $(k(1 - \rho_{RD}^2) + 1)) < 1$ , when  $\sigma_{RD}^2$  and  $\sigma_{RE}^2$  are comparable,  
449  $\Lambda^{\text{TBRS}}$  will be reduced as  $\rho_{SR}$  increases. This observation  
450 implies that, although both  $P_{co}$  and  $(1 - P_{so})$  are reduced  
451 when the first-hop CSI becomes better, the improvement of

the reliability is more substantial than the security loss, as  $\rho_{SR}$  452  
increases. 453

### C. Effective Secrecy Throughput

454 It should be noted that the COP and SOP metrics ignore the 455  
correlation between these two outage events. More specifically, 456  
in contrast to the point-to-point transmission case, since the 457  
 $S \rightarrow R$  link's SNR included in the MI expressions of (3) and 458  
(4), the secrecy outage and the connection outage are definitely 459  
not independent of each other. Therefore, it might be of limited 460  
benefit in evaluating the reliability or the security separately. 461  
We note furthermore that, although another metric referred to 462  
as the secrecy throughput was introduced as the product of the 463  
successful decoding probability and of the secrecy rate [21], 464  
[22], this definition ignores the fact that a reliable transmission 465  
may be insecure, and the SOP is not taken into consideration. 466  
Hence, this metric is unable to holistically characterize the 467  
efficiency of our scheme, while capable of achieving both re- 468  
liable and secure transmission. Therefore, here, we redefine the 469  
effective secrecy throughput as the probability of a successful 470  
transmission (reliable and secure) multiplied by the secrecy 471  
rate, namely, as  $\varsigma = R_s P_{R\&S}$ , where the RSCP is defined as 472

$$P_{R\&S} = \Pr\{I_D > R_0, I_E < R_0 - R_s\}. \quad (18)$$

473 Upon substituting the expressions of  $I_D$  and  $I_E$  in (3) and (4) 474  
into (18), we can rewrite  $P_{R\&S}$  for the TBRS strategy in (19), 475  
shown at the bottom of the page.

476 Finally, using the corresponding CDFs and PDFs of (8)–(10) 477  
from our previous analysis, we can obtain  $P_{R\&S}^{\text{TBRS}}$  in (20), 478  
shown at the bottom of the next page, as well as the secrecy 479  
throughput.

480 Furthermore, considering the asymptotic result for RSCP at 481  
high SNRs in (20) by applying the approximation  $K_v(x) \approx$  482  
 $(v-1)!/2(x/2)^v$  and closing the highest terms of  $\eta$  after 483  
invoking the McLaurin series representation for the exponential 484  
function, the asymptotic effective secrecy throughput can be 485  
approximated as

486 *Remark 3:* Given the definition of COP, SOP, and the secrecy 487  
throughput result of (21), shown at the bottom of the next page, 488  
it can be shown that, for a fixed  $R_s$ , if  $R_0$  is too small, although 489

$$\Lambda^{\text{TBRS}} = \frac{\left[ (1 - \rho_{SR}^2)^{N_t-1} + \sum_{k=0}^{K_r-1} (-1)^k \binom{K_r-1}{k} \frac{K_r \sigma_{SR}^2}{[k(1 - \rho_{RD}^2) + 1] \sigma_{RD}^2} \right] (2^{2R_0} - 1)}{\left[ N_t (1 - \rho_{SR}^2)^{N_t-1} + \sigma_{SR}^2 / \sigma_{RE}^2 \right] (2^{2(R_0 - R_s)} - 1)} \quad (17)$$

$$\begin{aligned} P_{R\&S}^{\text{TBRS}} &= \Pr \left\{ \left\{ \gamma_{SR} > \gamma_{th}^D, \gamma_{R^*D} > \frac{\gamma_{th}^D \gamma_{SR} + \gamma_{SR}}{\gamma_{SR} - \gamma_{th}^D} \right\} \cap \left[ \left\{ \gamma_{SR} > \gamma_{th}^E, \gamma_{R^*E} < \frac{\gamma_{th}^E \gamma_{SR} + \gamma_{SR}}{\gamma_{SR} - \gamma_{th}^E} \right\} \cup \left\{ \gamma_{SR} < \gamma_{th}^E \right\} \right] \right\} \\ &= \Pr \left\{ \gamma_{SR} > \gamma_{th}^D, \gamma_{R^*D} > \gamma_{th}^D + \frac{\gamma_{th}^D (\gamma_{th}^D + 1)}{\gamma_{SR} - \gamma_{th}^D}, \gamma_{R^*E} < \gamma_{th}^E + \frac{\gamma_{th}^E (\gamma_{th}^E + 1)}{\gamma_{SR} - \gamma_{th}^E} \right\} \end{aligned} \quad (19)$$

<sup>3</sup>Assume equal power allocation between  $S$  and the relay, yielding  $P_S = P_R = P$ , and define  $\eta = P/N_0$  as the transmit SNR [24].

489  $P_{RS}$  may be high (i.e., close to 1), the value of  $\varsigma$  remains small.  
 490 By contrast, if  $R_0$  is too large, the value of  $P_{co}$  is close to 1,  
 491 and therefore,  $\varsigma$  will also become small. This observation is  
 492 also suitable for  $R_s$ . Thus, as pointed out in the RSR analysis,  
 493 it is elusive to improve both the reliability and the security  
 494 simultaneously, but both of them are equally crucial in terms  
 495 of the effective secrecy throughput, which depends on the rate  
 496 pair  $(R_0, R_s)$ .

497 Additionally, (21) also reveals that increasing the SNR would  
 498 drastically reduce the effective secrecy throughput. For high  
 499 transmit SNRs, a high reliability can indeed be perfectly guar-  
 500 anteed, but at the same time, the grade of the security is severely  
 501 degraded. However, the probability of a reliable and simultane-  
 502 ously secure transmission will tend toward zero. Hence, we may  
 503 conclude that there exists an optimal SNR, which achieves the  
 504 maximal secrecy throughput.

505 In conclusion, adopting the appropriate code rate pair and  
 506 transmit SNR is crucial for achieving the maximum effective  
 507 secrecy throughput, which can be formulated as

$$\begin{aligned} \max_{R_0, R_s, \eta} \quad & \varsigma(R_0, R_s) = R_s P_{R\&S}^{\text{TBRS}} \\ \text{s.t.} \quad & P_{co} \leq \nu, P_{so} \leq \delta, 0 < R_s < R_0 \end{aligned} \quad (22)$$

508 where  $\nu$  and  $\delta$  denote the system's reliability and security  
 509 requirements. Unfortunately, it is quite a challenge to find  
 510 the closed-form optimal solution to this problem due to the  
 511 complexity of the expressions. Although suboptimal solutions  
 512 can be found numerically (with the aid of gradient-based search  
 513 techniques), the secrecy throughput optimization problem and  
 514 the corresponding complexity analysis and performance com-  
 515 parisons are beyond the scope of this work.

#### 516 IV. SECURE TRANSMISSION WITH JAMMING

517 Here, we consider the extension of the aforementioned relay  
 518 selection approaches to systems additionally invoking relay-

aided jamming. JRJS is based on the outdated but perfectly  
 estimated CSI, and the details have been presented in Section II.  
 We would also like to investigate the security performance  
 from an outage-based perspective. The COP, SOP, RSCP, and  
 effective secrecy throughput will be included.

#### A. COP and SOP

It is plausible that the main differences between the JRJS and  
 TBRS schemes are determined by the instantaneous SNR of the  
 $R \rightarrow D$  hop, where, now, a jammer is included. Based on our  
 preliminary results detailed for the point-to-point SNRs in (8)  
 and (10), we now focus our attention on the statistical analysis  
 of the SNR, including  $J^*$ . As stated for the JRJS scheme in  
 Section II,  $J^*$  corresponds to the lowest  $\tilde{\gamma}_{R_k D}$  and is selected  
 from the set  $\{\mathcal{R} - R^*\}$ . Recalling that  $R^*$  is the best relay  
 of the second hop, we have  $\tilde{\gamma}_{J^* D} = \min_{R_k \in \mathcal{R} - R^*} \{\tilde{\gamma}_{R_k D}\} \triangleq$   
 $\min_{R_k \in \mathcal{R}} \{\tilde{\gamma}_{R_k D}\}$  for  $K_r > 1$ . Using the induced order statis-  
 tics, the corresponding CDF of  $\gamma_{R^* D}$  is presented in (10),  
 whereas the PDF of  $\gamma_{J^* D}$  can be formulated as

$$f_{\gamma_{J^* D}}(x) = \frac{K_r \exp\left(\frac{-K_r x}{[(K_r - 1)(1 - \rho_{RD}^2) + 1]\tilde{\gamma}_{JD}}\right)}{[(K_r - 1)(1 - \rho_{RD}^2) + 1]\tilde{\gamma}_{JD}}. \quad (23)$$

Although the relay and jammer selection processes are not  
 entirely disjoint, we may exploit the assumption that  $\gamma_{R^* D}$  and  
 $\gamma_{J^* D}$  are independent of each other, which is valid when the  
 number of relays is sufficiently high, as justified in [24]. Let us  
 define the signal-to-interference-plus-noise ratio of the second  
 hop as  $\xi_D = \gamma_{R^* D} / (\gamma_{J^* D} + 1)$ , using (10) and (23), whose  
 CDF can be formulated as

$$F_{\xi_D}(x) = 1 - K_r \sum_{k=0}^{K_r - 1} (-1)^k \binom{K_r - 1}{k} \frac{\varphi_k e^{-\frac{x}{\tilde{\gamma}_{RD} \omega_k}}}{(k + 1)(x + \varphi_k)} \quad (24)$$

where we have  $\varphi_k = \lambda K_r \omega_k / ((K_r - 1)(1 - \rho_{RD}^2) + 1)(1 - \lambda)$ .

$$\begin{aligned} P_{R\&S}^{\text{TBRS}} &= \int_{\gamma_{th}^D}^{\infty} \left[ 1 - F_{\gamma_{R^* D}} \left( \gamma_{th}^D + \frac{\gamma_{th}^D (\gamma_{th}^D + 1)}{x - \gamma_{th}^D} \right) \right] F_{\gamma_{R^* E}} \left( \gamma_{th}^E + \frac{\gamma_{th}^E (\gamma_{th}^E + 1)}{x - \gamma_{th}^E} \right) f_{\gamma_{SR^*}}(x) dx \\ &\approx 2 \sum_{n=0}^{N_t - 1} \sum_{k=0}^{K_r - 1} \sum_{m=0}^{N_t - 1 - n} (-1)^k \binom{K_r - 1}{k} \binom{N_t - 1}{n} \binom{N_t - 1 - n}{m} \frac{K_r \rho_{SR}^{2(N_t - 1 - n)} (1 - \rho_{SR}^2)^n (\gamma_{th}^D)^{N_t - 1 - n - m}}{(N_t - 1 - n)! (k + 1) \tilde{\gamma}_{SR}^{N_t - n - (m + 1)/2}} \\ &\quad \times \exp \left[ - \left( \frac{\gamma_{th}^D}{\tilde{\gamma}_{SR}} + \frac{\gamma_{th}^D}{\omega_k \tilde{\gamma}_{RD}} \right) \right] \left[ \left( \frac{\gamma_{th}^D (\gamma_{th}^D + 1)}{\omega_k \tilde{\gamma}_{RD}} \right)^{\frac{m+1}{2}} K_{m+1} \left( 2 \sqrt{\frac{\gamma_{th}^D (\gamma_{th}^D + 1)}{\omega_k \tilde{\gamma}_{SR} \tilde{\gamma}_{RD}}} \right) \right. \\ &\quad \left. - \exp \left( \frac{-\gamma_{th}^E}{\tilde{\gamma}_{RE}} \right) \left( \frac{\gamma_{th}^D (\gamma_{th}^D + 1)}{\omega_k \tilde{\gamma}_{RD}} + \frac{\gamma_{th}^E (\gamma_{th}^E + 1)}{\tilde{\gamma}_{RE} + \gamma_{th}^D - \gamma_{th}^E} \right)^{\frac{m+1}{2}} K_{m+1} \left( 2 \sqrt{\frac{\gamma_{th}^D (\gamma_{th}^D + 1)}{\omega_k \tilde{\gamma}_{SR} \tilde{\gamma}_{RD}} + \frac{\gamma_{th}^E (\gamma_{th}^E + 1)}{\tilde{\gamma}_{SR} (\tilde{\gamma}_{RE} + \gamma_{th}^D - \gamma_{th}^E)}} \right) \right] \end{aligned} \quad (20)$$

$$\zeta^{\text{TBRS}}(R_0, R_s, \eta) = R_s \left\{ 1 - \left[ \frac{N_t (1 - \rho_{SR}^2)^{N_t - 1}}{\sigma_{SR}^2} + \sum_{k=0}^{K_r - 1} \frac{K_r (-1)^k}{[k (1 - \rho_{RD}^2) + 1] \sigma_{RD}^2} \binom{K_r - 1}{k} \right] \times \frac{2^{2R_0} - 1}{\eta} \right\} \frac{2^{2(R_0 - R_s)} - 1}{\sigma_{RE}^2 \eta} \quad (21)$$



545 As far as the eavesdropper is concerned,  $\gamma_{R^*E}$  and  $\gamma_{J^*E}$   
546 are independent and exponentially distributed. Furthermore, for  
547  $\xi_E = \gamma_{R^*E}/(\gamma_{J^*E} + 1)$ , we have

$$F_{\xi_E}(x) = 1 - \frac{\phi}{x + \phi} e^{-\frac{x}{\phi}} \quad (25)$$

548 where  $\phi = \lambda/(1 - \lambda)$ . According to the definition of COP and  
549 SOP in Section III-A, we can obtain the following closed-form  
550 approximations of the COP and the SOP.<sup>4</sup>

551 *Lemma 1:* The COP and the SOP of the JRJS strategy  
552 associated with feedback delays are approximated by

$$\begin{aligned} P_{\text{co}}^{\text{JRJS}}(R_0) &\approx 1 - \sum_{n=0}^{N_t-1} \sum_{k=0}^{K_r-1} \sum_{m=0}^{N_t-1-n} \binom{N_t-1}{n} \\ &\times \binom{K_r-1}{k} \binom{N_t-1-n}{m} \\ &\times \frac{(-1)^k (K_r+1) \rho_{SR}^{2(N_t-1-n)} (1 - \rho_{SR}^2)^n}{(N_t-1-n)! (k+1) \bar{\gamma}_{SR}^{N_t-n}} \\ &\times \frac{\Gamma(m+2) \hat{\phi}_k (\gamma_{th}^D)^{N_t-n} (\gamma_{th}^D + 1)^{m+1}}{(\gamma_{th}^D + \hat{\phi}_k)^{m+2}} \\ &\times \exp \left[ -\frac{\gamma_{th}^D (\hat{\phi}_k - 1)}{\bar{\gamma}_{SR} (\gamma_{th}^D + \hat{\phi}_k)} \right] \\ &\times \Gamma \left( -m-1, \frac{\gamma_{th}^D (\gamma_{th}^D + 1)}{\bar{\gamma}_{SR} (\gamma_{th}^D + \hat{\phi}_k)} \right) \end{aligned} \quad (26)$$

553 where  $\hat{\phi}_k = K_r \lambda \omega_k \eta \sigma_{RD}^2 / [(K_r - 1)(1 - \rho_{RD}^2) + 1](1 -$   
554  $\lambda) \eta \sigma_{RD}^2 + K_r)$ , and

$$\begin{aligned} P_{\text{so}}^{\text{JRJS}}(R_0, R_s) &\approx \sum_{n=0}^{N_t-1} \sum_{m=0}^{N_t-1-n} \binom{N_t-1}{n} \\ &\times \frac{\rho_{SR}^{2(N_t-1-n)} (1 - \rho_{SR}^2)^n}{m! \bar{\gamma}_{SR}^m} \\ &\times \frac{(2\gamma_{th}^E)^m \phi}{(2\gamma_{th}^E + \phi)} \exp \left[ -\left( \frac{2\gamma_{th}^E}{\bar{\gamma}_{SR}} + \frac{2\gamma_{th}^E}{\bar{\gamma}_{RE}} \right) \right]. \end{aligned} \quad (27)$$

555 *Proof:* The proof is given in Appendix B.

556 The feasible range of the reliability constraint is similar to  
557 that of the TBRS strategy, and hence, it is omitted here.

### 558 B. Reliability–Security Ratio

559 *Lemma 2:* Recalling the definition in Section III, the RSR  
560 for the JRJS strategy may be expressed in (28), shown at the  
561 bottom of the page.

<sup>4</sup>When we have  $\lambda \rightarrow 1$ , (24) will degenerate into the TBRS case seen in (10). The performance analysis of the JRJS will be presented separately in the following, since several approximations have to be included.

562 It can be seen from the previous expression that, in contrast  
563 to the analysis of the TBRS strategy operating without jam-  
564 ming, for a fixed SNR threshold, the CDF of the second-hop  
565 SNR will converge to a nonzero limit. We also find that this  
566 limit is determined by the power sharing ratio between the  
567 relay and the jammer. Furthermore, according to the analy-  
568 sis of the TBRS strategy, for  $\eta \rightarrow \infty$ , we have  $F_{\gamma_{SR^*}}(x) \rightarrow$   
569 0. Thus, by exploiting the tight upper bound that  $\gamma_D^{\text{TBRS}} \leq$   
570  $\min\{\gamma_{SR}, \gamma_{R^*D}\}$  and  $\gamma_E^{\text{TBRS}} \leq \min\{\gamma_{SR}, \gamma_{R^*E}\}$ , we have  
571  $P_{\text{co}}^{\text{JRJS}, \infty} \rightarrow F_{\gamma_{\xi_D}}(\gamma_{th}^D)$  and  $1 - P_{\text{so}}^{\text{JRJS}, \infty} \rightarrow F_{\gamma_{\xi_E}}(\gamma_{th}^E)$ . Finally,  
572 substituting the corresponding results into (16), we arrive at the  
573 RSR of the JRJS strategy.

*Remark 4:* It can be seen from the RSR expression of (28)  
574 again that the rate-pair setting  $(R_0, R_s)$  has an inconsistent  
575 influence on the RSR, and hence, we have to carefully adjust  $R_0$   
576 and  $R_s$  to balance the reliability versus security performance.  
577 Let us now focus our attention on the differences between the  
578 JRJS scheme and the TBRS arrangement.

579 First, we may find that the power sharing ratio  $\lambda$  between  
580 the relay and the jammer plays a very important role. The  
581 optimization of  $\lambda$  will be investigated from an effective secrecy  
582 throughput optimization point of view in the following.

583 Second, it is plausible that, in contrast to the behavior of the  
584 TBRS strategy,  $\Lambda^{\text{JRJS}}$  of (28) is only related to the delay of the  
585 second hop, but it is still a monotonically decreasing function of  
586  $\rho_{RD}$ . This implies that the improvement of the channel quality  
587 of the JRJS will achieve a more pronounced COP improvement  
588 than the associated SOP improvement. Furthermore, recalling  
589 that the RSR is considered in the high-SNR region, it has no  
590 dependence on the first hop quality. This is due to the fact that  
591 if the first-hop channel quality is sufficiently high for ensuring  
592 a successful transmission, the asymptotic CDFs of  $\xi_D$  and  $\xi_E$   
593 in (29) and (30) associated with  $\eta \rightarrow \infty$  will converge to a  
594 nonzero limit at high SNRs, which ultimately dominates the  
595 COP and the SOP.

### 597 C. Effective Secrecy Throughput

598 Before proceeding to the effective secrecy throughput analy-  
599 sis, we also have to investigate the RSCP.

600 *Lemma 3:* The RSCP of our JRJS strategy may be approxi-  
601 mated as in (31), shown at the bottom of the next page, where  
602 we have  $\theta_{1,k} = (\gamma_{th}^D (\gamma_{th}^D + 1)) / (\gamma_{th}^D + \hat{\phi}_k)$ ,  $\theta_2 = \gamma_{th}^D - \gamma_{th}^E +$   
603  $(\gamma_{th}^E (\gamma_{th}^E + 1)) / (\gamma_{th}^E + \hat{\phi})$ , and  $\hat{\phi} = \lambda \eta \sigma_{RE}^2 / ((1 - \lambda) \eta \sigma_{RE}^2 + 1)$ .

604 *Proof:* The proof is given in Appendix C.

605 Apart from the rate pair  $(R_0, R_s)$ , the aforementioned  $P_{R\&S}^{\text{JRJS}}$   
606 of (31) is also a function of the power sharing ratio  $\lambda$  between  
607 the selected relay and the jammer.

608 Given the complexity of the RSCP expression, it is quite  
609 a challenge to find a closed-form result for maximizing the

$$\Lambda^{\text{JRJS}} = \frac{(2^{2R_0} - 1)}{(2^{2(R_0 - R_s)} - 1)} \sum_{k=0}^{K_r-1} \binom{K_r-1}{k} \frac{(-1)^k K_r [(K_r - 1)(1 - \rho_{RD}^2) + 1] [(\lambda^{-1} - 1)(2^{2(R_0 - R_s)} - 1) + 1]}{[(K_r - 1)(1 - \rho_{RD}^2) + 1] (k+1)(\lambda^{-1} - 1)(2^{2R_0} - 1) + K_r [k(1 - \rho_{RD}^2) + 1]} \quad (28)$$

610 effective secrecy throughput that  $\max_{0 < \lambda < 1} \varsigma = R_s P_{R\&S}^{\text{JRJS}}$ . Al-  
 611 ternatively, we can focus on the asymptotic analysis in the high-  
 612 SNR region and try to find a general closed-form solution for  $\lambda$ .  
 613 Specifically, when we have  $\eta \rightarrow \infty$ ,  $P_{R\&S}^{\text{JRJS}}$  will be dominated  
 614 by the channel quality of the second hop; hence, we have

$$P_{R\&S}^{\text{JRJS},\infty}(R_0, R_s, \lambda) \approx \Pr \{ \xi_D > \gamma_{th}^D, \xi_E < \gamma_{th}^E \} \\ = [1 - F_{\xi_D}(\gamma_{th}^D)] F_{\xi_E}(\gamma_{th}^E) \quad (32)$$

615 where the approximation is based on the fact that, in contrast to  
 616 both  $F_{\xi_D}(\gamma_{th}^D)$  and  $F_{\xi_E}(\gamma_{th}^E)$ , which converge to a nonzero limit  
 617 regardless of  $\eta$ , the first hop's  $F_{\gamma_{SR}}(x)$  will tend to zero, and  
 618 hence, it can be neglected. Substituting the asymptotic results  
 619 of (29) and (30) into (33), we can obtain  $P_{R\&S}^{\text{JRJS},\infty}$ . In contrast to  
 620 the TBRS case operating without jamming, as the SNR tends to  
 621  $\infty$ , the RSCP will tend to a nonzero value and, upon increasing  
 622 the transmit SNR beyond a certain limit, will no longer increase  
 623 the effective secrecy throughput.

624 Then, based on (32), we arrive at the approximated optimal  
 625 value  $\lambda_{\text{opt}}$ , which is the solution of the following equation:

$$\frac{\partial P_{R\&S}^{\text{JRJS},\infty}(R_0, R_s, \lambda)}{\partial \lambda} = 0. \quad (33)$$

626 Then, by exploiting the approximation of  $[k(1 - \rho_{RD}^2) + 1]/$   
 627  $(k + 1) \approx 1 - \rho_{RD}^2$  in (29) for a large  $\rho_{RD}$  (practically, the CSI  
 628 delay is small, and  $\rho_{RD} \rightarrow 1$ ), we have

$$\lambda_{\text{subopt}} = \frac{\sqrt{[(K_r - 1)(1 - \rho_{RD}^2) + 1] \gamma_{th}}}{\sqrt{[(K_r - 1)(1 - \rho_{RD}^2) + 1] \gamma_{th} + \sqrt{K_r(1 - \rho_{RD}^2)}}} \quad (34)$$

629 where  $\gamma_{th} = (2^{2R_0} - 1)(2^{2(R_0 - R_s)} - 1)$ . It is clear that this  
 630 value is determined by the number of relays and  $(R_0, R_s)$ .

## 631 V. NUMERICAL RESULTS

632 Both our numerical and Monte Carlo simulation results are  
 633 presented here for verifying the theoretical PLS performance  
 634 analysis of the multiple-relay-aided network under CSI feed-

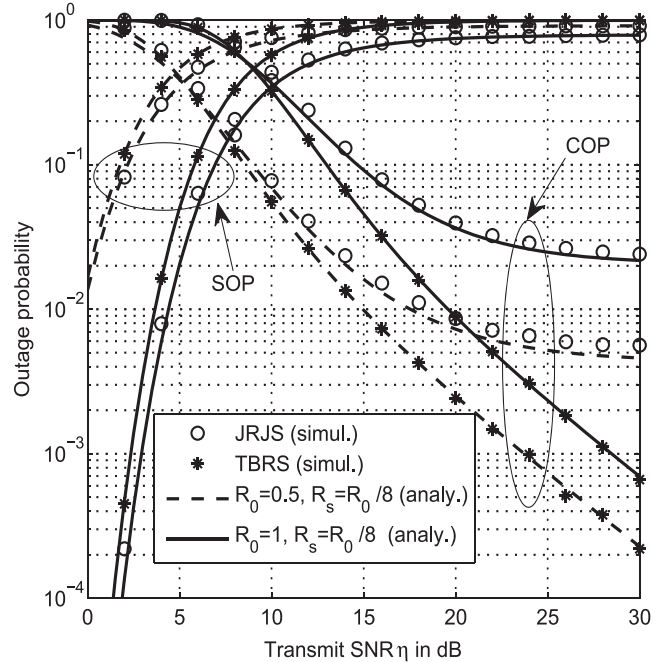


Fig. 2. COP and SOP versus transmit SNR for the TBRS and JRJS strategies in conjunction with different rate pairs, for  $N_t = K_r = 3$ ,  $f_d T_d = 0.1$ , and  $\lambda = 1/10$ .

back delays. Explicitly, the COP, SOP, RSCP, and RSR are 635  
 636 validated for both the TBRS and JRJS strategies. Furthermore, 637  
 638 the effects of feedback delays and system parameters (including 639  
 640 the transmission rate pair  $(R_0, R_s)$  and the power sharing ratio 641  
 642  $\lambda$  between the relay and the jammer) on the achievable effective 643  
 644 secrecy throughput are evaluated. The Rayleigh fading model 645  
 646 is employed for characterizing all communication links in our 647  
 648 system. Additionally, we set the total power to  $P = 1$  and 649  
 650  $\sigma_{SR}^2 = \sigma_{RD}^2 = \sigma_{RE}^2 = 1$ , and used  $T_{dSR} = T_{dRD} = T_d$ . 651

652 Fig. 2 plots the COP and the SOP versus the transmit SNR for 653  
 654 both the TBRS and JRJS strategies in conjunction with different 655  
 656 rate pairs. The analytical lines are plotted by using (11) and (14) 657  
 658 for the TBRS strategy and by using (26) and (27) for the JRJS 659

$$P_{R\&S}^{\text{JRJS}}(R_0, R_s, \lambda) \approx \sum_{n=0}^{N_t-1} \sum_{k=0}^{K_r-1} \sum_{m=0}^{N_t-1-n} (-1)^k \binom{N_t-1}{n} \binom{K_r-1}{k} \binom{N_t-1-n}{m} \\ \times \frac{K_r \rho_{SR}^{2(N_t-1-n)} (1 - \rho_{SR}^2)^n \hat{\varphi}_k (\gamma_{th}^D)^{N_t-1-n-m}}{(N_t-1-n)! (k+1) \bar{\gamma}_{SR}^{N_t-n} (\gamma_{th}^D + \hat{\varphi}_k) e^{\frac{\gamma_{th}^D}{\bar{\gamma}_{SR}} + \frac{\gamma_{th}^D}{\bar{\gamma}_{RD} \omega_k}}} \\ \times \left\{ \theta_{1,k}^{m+1} e^{\frac{\theta_1}{\bar{\gamma}_{SR}}} \Gamma(m+2) \Gamma\left(-m-1, \frac{\theta_{1,k}}{\bar{\gamma}_{SR}}\right) - \frac{\hat{\varphi} e^{-\gamma_{th}^E / \bar{\gamma}_{RE}}}{(\gamma_{th}^E + \phi) (\theta_{1,k} - \theta_2)} \Gamma(m+3) \right. \\ \times \left[ \theta_2^{m+2} e^{\frac{\theta_2}{\bar{\gamma}_{SR}}} \Gamma\left(-m-2, \frac{\theta_2}{\bar{\gamma}_{SR}}\right) - \theta_{1,k}^{m+2} e^{\frac{\theta_1}{\bar{\gamma}_{SR}}} \Gamma\left(-m-2, \frac{\theta_{1,k}}{\bar{\gamma}_{SR}}\right) \right] + \Gamma(m+2) (\gamma_{th}^D - \gamma_{th}^E) \\ \left. \times \left[ \theta_2^{m+1} e^{\frac{\theta_2}{\bar{\gamma}_{SR}}} \Gamma\left(-m-1, \frac{\theta_2}{\bar{\gamma}_{SR}}\right) - \theta_{1,k}^{m+1} e^{\frac{\theta_1}{\bar{\gamma}_{SR}}} \Gamma\left(-m-1, \frac{\theta_{1,k}}{\bar{\gamma}_{SR}}\right) \right] \right\} \quad (31)$$

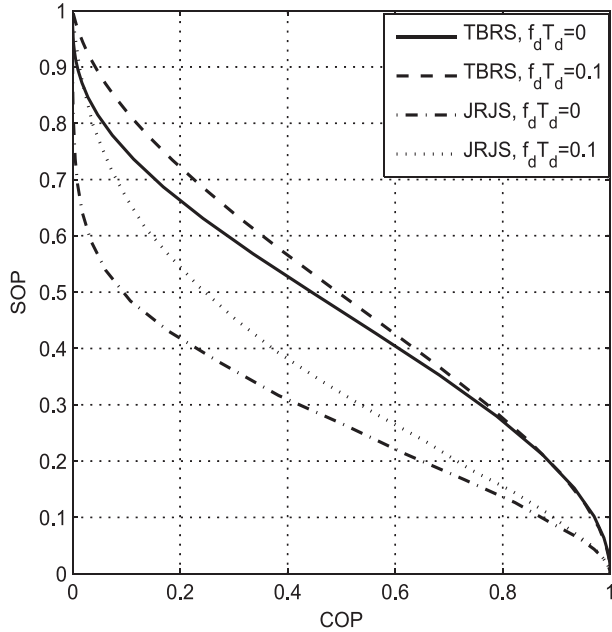


Fig. 3. SOP versus COP for the TBRS and JRJS strategies with different feedback delays for  $N_t = K_r = 3$ ,  $R_s = R_0/8$ , and  $\lambda = 1/10$ .

648 case, respectively. It can be clearly seen from the figure that the  
649 analytical and simulated outage probability curves match well,  
650 which confirms the accuracy of the mathematical analysis. As  
651 expected, compared with the TBRS strategy, the SOP of the  
652 JRJS strategy is much better, whereas the COP is worse. We  
653 can also find that both the COP and the SOP will converge to an  
654 outage floor at high SNRs for the JRJS strategy. The reason for  
655 this is that the jammer also imposes interference on the destina-  
656 tion and the interference inflicted increases with the SNR. Thus,  
657 the designers have to take into account the tradeoff between  
658 the reliability and the security and the interference imposed on  
659  $D$ , particularly when considering the JRJS strategy. Moreover,  
660 we can observe in Fig. 2 that increasing the transmission rate  
661 decreases the COP and increases the SOP.

662 Fig. 3 further characterizes the SOP versus COP for both the  
663 TBRS and JRJS strategies based on the numerical results in  
664 Fig. 2, which shows the tradeoff between the reliability and the  
665 security. It can be seen from the figure that the SOP decreases as  
666 the COP increases, and for a specific COP, the SOP of the JRJS  
667 scheme is strictly lower than that of TBRS. This confirms that  
668 the JRJS scheme performs better than the conventional TBRS  
669 scheme. Furthermore, the CSI feedback delay will also degrade  
670 the system tradeoff performance.

671 Fig. 4 illustrates the RSCP versus transmit SNR for the  
672 TBRS strategy in the context of different network configura-  
673 tions, including different rate pairs, different number of relays,  
674 and both perfect and outdated CSI feedback scenarios. The  
675 analytical lines are plotted by using the approximation in (20).  
676 We may conclude from the figure that the rate-pair setting  
677 ( $R_0, R_s$ ) determines both the reliability and security transmis-  
678 sion performance. These curves also show that the RSCP is a  
679 concave function of the transmit SNR, whereas the continued  
680 boosting of the SNR would only decrease the probability of  
681 a successful transmission. We can observe from Fig. 4 that,

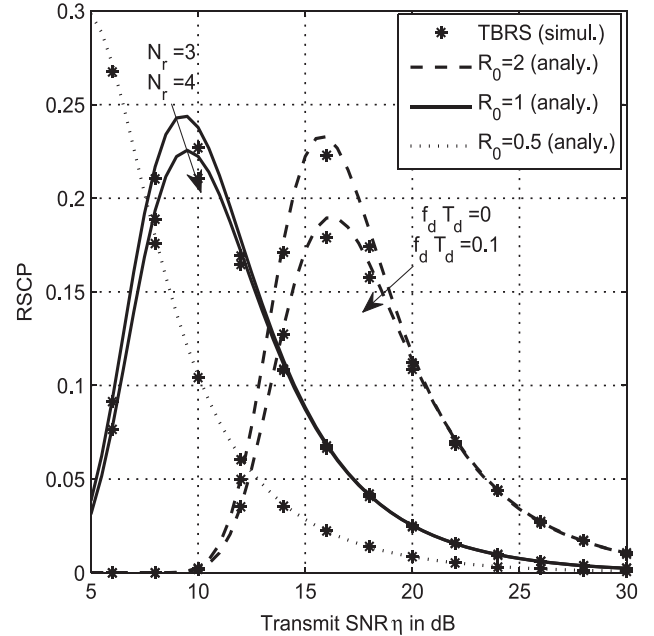


Fig. 4. RSCP versus transmit SNR for the TBRS strategy with different rate pairs for  $N_t = K_r = 3$ ,  $f_d T_d = 0.1$ .

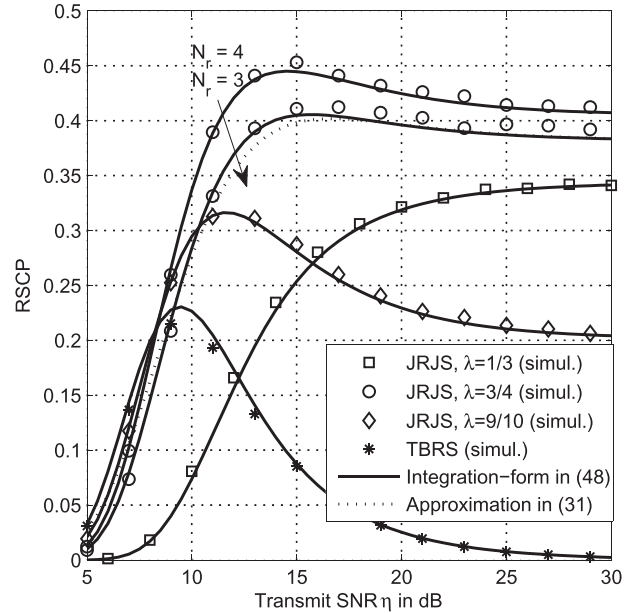


Fig. 5. RSCP versus transmit SNR for the JRJS strategy for different power sharing ratios  $\lambda$  and for  $N_t = K_r = 3$ ,  $f_d T_d = 0.1$ , and  $R_0 = 1$ ,  $R_s = R_0/8$ .

for a high transmit SNR, total reliability can be guaranteed,  
682 whereas the associated grade of security is severely eroded.  
683 Furthermore, increasing the number of relays and decreasing  
684 the feedback delay will improve both the reliability and security  
685 performance. 686

The RSCP of the JRJS strategy is presented in Fig. 5 for  
687 different power sharing ratios between relaying and jamming.  
688 Both the integration form (45) and the approximated closed  
689 form in (31) match well with the Monte Carlo simulations.  
690 The performance of the TBRS strategy is also included for  
691 comparison. The JRJS scheme outperforms the TBRS operating  
692

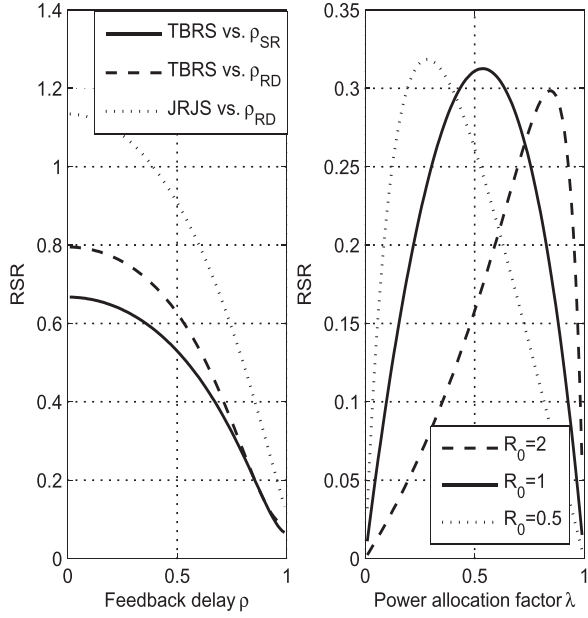


Fig. 6. RSR versus feedback delay coefficient ( $R_0 = 1$ ,  $R_s = R_0/8$ ,  $\lambda = 3/4$ ) and power sharing ratio  $\lambda$  ( $R_s = R_0/8$ ,  $\rho_{SR} = \rho_{RD} = 0.9$ ) for the TBRS and JRJS strategies, with  $N_t = K_r = 3$ .

693 without jamming under the scenario considered when encoun-  
 694 tering comparable relay–destination and relay–eavesdropper  
 695 channels. For some extreme configurations (when the relay–  
 696 eavesdropper links are comparatively weak), this statement  
 697 may not hold, but this scenario is beyond the scope of this  
 698 paper. The maximum RSCP appears at about  $\eta = 15$  dB  
 699 for the JRJS strategy using  $\lambda = 3/4$ , whereas it is  $\eta = 10$  dB  
 700 for the TBRS strategy. Furthermore, as expected, increasing the  
 701 number of available relays and jamming nodes will always be  
 702 able to improve the reliability and security performance. How-  
 703 ever, the continued boosting of the jammer’s power (decreasing  
 704  $\lambda$ ) will not always improve the overall performance, because  
 705 the interference improves initially the security, but then, it starts  
 706 to reduce the reliability as  $\lambda$  decreases. This further motivates  
 707 the designer to carefully take into account the power sharing  
 708 between relaying and jamming. The effect of the rate-pair  
 709 setting on the security and reliability of the JRJS strategy is  
 710 neglected here, which follows a similar trend to that of the  
 711 TBRS strategy.

712 Fig. 6 characterizes the RSR versus feedback delay and  
 713 power sharing ratio for both TBRS and JRJS, in which the  
 714 RSR curves are plotted by using (17) and (28), respectively.  
 715 The first illustration shows that the RSR decreases as the delay  
 716 coefficients ( $\rho_{SR}$  and  $\rho_{RD}$ ), which confirms that the im-  
 717 provement of reliability becomes more pronounced than the  
 718 reduction of the security as the feedback delay decreases.  
 719 This observation implies an improvement in terms of the  
 720 security–reliability tradeoff. In addition, the RSR versus  $\rho_{RD}$   
 721 is larger than that of  $\rho_{SR}$ , which indicates that the impact of the  
 722 second-hop CSI feedback delay is more prominent. The other  
 723 illustration in the right demonstrates that the RSR is a concave  
 724 function of the power sharing ratio, which reflects the tradeoff  
 725 between the reliability and the security struck by adjusting  $\lambda$ .

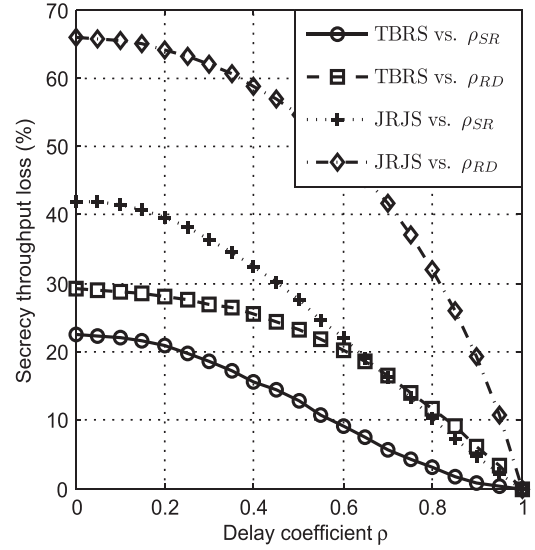


Fig. 7. Percentage secrecy throughput loss versus delay coefficients with  $N_t = K_r = 3$ ,  $R_0 = 1$ ,  $R_s = R_0/8$ ,  $\lambda = 3/4$ , and  $\eta = 10$  dB.

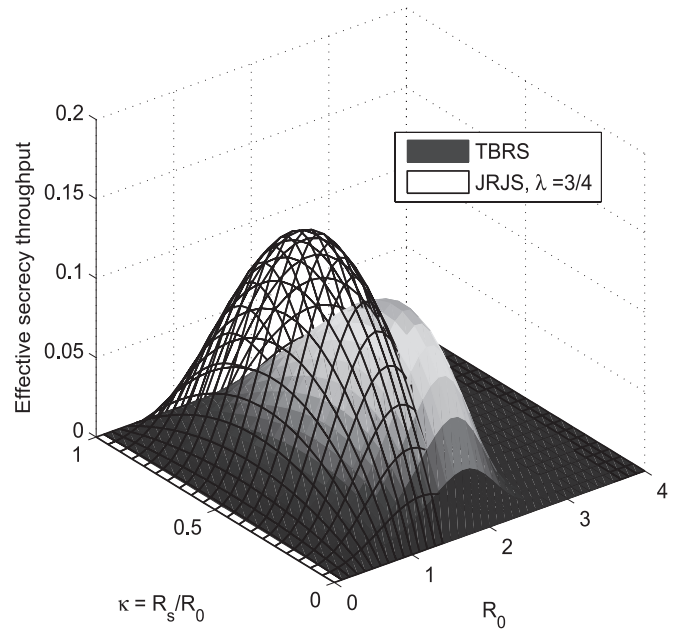


Fig. 8. Secrecy throughput versus  $R_0$  and  $\kappa = R_s/R_0$  for both the TBRS and JRJS strategies with  $N_t = K_r = 3$ ,  $f_d T_d = 0.1$ , and  $\eta = 15$  dB.

To further evaluate the effect of feedback delays on the  
 726 secrecy performance, Fig. 7 plots the resultant percentage of 727  
 secrecy throughput loss versus the delay, which is defined as 728

$$\text{S}_{\text{loss}} = \frac{S_{\text{no-delay}} - S_{\text{delay}}}{S_{\text{no-delay}}}. \quad (35)$$

It can be seen from the figure that, compared with the TBRS  
 729 scheme, JRJS is more sensitive to the feedback delays. Further-  
 730 more, recalling that increasing the delay coefficient  $\rho_{SR}$  of the  
 731 first hop improves the reliability, but at the same time also helps  
 732 the eavesdropper, it is not surprising that the secrecy throughput  
 733 loss due to the second-hop feedback delay is more pronounced. 734

Fig. 8 illustrates the achievable effective secrecy throughput  
 735 for both the TBRS and JRJS strategies versus the codeword 736

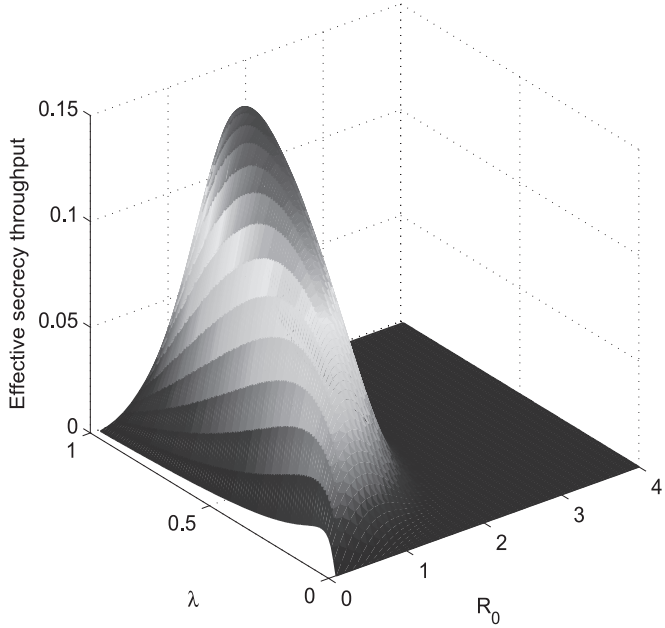


Fig. 9. Secrecy throughput versus  $R_0$  and  $\lambda$  for the JRJS strategy with  $N_t = K_r = 3$ ,  $f_d T_d = 0.1$ ,  $\eta = 15$  dB, and  $R_s/R_0 = 1/8$ .

737 transmission rate  $R_0$  and the secrecy code ratio  $\kappa = R_s/R_0$   
 738 with no outage constraints ( $v = \delta = 1$ ). The values of the  
 739 effective secrecy throughput are plotted by using  $\zeta = R_s P_{R\&S}$ .  
 740 We can observe in Fig. 8 that, subject to a fixed code rate  
 741 ratio  $\kappa$ , the effective secrecy throughput increases to a peak  
 742 value as  $R_0$  reaches its optimal value and then decreases. This  
 743 phenomenon can be explained as follows. At a low transmission  
 744 rate, although the COP increases with  $R_0$ , which has a negative  
 745 effect on the effective secrecy throughput, both the secrecy  
 746 rate and the SOP performance will benefit. However, after  
 747 reaching the optimal  $R_0$ , the effective secrecy throughput drops  
 748 since the main link cannot afford a reliable transmission, and  
 749 the resultant COP increase becomes dominant. On the other  
 750 hand, subject to a fixed  $R_0$  (which results in a constant COP),  
 751 the effective secrecy throughput is also a concave function  
 752 of  $\kappa$ , and increasing the code rate ratio ultimately results  
 753 in an increased secrecy information rate at the cost of an  
 754 increased SOP.

755 The achievable effective secrecy throughput for the JRJS  
 756 strategy is also presented in Fig. 8, and similar conclusions and  
 757 trends can be observed to that of the TBRS case. Additionally,  
 758 the comparison of the two strategies indicates that the JRJS  
 759 scheme attains a higher effective secrecy throughput than the  
 760 TBRS scheme operating without jamming, even if no power  
 761 sharing optimization has been employed.

762 Fig. 9 further illustrates the impact of power sharing between  
 763 the relay and the jammer on the achievable effective secrecy  
 764 throughput of the JRJS strategy versus  $R_0$  in the absence of  
 765 outage constraints. Given a fixed code rate pair  $(R_0, R_s)$ , the  
 766 effective secrecy throughput follows the trend of the RSCP,  
 767 which is a concave function of  $\lambda$ , as shown in Fig. 6. The  
 768 interference introduced by the jammer initially improves both  
 769 the reliability and the security as  $\lambda$  increases, but this trend is  
 770 reversed beyond a certain point.

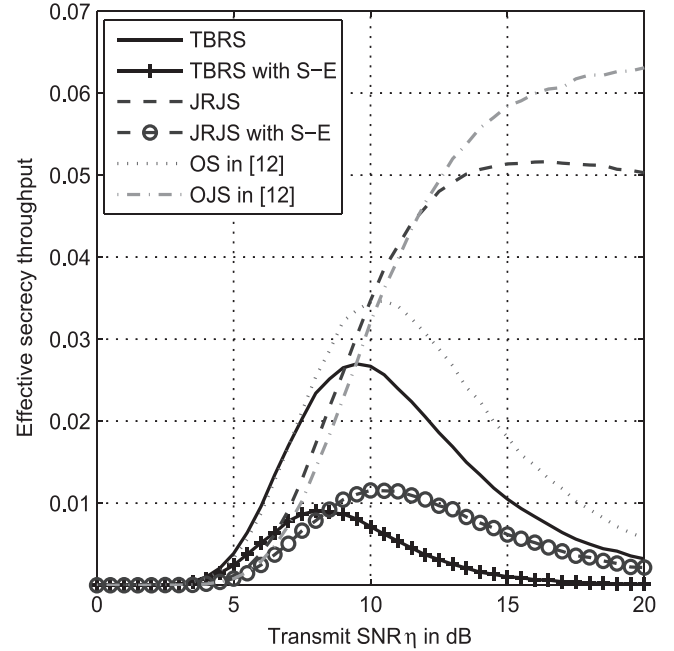


Fig. 10. Comparisons for different strategies with and without the S-E link, for  $N_t = K_r = 3$ ,  $R_0 = 1$ ,  $R_s = R_0/8$ ,  $f_d T_d = 0.1$ , and  $\lambda = 3/4$ .

## VI. DISCUSSION

771

### A. Impact of the S-E Link

772

We note that the introduction of the S-E link, i.e., the  
 773 information leakage in the first phase, is very critical to the  
 774 security. There are also some research studies focusing on  
 775 the corresponding secure transmission design and performance  
 776 evaluation for cooperative networks with the S-E link, such  
 777 as [15] and [16]. Here, we assume that the eavesdropper can  
 778 receive information directly from the source in the first phase.  
 779 Thus, following the steps in the prior sections, for the TBRS  
 780 and JRJS schemes, it is clear that the SNR experienced at the  
 781 eavesdropper should be rewritten as

$$\tilde{\gamma}_E^T = \gamma_{SE} + \gamma_E^T \quad (36)$$

where  $\gamma_{SE} = P_s |\mathbf{w}_{\text{opt}}(t|T_{dSR}) \mathbf{h}_{SE}(t)|^2 / N_0$  follows the ex-  
 783 ponential distribution with the average value  $\bar{\gamma}_{SE}$ ,  $\tau =$   
 784  $\{\text{TBRS, JRJS}\}$ , and  $\gamma_E^T$  has been defined in (4) and (7).  
 785

Then, the corresponding SOP, RSCP, and effective secrecy  
 786 throughput have to be reconsidered. Unfortunately, to the best  
 787 of our knowledge, it is a mathematically intractable problem  
 788 to obtain closed-form results for the related performance eval-  
 789 uations. Therefore, we resorted to numerical simulations for  
 790 further investigating the impact of the S-E link. Fig. 10 com-  
 791 pares the effective secrecy throughput of the TBRS and JRJS  
 792 schemes both with and without considering the direct S-E  
 793 link. It becomes clear that the information leakage in the first  
 794 phase will lead to a severe security performance degradation,  
 795 particularly for the JRJS scheme, which will no longer be  
 796 capable of maintaining a steady throughput at high SNRs. The  
 797 reason for this trend is that increasing the transmit SNR will  
 798 help the eavesdropper in the presence of the direct S-E link.  
 799



## 800 B. Comparisons

801 Here, based on the outdated CSI assumption, we provide per-  
802 formance comparisons with a range of other schemes advocated  
803 in [12] with the aid of the proposed outage-based characteriza-  
804 tion. Fig. 10 also incorporates our effective secrecy throughput  
805 performance comparison, where the optimal selection (OS)  
806 regime and the optimal selection combined with jamming (OSJ)  
807 were proposed in [12]. They are formulated as

$$808 \text{ OS : } R^* = \arg \max_{R_k \in \mathcal{R}} \left\{ \frac{\tilde{\gamma}_{R_k D}}{\tilde{\gamma}_{R_k E}} \right\} \quad (37)$$

$$809 \text{ OSJ : } \begin{cases} R^* = \arg \max_{R_k \in \mathcal{R}} \left\{ \frac{\tilde{\gamma}_{R_k D}}{\tilde{\gamma}_{R_k E}} \right\} \\ J^* = \arg \min_{R_k \in \mathcal{R} - R^*} \left\{ \frac{\tilde{\gamma}_{R_k D}}{\tilde{\gamma}_{R_k E}} \right\} \end{cases} \quad (38)$$

810 where  $\tilde{\gamma}_{R_k E}$  is the delayed version of the instantaneous CSI of  
811 the R–E link. It should be noted that this constitutes an entirely  
812 new performance characterization of these schemes from the  
813 perspective of the effective secrecy throughput. It is shown in  
814 Fig. 1 that the selection combined with jamming outperforms  
815 the corresponding nonjamming techniques at high SNRs, albeit  
816 this trend may no longer prevail at low SNRs. In comparison,  
817 compared with those selections relying on the average SNRs of  
818 the R–E link, the optimal selections relying on the idealized  
819 simplifying assumptions of having global CSI (OS and OSJ  
820 schemes) knowledge can only achieve throughput gains at high  
821 SNRs due to the inevitable feedback delay.

## 822 VII. CONCLUSION

823 An outage-based characterization of cooperative relay net-  
824 works has been provided in the face of CSI feedback delays.  
825 Two types of relaying strategies were considered, namely, the  
826 TBRS strategy and the JRJS strategy. Closed-form expressions  
827 of the COP, the SOP, and the RSCP, as well as of the RSR,  
828 were derived. The RSR results demonstrated that the reliability  
829 is improved more substantially than the security performance  
830 when the CSI feedback delays are reduced. Furthermore, we  
831 presented a modified effective secrecy throughput definition  
832 and demonstrated that the JRJS strategy achieves a significant  
833 effective secrecy throughput gain over the TBRS strategy. The  
834 transmit SNR, the secrecy codeword rate setting, and the power  
835 sharing ratio between the relay and jammer nodes play impor-  
836 tant roles in striking a balance between the reliability and the  
837 security in terms of the secrecy throughput. The impact of the  
838 direct S–E link and the performance comparisons with other  
839 selection schemes were also included. Additionally, our results  
840 demonstrate that JRJS is more sensitive to the feedback delays  
841 and that the secrecy throughput loss due to the second-hop  
842 feedback delay is more pronounced than that due to the first-  
843 hop one.

## 844 APPENDIX A

### 845 PROOF OF PROPOSITION 1

846 To simplify the asymptotic performance analysis, (3) can be  
847 expressed in a more mathematically tractable form by the com-  
848 monly used tight upper bound of  $\gamma_D^{\text{TBRS}} \leq \min\{\gamma_{SR}, \gamma_{R^*D}\}$

849 and  $\gamma_E^{\text{TBRS}} \leq \min\{\gamma_{SR}, \gamma_{R^*E}\}$ . When we have  $\eta \rightarrow \infty$ , based  
850 on the CDFs in (9) and (10) and closing the smallest order terms  
851 of  $x/\eta$ , we have

$$852 F_{\gamma_{SR}}(x) \rightarrow 1 - \left[ \sum_{n=0}^{N_t-1} \binom{N_t-1}{n} \rho_{SR}^{2(N_t-1-n)} (1 - \rho_{SR}^2)^n \right. \\ 853 + \sum_{n=0}^{N_t-2} \binom{N_t-1}{n} \times \rho_{SR}^{2(N_t-1-n)} \\ 854 \left. \times (1 - \rho_{SR}^2)^n \frac{x}{\tilde{\gamma}_{SR}} + \mathcal{O}\left(\frac{x}{\tilde{\gamma}_{SR}}\right) \right] \\ 855 \times \left[ 1 - \frac{x}{\tilde{\gamma}_{SR}} + \mathcal{O}\left(\frac{x}{\tilde{\gamma}_{SR}}\right) \right] \\ 856 = 1 - \left[ 1 + (1 - (1 - \rho_{SR}^2)^{N_t-1}) \frac{x}{\tilde{\gamma}_{SR}} + \mathcal{O}\left(\frac{x}{\tilde{\gamma}_{SR}}\right) \right] \\ 857 \times \left[ 1 - \frac{x}{\tilde{\gamma}_{SR}} + \mathcal{O}\left(\frac{x}{\tilde{\gamma}_{SR}}\right) \right] \\ 858 = (1 - \rho_{SR}^2)^{N_t-1} \frac{x}{\tilde{\gamma}_{SR}} + \mathcal{O}\left(\frac{x}{\tilde{\gamma}_{SR}}\right) \quad (39)$$

859 where  $\mathcal{O}(x)$  denotes the high-order infinitely small contribu-  
860 tions as a function of  $x$ , and

$$861 F_{\gamma_{R^*D}}(x) \rightarrow 1 - \sum_{k=0}^{K_r-1} (-1)^k \frac{K_r}{k+1} \binom{K_r-1}{k} \\ 862 \times \left[ 1 - \frac{k+1}{k(1 - \rho_{RD}^2) + 1} \frac{x}{\tilde{\gamma}_{RD}} + \mathcal{O}\left(\frac{x}{\tilde{\gamma}_{RD}}\right) \right] \\ 863 = \sum_{k=0}^{K_r-1} (-1)^k \binom{K_r-1}{k} \frac{K_r}{k(1 - \rho_{RD}^2) + 1} \\ 864 \times \frac{x}{\tilde{\gamma}_{RD}} + \mathcal{O}\left(\frac{x}{\tilde{\gamma}_{RD}}\right). \quad (40)$$

865 Then, applying the upper bound of the receiver SNR, we may  
866 rewrite the COP and the SOP of the TBRS strategy at high  
867 SNRs as

$$868 P_{\text{co}}^{\text{TBRS}, \infty} = 1 - (1 - F_{\gamma_{SR^*}}(\gamma_{th}^D)) (1 - F_{\gamma_{R^*D}}(\gamma_{th}^D)) \\ 869 = \left[ \frac{(1 - \rho_{SR}^2)^{N_t-1}}{\sigma_{SR}^2} + \sum_{k=0}^{K_r-1} (-1)^k \right. \\ 870 \left. \times \binom{K_r-1}{k} \frac{K_r}{[k(1 - \rho_{RD}^2) + 1] \sigma_{RD}^2} \right] \frac{2^{2R_0} - 1}{\eta} \quad (41)$$

871 and according to the fact that  $\gamma_{R^*E}$  is exponentially distributed,  
872 we have

$$873 1 - P_{\text{so}}^{\text{TBRS}, \infty} = 1 - (1 - F_{\gamma_{SR^*}}(\gamma_{th}^E)) (1 - F_{\gamma_{R^*E}}(\gamma_{th}^E)) \\ 874 = \left[ \frac{(1 - \rho_{SR}^2)^{N_t-1}}{\sigma_{SR}^2} + \frac{1}{\sigma_{RE}^2} \right] \frac{2^{2(R_0 - R_s)} - 1}{\eta}. \quad (42)$$

875 Finally, substituting (41) and (42) into the definition of RSR  
876 in (16), we can obtain (17).

859 APPENDIX B  
860 PROOF OF LEMMA 1

861 According to the description of COP and SOP, replacing  
862  $F_{\gamma_{R^*D}}(x)$  and  $F_{\gamma_{R^*E}}(x)$  by  $F_{\xi_D}(x)$  and  $F_{\xi_E}(x)$  in (12) and (14)  
863 will involve a mathematically intractable integration of the form

$$\Upsilon(a, b, \mu, \nu) = \int_0^{\infty} \frac{z^a}{z+b} \exp\left(-\mu z - \frac{\nu}{z}\right) dz \quad (43)$$

864 which, to the best of our knowledge, does not have a closed-  
865 form solution. Alternatively, bearing in mind that the preceding  
866 integration has a great matter with  $\xi_D$ , we now focus our  
867 attention on the approximation of  $\xi_D$ . Based on the PDF  
868 results in (23), it may be seen that  $\gamma_{J^*D}$  obeys an exponential  
869 distribution. Then, we can approximate  $\hat{\gamma}_{J^*D} = \gamma_{J^*D} + 1$  by  
870 the exponential distribution as well, with an average value  
871 of  $\mathbb{E}\{\hat{\gamma}_{J^*D}\} = ((K_r - 1)(1 - \rho_{RD}^2) + 1)\bar{\gamma}_{RD} + K_r)/K_r$  by  
872 assuming that the AWGN term "1" is part of the stochastic  
873 mean terms. The approximation based on this method provides  
874 a very accurate analysis, and the accuracy of this method is  
875 verified by the numerical results of [34]. Thus, the CDF of  
876  $\hat{\xi}_D = \gamma_{R^*D}/\hat{\gamma}_{J^*D}$  can be derived as

$$F_{\hat{\xi}_D}(x) = \sum_{k=0}^{K_r-1} (-1)^k \binom{K_r-1}{k} \frac{K_r}{k+1} \frac{x}{x + \hat{\varphi}_k} \quad (44)$$

877 where  $\hat{\varphi}_k = \mathbb{E}\{\gamma_{R^*D}\}\mathbb{E}\{\hat{\gamma}_{J^*D}\}$ .

878 Then, substituting (44) into (11), we have

$$\begin{aligned} & F_{\gamma_{D}^{\text{JRJS}}}(x) \\ & \approx \sum_{n=0}^{N_t-1} \sum_{k=0}^{K_r-1} \sum_{m=0}^{N_t-1-n} \binom{N_t-1}{n} \binom{K_r-1}{k} \binom{N_t-1-n}{m} \\ & \times \frac{(-1)^k K_r \rho_{SR}^{2(N_t-1-n)} (1 - \rho_{SR}^2)^n \varphi_k x^{N_t-1-n-m} e^{-\frac{x}{\bar{\gamma}_{SR}}}}{(N_t-1-n)!(k+1)\bar{\gamma}_{SR}^{N_t-n} (x + \varphi_k)} \\ & \times \int_0^{\infty} \frac{z^{m+1}}{z + \frac{x(x+1)}{x+\varphi_k}} \exp\left(-\frac{z}{\bar{\gamma}_{SR}}\right) dz. \end{aligned} \quad (45)$$

879 Using [33, eq. (3.383.10)], we can obtain the CDF of  $\gamma_D^{\text{JRJS}}$  as

$$\begin{aligned} F_{\gamma_D^{\text{JRJS}}}(x) & \approx 1 - \sum_{n=0}^{N_t-1} \sum_{k=0}^{K_r-1} \sum_{m=0}^{N_t-1-n} \binom{N_t-1}{n} \\ & \times \binom{K_r-1}{k} \binom{N_t-1-n}{m} \\ & \times \frac{(-1)^k (K_r+1) \rho_{SR}^{2(N_t-1-n)} (1 - \rho_{SR}^2)^n}{(N_t-1-n)!(k+1)\bar{\gamma}_{SR}^{N_t-n}} \\ & \times \frac{\Gamma(m+2) \hat{\varphi}_k x^{N_t-n} (x+1)^{m+1}}{(x + \hat{\varphi}_k)^{m+2}} \\ & \times \exp\left[-\frac{x(\hat{\varphi}_k - 1)}{\bar{\gamma}_{SR}(x + \hat{\varphi}_k)}\right] \\ & \times \Gamma\left(-m-1, \frac{x(x+1)}{\bar{\gamma}_{SR}(x + \hat{\varphi}_k)}\right). \end{aligned} \quad (46)$$

880 Finally, substituting  $x = \gamma_{th}^D$  into (46), we obtain  $P_{\text{co}}^{\text{JRJS}}$ .

As far as the SOP is considered, we exploit the commonly 881  
used tight upper bound of  $\gamma_E^{\text{JRJS}} \geq (1/2) \min\{\gamma_{SR}, \xi_E\}$  to 882  
calculate it, which may be rewritten as 883

$$\begin{aligned} P_{\text{so}}^{\text{JRJS}} & \approx \Pr\left\{\frac{1}{2} \min\{\gamma_{SR}, \xi_E\} > \gamma_{th}^E\right\} \\ & = [1 - F_{\gamma_{SR}}(2\gamma_{th}^E)] [1 - F_{\xi_E}(2\gamma_{th}^E)]. \end{aligned} \quad (47)$$

Substituting (9) and (25) into (47), we obtain  $P_{\text{so}}^{\text{JRJS}}$ . 884

APPENDIX C 885  
PROOF OF LEMMA 3 886

According to the definition of the RSCP in (18), we can 887  
calculate it by 888

$$\begin{aligned} P_{RS}^{\text{JRJS}} & = \int_0^{\infty} \left[1 - F_{\xi_D}\left(\gamma_{th}^D + \frac{\gamma_{th}^D(\gamma_{th}^D + 1)}{z}\right)\right] \\ & \times F_{\xi_E}\left(\gamma_{th}^E + \frac{\gamma_{th}^E(\gamma_{th}^E + 1)}{z + \gamma_{th}^D - \gamma_{th}^E}\right) f_{\gamma_{SR^*}}(z + \gamma_{th}^D) dz. \end{aligned} \quad (48)$$

To make the integration mathematically tractable, we invoke 889  
a simple approximation for  $F_{\xi_E}(x)$  by treating the AWGN term 890  
"1" in  $\xi_E = \gamma_{R^*E}/(\gamma_{J^*E} + 1)$  as part of the stochastic mean 891  
terms. Hence, we have 892

$$F_{\xi_E}(x) = \frac{x}{x + \hat{\phi}} \quad (49)$$

where  $\hat{\phi} = \lambda\eta\sigma_{RE}^2/((1-\lambda)\eta\sigma_{RE}^2 + 1)$ . 893

Then, replacing the corresponding CDFs of the second hop 894  
with  $F_{\hat{\xi}_D}(x)$  and  $F_{\hat{\xi}_E}(x)$  in (26), the integration can be derived as 895

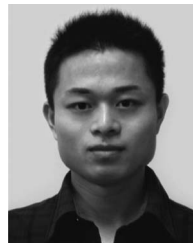
$$\begin{aligned} P_{RS}^{\text{JRJS}} & \approx 1 - \sum_{n=0}^{N_t-1} \sum_{k=0}^{K_r-1} \sum_{m=0}^{N_t-1-n} (-1)^k \binom{N_t-1}{n} \\ & \times \binom{K_r-1}{k} \binom{N_t-1-n}{m} \\ & \times \frac{(K_r+1) \rho_{SR}^{2(N_t-1-n)} (1 - \rho_{SR}^2)^n}{(N_t-1-n)!(k+1)\bar{\gamma}_{SR}^{N_t-n}} \\ & \times \frac{\hat{\varphi}_k (\gamma_{th}^D)^{N_t-1-n-m}}{\gamma_{th}^D + \hat{\varphi}_k} \exp\left(-\frac{\gamma_{th}^D}{\bar{\gamma}_{SR}} - \frac{\gamma_{th}^D}{\omega_k \bar{\gamma}_{RD}}\right) \\ & \times \int_0^{\infty} e^{-\frac{z}{\bar{\gamma}_{SR}}} z^{m+1} \left[ \frac{1}{z + \theta_{1,k}} - \frac{\hat{\phi} (z + \gamma_{th}^D - \gamma_{th}^E) e^{-\frac{\gamma_{th}^E}{\bar{\gamma}_{RE}}}}{(\gamma_{th}^E + \hat{\phi})(\theta_{1,k} - \theta_2)} \right] \\ & \times \left( \frac{1}{z + \theta_2} - \frac{1}{z + \theta_{1,k}} \right) dz \end{aligned} \quad (50)$$

where  $\hat{\varphi}_k$  and  $\hat{\phi}$  are introduced by relying on the similar approx- 896  
imation as in Appendix B. Then, using [33, eq. (3.383.10)], we 897  
obtain  $P_{R\&S}^{\text{JRJS}}$ . 898

899

## REFERENCES

- 900 [1] B. Schneier, "Cryptographic design vulnerabilities," *Computer*, vol. 31,  
901 no. 9, pp. 29–33, Sep. 1998.
- 902 [2] A. D. Wyner, "The wire-tap channel," *Bell Syst. Techn. J.*, vol. 54, no. 8,  
903 pp. 1355–1387, Oct. 1975.
- 904 [3] I. Csiszar and J. Korner, "Broadcast channels with confidential messages,"  
905 *IEEE Trans. Inf. Theory*, vol. IT-24, no. 3, pp. 339–348, May 1978.
- 906 [4] W. K. Harrison, J. Almeida, M. R. Bloch, S. W. McLaughlin, and  
907 J. Barros, "Coding for secrecy: An overview of error-control coding tech-  
908 niques for physical-layer security," *IEEE Signal Process. Mag.*, vol. 30,  
909 no. 5, pp. 41–50, Sep. 2013.
- 910 [5] P. K. Gopala, L. Lai, and H. E. Gamal, "On the secrecy capacity of  
911 fading channels," *IEEE Trans. Inf. Theory*, vol. 54, no. 10, pp. 4687–4698,  
912 Oct. 2008.
- 913 [6] Y. W. P. Hong, P. C. Lan, and C. C. J. Kuo, "Enhancing physical-layer  
914 secrecy in multi-antenna wireless systems: An overview of signal process-  
915 ing approaches," *IEEE Signal Process. Mag.*, vol. 30, no. 5, pp. 29–40,  
916 Sep. 2013.
- 917 [7] R. Bassily *et al.*, "Cooperative security at the physical layer: A summary  
918 of recent advances," *IEEE Signal Process. Mag.*, vol. 30, no. 5, pp. 16–28,  
919 Sep. 2013.
- 920 [8] L. Dong, Z. Han, A. P. Petropulu, and H. V. Poor, "Improving wire-  
921 less physical layer security via cooperating relays," *IEEE Trans. Signal  
922 Process.*, vol. 58, no. 3, pp. 1875–1888, Mar. 2010.
- 923 [9] J. Huang and A. L. Swindlehurst, "Cooperative jamming for secure com-  
924 munications in MIMO relay networks," *IEEE Trans. Signal Process.*,  
925 vol. 59, no. 10, pp. 4871–4884, Oct. 2011.
- 926 [10] Y. Zou, X. Wang, and W. Shen, "Optimal relay selection for physical-layer  
927 security in cooperative wireless networks," *IEEE J. Sel. Areas Commun.*,  
928 vol. 31, no. 10, pp. 2099–2111, Oct. 2013.
- 929 [11] Y. Zou, X. Wang, W. Shen, and L. Hanzo, "Security versus reliability  
930 analysis of opportunistic relaying," *IEEE Trans. Veh. Technol.*, vol. 63,  
931 no. 6, pp. 2653–2661, Jul. 2014.
- 932 [12] I. Krikidis, J. S. Thompson, and S. McLaughlin, "Relay selection  
933 for secure cooperative networks with jamming," *IEEE Trans. Wireless  
934 Commun.*, vol. 8, no. 10, pp. 5003–5011, Oct. 2009.
- 935 [13] J. Chen, R. Zhang, L. Song, Z. Han, and B. Jiao, "Joint relay and jammer  
936 selection for secure two-way relay networks," *IEEE Trans. Inf. Forensic  
937 Security*, vol. 7, no. 1, pp. 310–320, Feb. 2012.
- 938 [14] Z. Ding, M. Xu, J. Lu, and F. Liu, "Improving wireless security  
939 for bidirectional communication scenarios," *IEEE Trans. Veh. Technol.*,  
940 vol. 61, no. 6, pp. 2842–2848, Jul. 2012.
- 941 [15] C. Wang, H. M. Wang, and X. G. Xia, "Hybrid opportunistic relay-  
942 ing and jamming with power allocation for secure cooperative net-  
943 works," *IEEE Trans. Wireless Commun.*, vol. 14, no. 2, pp. 589–605,  
944 Feb. 2015.
- 945 [16] H. Deng, H. M. Wang, W. Guo, and W. Wang, "Secrecy transmission with  
946 a helper: To relay or to jam," *IEEE Trans. Inf. Forensic Security*, vol. 10,  
947 no. 2, pp. 293–307, Feb. 2015.
- 948 [17] B. He, X. Zhou, and T. D. Abhayapala, "Wireless physical layer secu-  
949 rity with imperfect channel state information: A survey," *ZTE Commun.*,  
950 vol. 11, no. 3, pp. 11–19, Sep. 2013.
- 951 [18] A. Mukherjee and A. L. Swindlehurst, "Robust beamforming for security  
952 in MIMO wiretap channels with imperfect CSI," *IEEE Trans. Signal  
953 Process.*, vol. 59, no. 1, pp. 351–361, Jan. 2011.
- 954 [19] J. Zhang and M. C. Gursoy, "Relay beamforming strategies for physical-  
955 layer security," in *Proc. CISS*, Princeton, NJ, USA, Mar. 2010, pp. 1–6.
- 956 [20] M. Bloch, J. Barros, M. R. D. Rodrigues, and S. W. McLaughlin, "Wire-  
957 less information-theoretic security," *IEEE Trans. Inf. Theory*, vol. 54,  
958 no. 6, pp. 2515–2534, Jun. 2008.
- 959 [21] X. Zhou, M. R. McKay, B. Maham, and A. Hjørungnes, "Rethink-  
960 ing the secrecy outage formulation: A secure transmission design  
961 perspective," *IEEE Commun. Lett.*, vol. 15, no. 3, pp. 302–304,  
962 Mar. 2011.
- 963 [22] J. Hu, Y. Cai, N. Yang, and W. Yang, "A new secure transmission scheme  
964 with outdated antenna selection," *IEEE Trans. Inf. Forensics Security*,  
965 to be published.
- 966 [23] J. Hu, W. Yang, N. Yang, X. Zhou, and Y. Cai, "On-off-based secure trans-  
967 mission design with outdated channel state information," *IEEE Trans.  
968 Veh. Technol.*, to be published.
- 969 [24] N. E. Wu and H. J. Li, "Effect of feedback delay on secure cooperative  
970 networks with joint relay and jammer selection," *IEEE Wireless Commun.  
971 Lett.*, vol. 2, no. 4, pp. 415–418, Aug. 2013.
- 972 [25] X. Guan, Y. Cai, and Y. Yang, "Secure transmission design and perfor-  
973 mance analysis for cooperation exploring outdated CSI," *IEEE Commun.  
974 Lett.*, vol. 18, no. 9, pp. 1637–1640, Sep. 2014.
- [26] L. Wang, S. Xu, W. Yang, W. Yang, and Y. Cai, "Security performance  
975 of multiple antennas multiple relaying networks with outdated relay  
976 selection," in *Proc. WCSP*, Hefei, China, Oct. 2014, pp. 1–6. 977
- [27] J. Huang and A. L. Swindlehurst, "Buffer-aided relaying for two-hop  
978 secure communication," *IEEE Trans. Wireless Commun.*, vol. 14, no. 1,  
979 pp. 152–164, Jan. 2015.
- [28] S. I. Kim, I. M. Kim, and J. Heo, "Secure transmission for multiuser relay  
980 networks," *IEEE Trans. Wireless Commun.*, vol. 14, no. 7, pp. 3724–3737,  
981 Jul. 2015. 982
- [29] Y. Ma, D. Zhang, A. Leith, and Z. Wang, "Error performance of transmit  
983 beamforming with delayed and limited feedback," *IEEE Trans. Wireless  
984 Commun.*, vol. 8, no. 3, pp. 1164–1170, Mar. 2009. 985
- [30] Z. Rezki, A. Khisti, and M. S. Alouini, "Ergodic secret message capac-  
986 ity of the wirechannel with finite-rate feedback," *IEEE Trans. Wireless  
987 Commun.*, vol. 13, no. 6, pp. 3364–3379, Jun. 2014. 988
- [31] X. Tang, R. Liu, P. Spasojevic, and H. V. Poor, "On the throughput of  
989 secure hybrid-ARQ protocols for Gaussian block-fading channels," *IEEE  
990 Trans. Inf. Theory*, vol. 55, no. 4, pp. 1575–1591, Apr. 2009. 991
- [32] H. A. Suraweera, M. Soysa, C. Tellambura, and H. K. Garg, "Performance  
992 analysis of partial relay selection with feedback delay," *IEEE Signal  
993 Process. Lett.*, vol. 17, no. 6, pp. 531–534, Jun. 2010. 994
- [33] I. S. Gradshteyn and I. M. Ryzhik, *Table of Integrals, Series and Products*,  
995 6th ed. San Diego, CA, USA: Academic, 2000. 996
- [34] S. Kim and J. Heo, "Outage probability of interference-limited amplify-  
997 and-forward relaying with partial relay selection," in *Proc. IEEE VTC*,  
998 Yokohama, Japan, May 2011, pp. 1–5. 999



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AQ2 = Equations (29) and (30) are missing in the document. Please check.

AQ3 = Please provide publication update in Ref [22].

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AQ6 = Please confirm that Dr. Zou has received two Ph.D. degrees.

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